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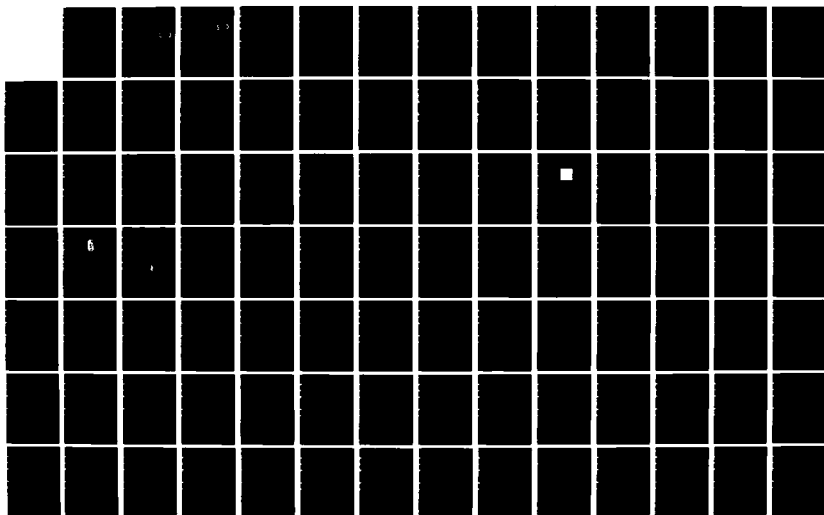
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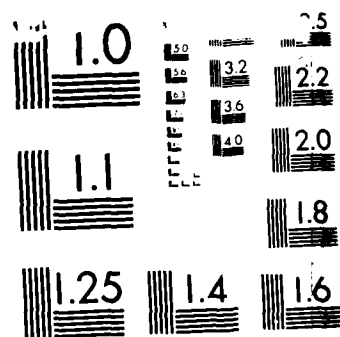
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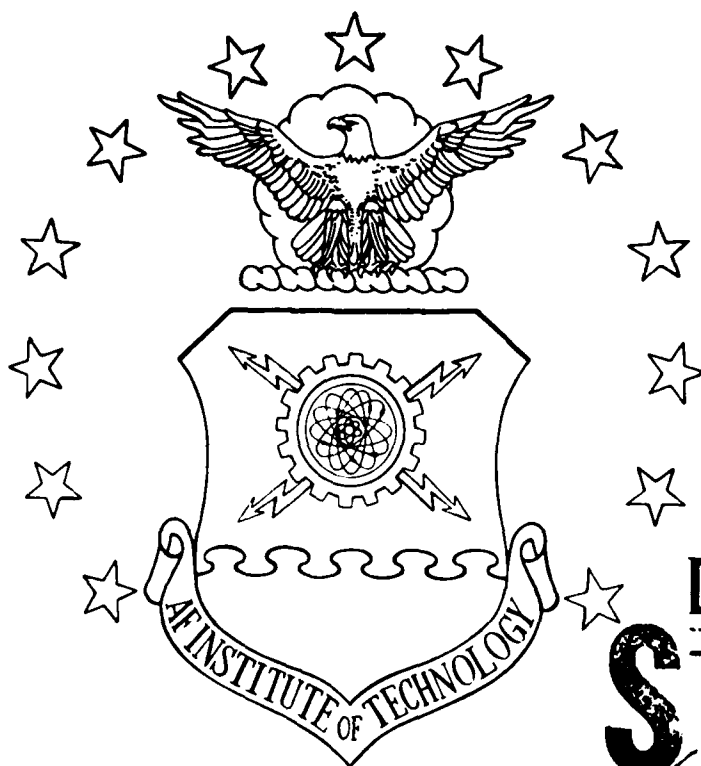




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THESIS

Roland J. Bloom
Captain, USAF

William J. Ramey, Jr.
Captain, USAF

AFIT/GA/GE/ENG/85D-33

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FOR A MOBILE AUTONOMOUS ROBOT

THESIS

Presented to the Faculty of the School of Engineering
of the Air Force Institute of Technology

Air University

In Partial Fulfillment of the
Requirements for the Degree of
Master of Science in Astronautical Engineering
and Electrical Engineering, Respectively

Roland J. Bloom, B.S.A.E
Captain, USAF

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Captain, USAF

December 1985

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Preface

It is only a matter of time until the autonomous mobile robot becomes a reality. The key to achieving this goal lies in the development of a navigation system capable of accurate position determination and intelligent, efficient, collision-free, path planning through the robot's environment. Hopefully, our efforts have provided some advancement toward creating such a robot navigation system.

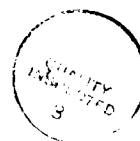
The success of this thesis was a direct result of the endless support provided by several individuals and organizations. A special thanks goes to our thesis advisor Dr. Matthew Kabrisky for having the confidence to turn us loose on this project. Additionally, we would like to thank our sponsor, Tim Anderson of AFMRL; Robert Durham, Orville Wright, Dick Wager, and Stan Bashore of AFIT/ENG; Carl Short and Ron Ruley of AFIT/RMF; Mrs. Allis Moore of AFWAL; Allen Cooper, Berny Swagert, and George Kelsh of King Radio Corporation; Diane White, Ed Freedman, and Cal Watson of Analog Devices Incorporated; Bill Lee and Roger Heilman of Sundstrand Data Control Incorporated; and a generous thanks to King Radio Corporation for their donation of equipment to the project.

We would also like to express our appreciation to Tom Clifford and Bert Schneider for creating the MARRS-1 robot and putting together the robot laboratory. Their efforts made our endeavor possible. In addition, we owe Tom

Clifford a special thanks for his assistance and advice throughout the project.

We would also like to extend our gratitude to Lt. Col. Dan Biezad for his tireless assistance from the very beginning. To him we credit the aquisition of our directional gyro system.

Most of all we wish to acknowledge the many sacrifices made by our wives and children. Without their patience, love, and understanding we could not have completed this thesis.



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Abstract

A navigation system for a mobile autonomous robot is presented. The navigation system is based upon a directional gyroscope and a single axis accelerometer which enables a robot to navigate independent of wheel optical shaft encoders and other commonly used positioning apparatus. The computer controlled navigation system is capable of providing absolute heading, heading rate (angular velocity), and linear velocity to a user computer. These data from the navigation system (heading and velocity) are used to compute the present location of the robot. In addition, the heading data is used to form a closed loop feedback control system for maintaining the robot on a desired course. The navigation system was designed specifically for application on an existing Air Force Institute of Technology (AFIT) robot; however, it could be easily adapted to any robot system with a standard IEEE RS-232 serial communication interface. Test results are provided which demonstrate the use of closed loop heading control on the AFIT robot and which identify problems associated with the use of an accelerometer system for distance measurement. This thesis includes all schematics, parts lists, software listings, and operating instructions for the navigation system. A new robot world modeling and path planning technique is also presented.

GYRO AND ACCELEROMETER BASED NAVIGATION SYSTEM FOR A MOBILE AUTONOMOUS ROBOT

I. Introduction

BACKGROUND

In general, robots fall into three broad categories: fixed arm robots, mobile robots, and mobile autonomous robots. It is important to understand how these three classes of robots differ.

A fixed arm robot is a machine modeled after the human arm and hand. It is capable of armlike movement and has a handlike manipulator, but by itself is incapable of movement from one location to another. Fixed arm robots are by far the largest category of present day robots, encompassing virtually all of the currently available robots. They are used primarily for repetitive manipulative tasks such as industrial assembly line work. Because these robots do not in general possess the ability to move themselves about their environment, they will not be discussed further.

A mobile robot is "a robot mounted on a moveable platform" [11:17]. They are distinguished by their ability to move freely about their environment, but with command and control provided external to the robot. Underwater salvage robots provide an excellent example of mobile robotics. They can move freely about their environment, but are

controlled from a surface vessel by a team of human operators. Since mobile robots are not self controlled, they too will not be discussed further.

A mobile autonomous robot is "a robot acting independently and of its own volition" [11:17]. They are distinguished by their ability to move freely about their environment completely under internal control independent of external machine or human assistance. Mobile autonomous robots have many potential applications, but currently they do not exist outside of research labs. Expansion of this concept is the basis of this thesis.

In recent years, there has been a substantial increase of interest in mobile autonomous robots. A report regarding the First World Conference on Robotics Research, held in early 1985, noted that one of the most actively researched fields was mobile autonomous robots [11:17]. This surge of interest is a direct result of the "microelectronics revolution" which has resulted in today's microprocessors and digital integrated circuits (IC's). Tremendous computational power is now available in very small packages making the internal control required for a mobile autonomous robot feasible.

The potential to develop a mobile autonomous robot has generated many possible applications. Chief among these is the performance of tasks hazardous to human personnel such as fire fighting, bomb disposal, nuclear waste disposal,

underwater salvage and repair, deep sea exploration, chemical production, mining, sentry duty, and military reconaissance and attack missions.

This effort is being led by the Department of Defense (DOD). The Defense Advanced Research Projects Agency (DARPA) currently has a 17 million dollar contract with Martin Marrietta Corporation to develop an autonomous land vehicle [21] and has asked General Dynamics and others for proposals on battlefield robots [13:48]. DARPA is also funding multi-legged mobile robotics research at Ohio State University [13:48].

The Army's General Officer Steering Committee for Artificial Intelligence and Robotics recently issued a report outlining near-term (1980's) robotic systems applications [14]. They include: a light weight robotic vehicle to mount antitank guided missiles, mortars, and small arms; a robotic obstacle/mine breaching tank; a robotic transport and resupply vehicle; a security sentry robot; an explosive ordnance disposal robot; and a robot scout vehicle.

The Air Force Medical Research Laboratory at Wright-Patterson Air Force Base (WPAFB) is currently working on the development of a mobile autonomous robot to service aircraft in a nuclear, biological, or chemical (NBC) contaminated environment. This thesis is an extension of that effort.

All of the above applications require the robot to be

able to perform three primary functions: computation, manipulation, and navigation.

Computation is the robot's ability to make job application related control decisions. For example, an aircraft servicing robot would have to decide which subsystems to test and then which part to replace. In the field of artificial intelligence, this would be called an expert system. Since numerous expert systems of the type required for robotics have already been developed, this function will not be discussed further.

Manipulation is the robot's ability to skillfully handle and move objects in its environment. This function has been well developed and is commercially available in the form of a fixed arm robot. Therefore this function also will not be discussed further.

Navigation is the robot's ability to direct itself toward some destination. This problem has yet to be adequately addressed and is the major impediment to the development of a mobile autonomous robot. Robot navigation is the focus of this thesis.

Navigation requires that the robot be able to follow a given or calculated course of travel making course corrections as necessary. This means the robot must be capable of finding its current location in reference to the desired course. This information is required for steering control feedback in order to accurately follow a given

course.

In addition, the robot must know its starting location and final destination on the same frame of reference as its current location. This will give it the ability to know when it has arrived at the desired location.

Finally, a navigation system must provide a means for recognizing obstacles, both known and unknown. It must allow for course changes to avoid these obstacles and yet still arrive at the destination in a time efficient manner. A robot must be able to deal with a dynamic environment.

Physically, the navigation system is composed of sensors and steering control. The sensor subsystems collect position and obstacle data. The steering control subsystem uses the position and obstacle data to make any necessary course changes. All but the most simple steering control subsystems require many intelligent decisions to be made and therefore require a computer as the controller.

Past research work on mobile robot and mobile autonomous robot navigation has produced vision, embedded wire, beacon, shaft encoder, and sonar based navigation systems, where the environmental data from each system's sensors is interpreted by a computer to control the path of travel.

Vision navigation systems use television cameras to "see" the desired path and obstacles much like a human. Vision systems today are relatively crude in spite of modern

advances in high speed computer systems. They are not capable of interpreting the digital video information fast enough to allow real time navigation.

Dr. Hans Moravec has done considerable research in the area of robotic vision. During his Ph.D. work at Stanford University, he experimented with the "Stanford cart", a mobile robot which he equipped with a video camera [17]. The cart's ultimate achievement was travelling through an obstacle laden area to a goal about 60 feet away. The trip took a total of about five hours. In addition, the cart did no onboard processing of the video images. They were transmitted by radio link to a VAX 780 computer which interpreted the data and issued the resulting steering commands. This reduces the robot to the mobile category as opposed to the mobile autonomous classification.

Dr. Moravec, now a professor at Carnegie Mellon University, is still diligently pursuing his research into creating human like vision for a mobile robot. However, he admits that to reach his goal he needs a computer which is about 1000 times faster than his current computer [9:30]. Present vision systems are not fast enough to allow real time navigation and the computers are not small enough to be built on board the robot for a mobile autonomous system.

Embedded wire systems find the desired path by following wires on or beneath the travel surface like a train on railroad tracks. This system is very simple,

requires little computational ability, and navigation can be done in real time. However, it is not very flexible. Places of travel are limited to where the wire tracks are located and continuity of the wire from start to destination. Obstacle avoidance is impossible without additional sensors and even then would be limited to stopping and waiting for the obstacle to move.

Beacon systems determine the desired path from position reference information broadcast by beacons found near the path of travel. This system is also simple, easy to implement, and provides a real time navigation capability. It suffers from the need to provide beacons in the robots working environment for navigation. Additional sensors are required to provide an obstacle avoidance capability.

Optical shaft encoder systems rely on "dead reckoning" navigation by starting at a known position and moving precise distances as measured by the optical shaft encoder. This is also a relatively simple system, easy to implement, and can be done in real time. It is more flexible than the embedded wire system because it is not limited to following wire tracks, but it suffers from errors in sensor data. The distance moved data must be very accurate or error will be introduced that is cumulative and will eventually result in the robot becoming lost.

Optical shaft encoders have proven to be fairly accurate under ideal conditions. However, several sources

of position error accompany their use. These position errors are unbounded and cumulative. One error source is wheel slippage, resulting in an inaccurate measurement of actual distance traveled.

A second problem lies in knowing the exact circumference of the wheel. The distance traveled by the robot is determined from the number of revolutions made by the wheel; one revolution of wheel travel equals the distance of one circumference in linear travel. As the tire wears out, the error in the measured distance grows larger.

A third problem results from computational errors. This error is most prominent when the robot has negotiated many turns, requiring numerous calculations based on geometric approximations to determine the current location of the robot. After a short period of travel, the robot will begin to stray off course due to an accumulation of error. The performance of optical shaft encoders degrades considerably as the number of maneuvers performed by the robot increases.

A fourth problem with optical shaft encoders results from not having an absolute heading reference. All heading computations from encoder data produce a relative heading to some initial heading. Initial heading of the robot must be externally provided and can itself be a very large error source. In Figure 1.1, the effect of initial heading error is illustrated. Without an accurate initial heading, the robot will eventually become lost. A robot system using

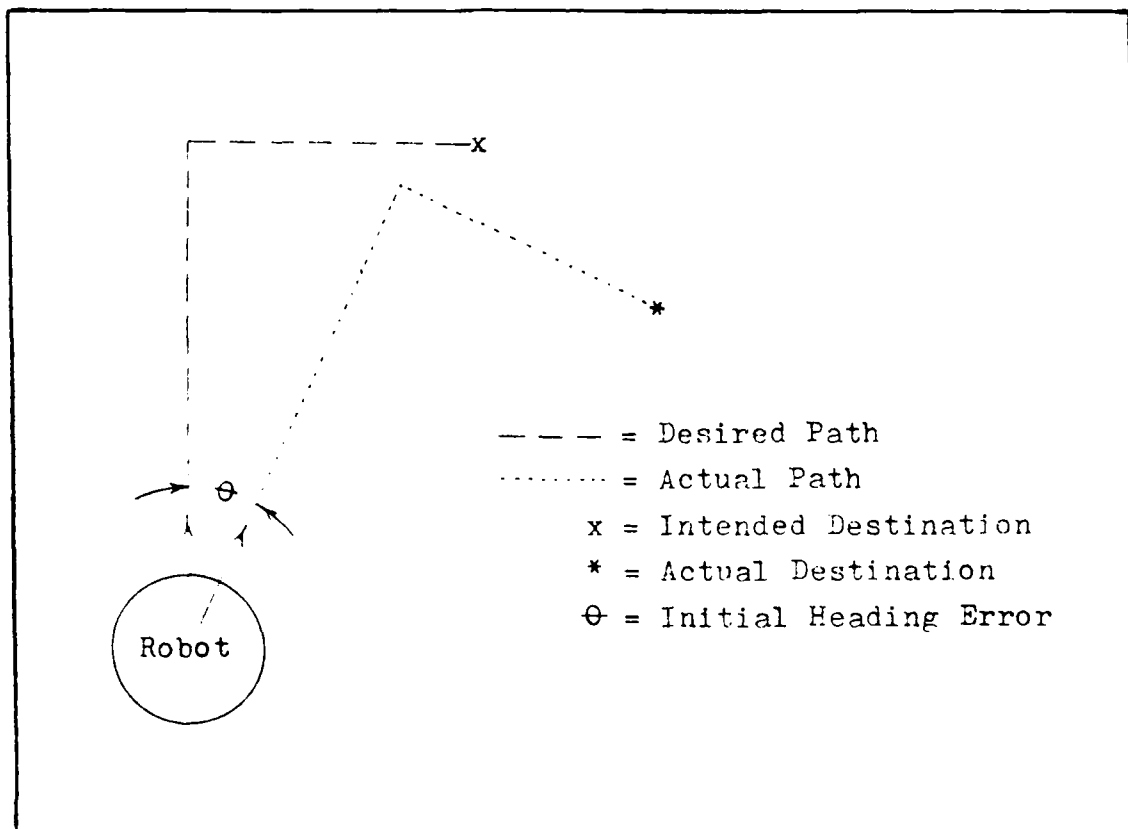


Figure 1.1. Effect of initial heading error.

optical shaft encoders for navigation is highly dependent on this initial heading reference.

Finally, optical shaft encoders offer no means for detecting obstacles. Therefore, additional sensors are required to provide an obstacle avoidance capability. Sonars (ultrasonic ranging units) can be used to complement the optical shaft encoders and together provide position and obstacle detection information. In addition, the sonars can be used to compensate for errors in the encoders by providing range information to known obstructions. In other

words, the sonars can provide a position "fix" so the robot knows its location with respect to its known environment.

Dr. James Crowley [12] is experimenting with just such a system. He uses a focused rotating ultrasonic ranging device to maintain a description of the robots external environment. This description is called a sensor model. The estimated position of the robot, obtained from optical shaft encoders, is compared to the position of the robot determined from the sensor model to create a composite model. This composite model represents an average position of the robot. In this way, Crowley seeks to maintain an accurate estimate of the robots position.

It is not very feasible to use sonars as the only source of navigation data since their range is limited (making navigation through a large open area impossible) and unknown obstacles could be misinterpreted as known obstacles causing the robot to become lost. It is also important to note that sonar data is useless for navigation without an accurate heading reference. A detected object cannot be used as a reference without knowing the direction of the object in the navigation reference frame.

Previous thesis work at the Air Force Institute of Technology (AFIT) has produced a mobile robot, the Mobile Autonomous Robot Research System-1 (MARRS-1), that can map its environment and provide obstacle detection with sonar sensors. It can also measure distance travelled with

optical shaft encoders. In addition, the robot has an onboard computer for dedicated navigation calculations. However, MARRS-1 is not capable of autonomous navigation. The current navigation system can only traverse a course preprogrammed into the drive computer through the robot's keyboard or by use of the teaching pendant. It can only collect, not use for navigation, sonar obstacle information and shaft encoder distance data (as presently programmed). In addition, it is not able to compensate for errors in wheel direction and distance measurements, which may cause MARRS-1 to wander from the programmed course (open loop steering control).

It is obvious that past robot navigation research, both at AFIT and abroad, has not produced an accurate, real time, autonomous, mobile navigation system. A new approach is indicated by the weaknesses and liabilities of past systems. First, an absolute and directly measurable robot heading reference is required. Second, an accurate and preferably non-mechanical method of distance measurement is necessary. Third, an accurate, long distance (5 to 50 feet or more) method of obstacle detection is required. Finally, a navigation algorithm is required that models the sensor data into a real world map and issues robot steering commands based on this mapping.

This thesis attempts to solve these problems by adapting an aircraft directional gyroscope system for the

heading reference, an accelerometer for distance measurement, and sonars for obstacle detection. Robot "world modeling" and path planning is discussed, but not implemented.

PROBLEM AND SCOPE

The problem is to design and fabricate a real time, point to point, closed loop, mobile autonomous robot navigation system for the MARRS-1 robot. In addition, it must be capable of detecting and avoiding both previously known and unknown obstacles.

Real time navigation is defined to be navigation at the maximum constant motion travel speed of MARRS-1. Point to point refers to navigation from a given starting point to a given destination and includes in-course obstacle detection and avoidance. Closed loop refers to the ability to detect and correct course errors. Autonomous, in this case, will be broadened to mean needing no external support except power.

Design and fabrication will include: construction of a new third body tier to house the gyro and accelerometer based navigation system (GYRAC); physical and electrical modification of MARRS-1 to allow GYRAC integration; fabrication of a GYRAC control computer; implementation of GYRAC computer software; fabrication of a digital electronic interface between the gyro/accelerometer and GYRAC computer;

integration of the MARRS-1 drive and navigation computers with the GYRAC computer; a simple point to point navigation control program (no obstacle avoidance) to demonstrate course tracking; full testing of all new hardware and compilation of test results; complete schematic diagrams and operating manuals for all new hardware; and fully documented software listings for all operational and test programs.

Obstacle detection and avoidance were not implemented, but are available since this has been demonstrated on a previous thesis [19]. Finally, the design options available were limited by hardware and software restrictions imposed by systems previously added to MARRS-1, limited space internal to the robot, limited time, and insufficient funds.

Assumptions

Several basic assumptions are necessary for a navigation system based on a directional gyroscope and an accelerometer. This thesis is predicated on the following assumptions:

1. Assume no local disturbances will be present in the earth's magnetic field. This assumption is necessary since the directional gyro output is slaved to a magnetic flux detector for absolute reference to magnetic north.
2. Assume the operational environment of the robot is a perfectly smooth and level surface. This assumption is necessary since a single accelerometer cannot distinguish true acceleration from local gravity. Therefore, any tilt of the accelerometer input axis into the vertical plane will induce an error in measured acceleration.

3. Assume a perfect integrator. An operational amplifier integrator circuit is used to integrate accelerometer output to obtain velocity.
4. Assume velocity of the robot is constant over sample period. This is necessary for simplicity and ease of calculation.
5. Assume sample time is known precisely. Using accelerometer output to ultimately obtain distance traveled by the robot is a time dependent problem.

All the above assumptions, excluding the first, are tied directly to the use of an accelerometer. Each of these assumptions will be addressed again in Chapter V.

Evolution and Capability of MARRS-1

The MARRS-1 began as a Heathkit HERO-1 robot. This original HERO-1 has since undergone substantial modification and today bears little resemblance to its former self. Lieutenant Owen [19] was the first to modify the original HERO-1. He added a laser barcode scanner and several Polaroid ultrasonic ranging units to the robot. The scanner was used to determine the location of the robot by reading barcodes taped to the floor. The first step had been taken toward creating an autonomous mobile robot. The HERO-1 could collect useful but limited navigation data on its own.

Follow-on work by Clifford and Schneider [10] was a major leap toward the goal of creating a mobile autonomous robot. Under their thesis effort, the HERO-1 was completely rebuilt. The following is a list of the major modifications performed by Clifford and Schneider:

1. A new main body was fabricated. This new body is 12

sided and consists of two separate levels. The top level can rotate with respect to the lower level.

2. The HERO-1 CPU (MC6808) was upgraded with the addition of a Virtual Devices Inc. MENOS-1 MC6801 CPU board which added RS-232 serial communication capability to the HERO-1.
3. Two dozen Polaroid sonar transducers were attached to the new robot body (one on each segment of the upper and lower body decks).
4. Optical shaft encoders were placed on the two rear wheel shafts and on the front (drive) wheel steering shaft to provide distance moved data.
5. An MC6802 based computer was added to control the optical shaft encoder subsystem and the sonar subsystem. This computer is called the Navigation computer.

The resulting heavily modified HERO-1 was renamed MARRS-1. MARRS-1 was then used to generate sonar range data and wheel distance data. This data was post processed on an external computer to create a composite map of the robot's local environment relative to the robot (whose position was determined from the shaft encoder data). This represents a significant improvement over the capability of the Owen modified HERO-1. However, like Owen's modified HERO-1, the MARRS-1 served only as a collector of navigation data. The actual movement of the robot was controlled by a human operator. MARRS-1 was not yet an autonomous robot, but it possessed the capability to become one.

The capability of the MARRS-1 at the beginning of this thesis can be summarized as follows:

1. All functions of the original HERO-1 still existed, except for the arm [10:III-3]. This capability included programable movement of the robot.

Unfortunately, this programed motion was open loop, resulting in unreliable and nonrepeatable movement. The robot could not follow a straight line path.

2. The laser barcode scanner was removed but all provisions for attaching it (both mechanically and electrically) to the robot still existed.
3. Sonar and optical shaft encoder data could be gathered by the navigation computer and relayed via an RS-232 computer interface to an external computer for storage on a floppy disk. Firmware was resident onboard the Navigation computer to obtain and transmit the data.
4. Four RS-232 computer interface ports were available for use. One port allowed communication with the main CPU (MENOS-1 upgrade board). The other three ports allowed access to the navigation computer.

General Approach

Development of a mobile autonomous robot navigation system for the MARRS-1 required a rigorous analysis of past and present robot navigation research literature. Based on this literature search, a new approach to world modeling for mobile autonomous robots was developed.

This new navigation approach required collection of specific types of data. The current systems and sensors on the MARRS-1 were compared against these new requirements and a list of sensors and systems to be constructed were compiled. A prime requisite during this analysis was to limit the data collected so that onboard computers could process the data fast enough to allow real time navigation.

The required sensors and subsystems were built and tested as was the necessary software to drive each subsystem. Modifications were then made to the MARRS-1

system to allow the GYRAC subsystem to be physically, electrically, and functionally (through software) integrated onto the MARRS-1 platform. Navigation system testing was performed and supporting data collected to validate the new navigation system.

Finally, MARRS-1 system software was developed, tested, and support data collected to demonstrate the feasibility of gyro based steering control. Conclusions and recommendations are provided based on testing at each of the various stages.

Sequence of Presentation

Chapter One provides a detailed problem background, problem statement and scope, assumptions, evolution of MARRS-1, general approach, and sequence of presentation. Chapter Two covers the GYRAC system design and theory of operation. Chapter Three discusses GYRAC system integration onto the MARRS-1 platform. Chapter Four presents a new mobile autonomous robot navigation theory for world modelling and path planning. Chapter Five discusses testing results and analysis of the various stages of the GYRAC development. Chapter Six gives a system summary and presents conclusions and recommendation. The appendices contain hardware data sheets, schematic diagrams, device layouts, wiring diagrams, connector diagrams, an equipment list, and a parts list. Included are software structure

charts, program listings, collected data tables, and software operating instructions.

II. GYRAC Design and Theory of Operation

The hardware design goal of this thesis is to create a navigation data system capable of providing absolute heading and velocity data, in binary digital form, to an external computer. Three primary requirements are the basis for this design:

1. A KCS-55A Pictorial Navigation System (produced by King Radio Corp.) was ordered after selection as the best directional gyro system for this project and a QA-1100 accelerometer (produced by Sundstrand Data Control Inc.) was already on hand at AFIT. Therefore, the system must be centered around the KCS-55A Pictorial Navigation System and the QA-1100 accelerometer.
2. The navigation data system must be compatible with the MARRS-1 but also easily transportable to another robot system.
3. The system must be attainable with resources readily available to AFIT.

Since the foundation of the system is a gyro and an accelerometer, the navigation data system will hereafter be referred to as the "GYRAC" system (for gyro and accelerometer system).

Summary of GYRAC System

A general layout of the GYRAC system can be seen in the block diagram in Figure 2.1. The gyro subsystem provides analog heading information which is then converted to a 12 bit TTL (Transistor Transistor Logic) digital data signal. Angular velocity is a byproduct of this conversion in analog signal form. Thus, the analog angular velocity is digitized

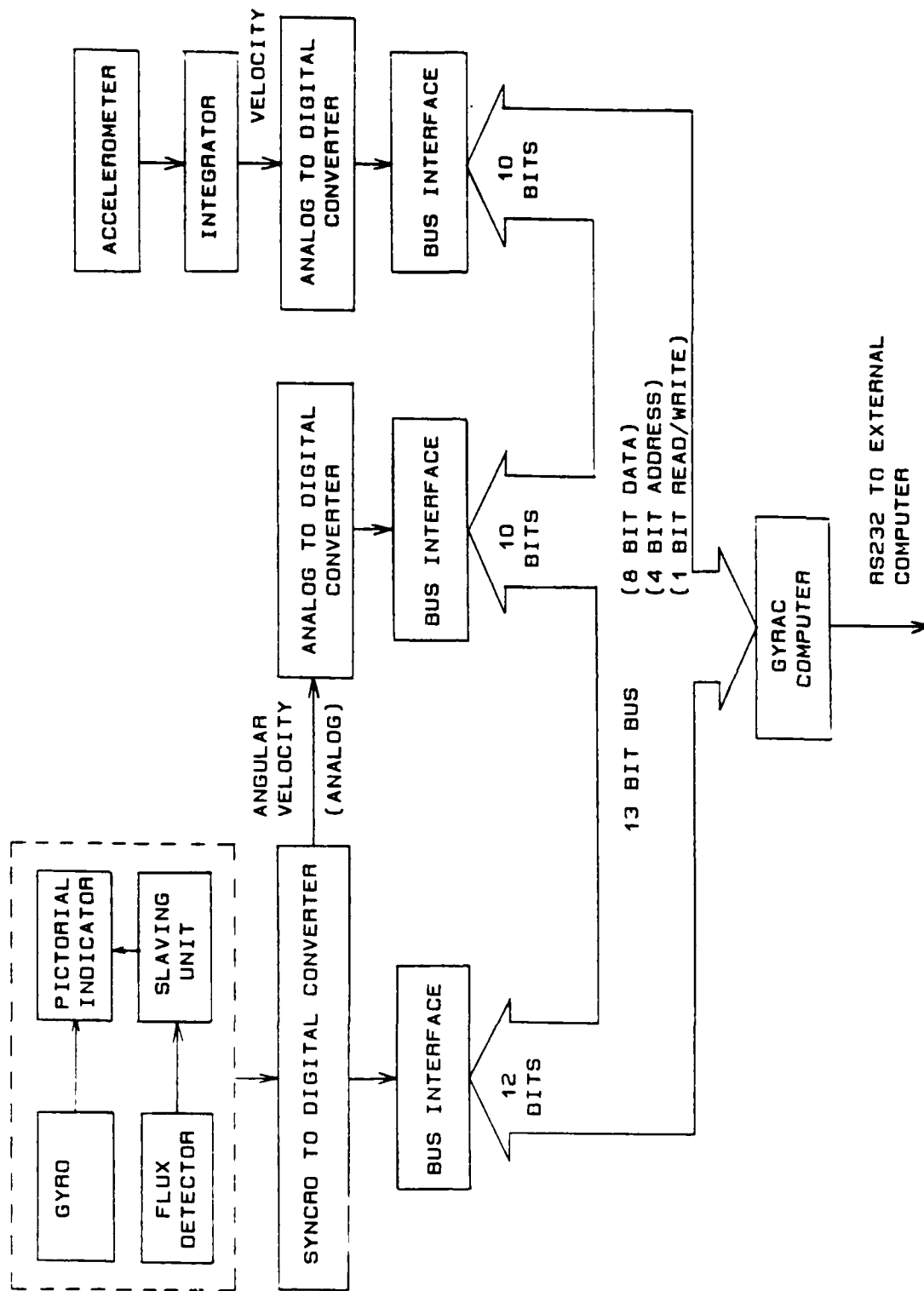


Figure 2.1. GYRAC System Block Diagram

through an analog-to-digital converter resulting in a 10 bit digital data signal. The accelerometer produces an analog signal which is integrated via an operational amplifier circuit and converted to digital TTL data by an analog-to-digital converter. The three digital data signals are connected to a common thirteen wire bus containing eight data lines, four address lines, and a read/write line. This bus serves as a standard sensor interface to the GYRAC computer. The GYRAC computer interprets commands received from an external computer via an RS-232 serial data link and acquires the appropriate sensor data as directed by the external command. The GYRAC computer then performs any necessary preprocessing of the data, converts it to serial format, and transmits it to the external computer via the RS232 link.

The Gyro Subsystem.

The gyro subsystem is composed of four major elements: a directional gyro; an indicator unit; a magnetic flux detector; and a slaving unit. The directional gyro provides a gyro stabilized magnetic heading to the indicator. The directional gyro consists of two primary components: the gyro itself and a base assembly, see Figure 2.2.

The gyro is a spinning mass precision gyro with two degrees of freedom, see Figure 2.3. Relative angular displacement is sensed by an optical encoder assembly mounted on the gyro's outer gimbal. Electrical outputs from

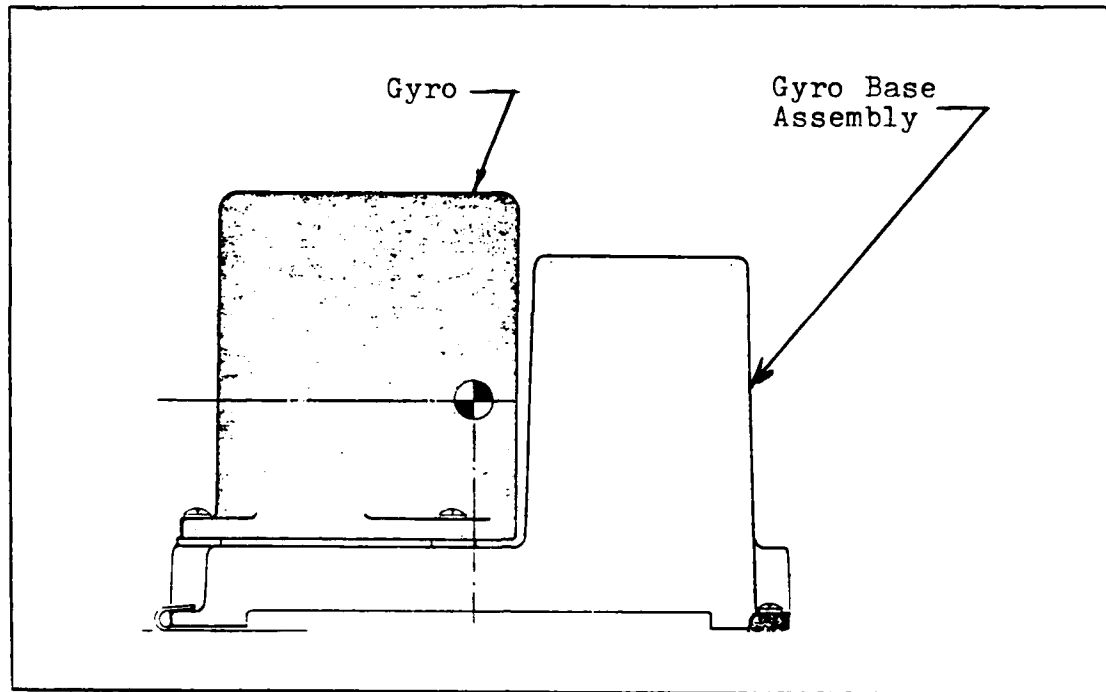


Figure 2.2. Directional Gyro [1:2-11]

the optical encoder are two square waves which are used to drive a stepper motor in the indicator. The gyro base assembly contains the control logic for the gyro and the slaving logic for the indicator. The gyro base also serves as the power supply for the entire gyro subsystem. From the single 28 volts DC input into the gyro base, the following internal voltage supplies are generated: 26 VAC 400 Hz for the gyro spin motor, flux detector excitation, and heading synchro excitation; + and - 15 VDC regulated supply for the analog circuitry in the system; +15 VDC unregulated voltage for the stepper motor in the indicator; and +5 VDC regulated supply for the system digital logic circuitry [3:5-33].

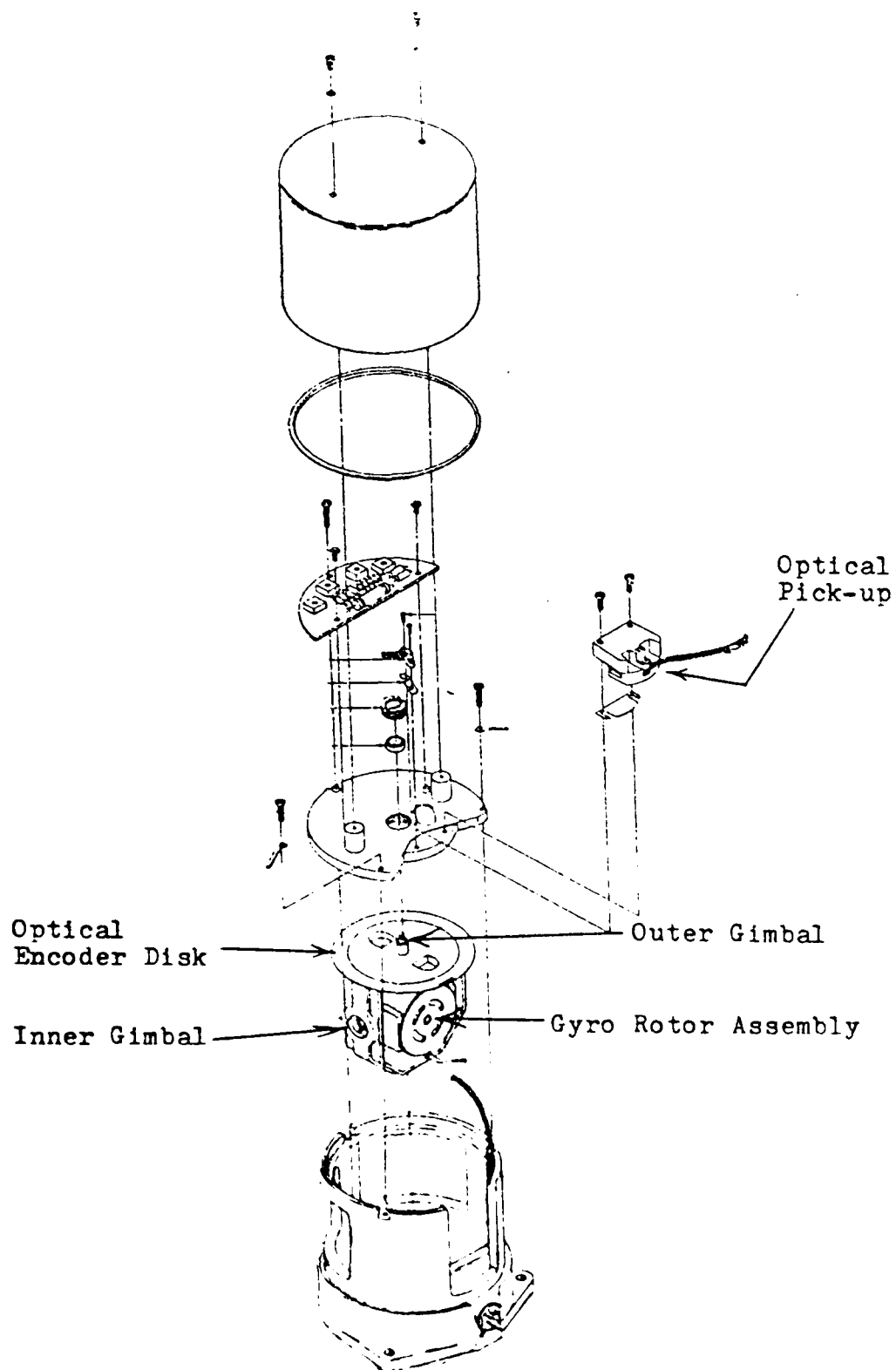


Figure 2.3. Gyro, Exploded View [2:5-33].

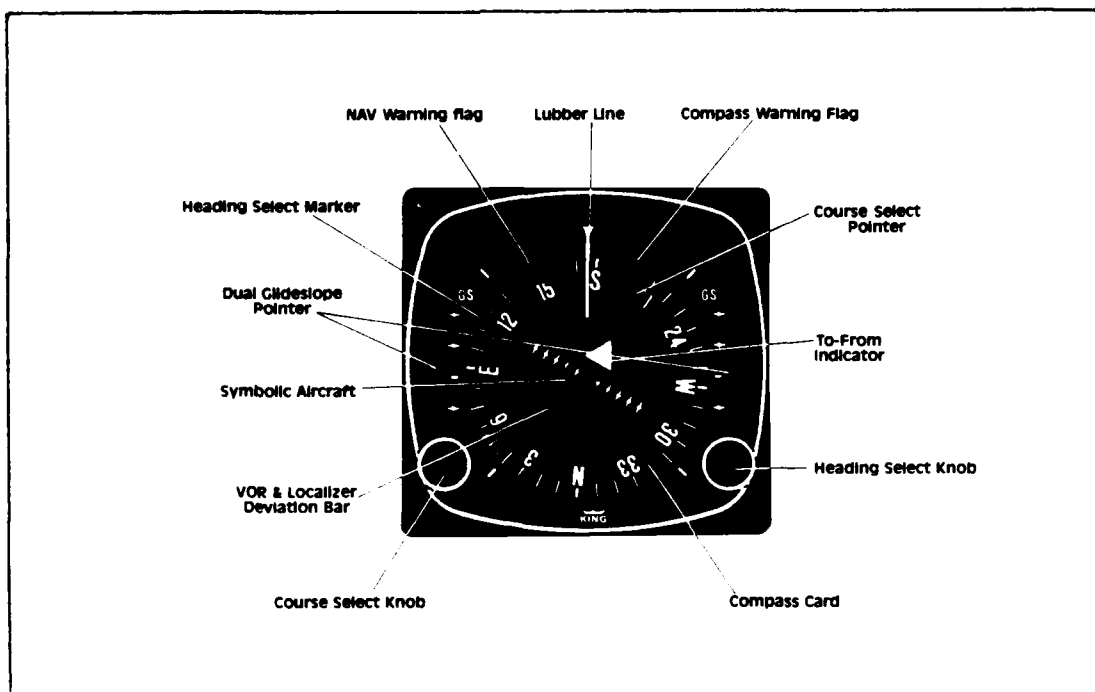


Figure 2.4. The Heading Indicator [1:2-9]

Also, separate regulated grounds are maintained for the analog and digital circuitry. For the GYRAC system, the above mentioned supplies and grounds are routed to a central power distribution panel to provide the necessary power for other GYRAC hardware (see Appendix E).

The indicator is typical of the type seen in the cockpit of small aircraft, as shown in Figure 2.4. A digital stepper motor is used to drive the heading display in response to the signals generated in the directional gyro. The signals from the gyro consist of a two phase excitation drive that is connected to the four stepper motor leads as shown in Figure 2.5. Each time the A or B waveforms (see

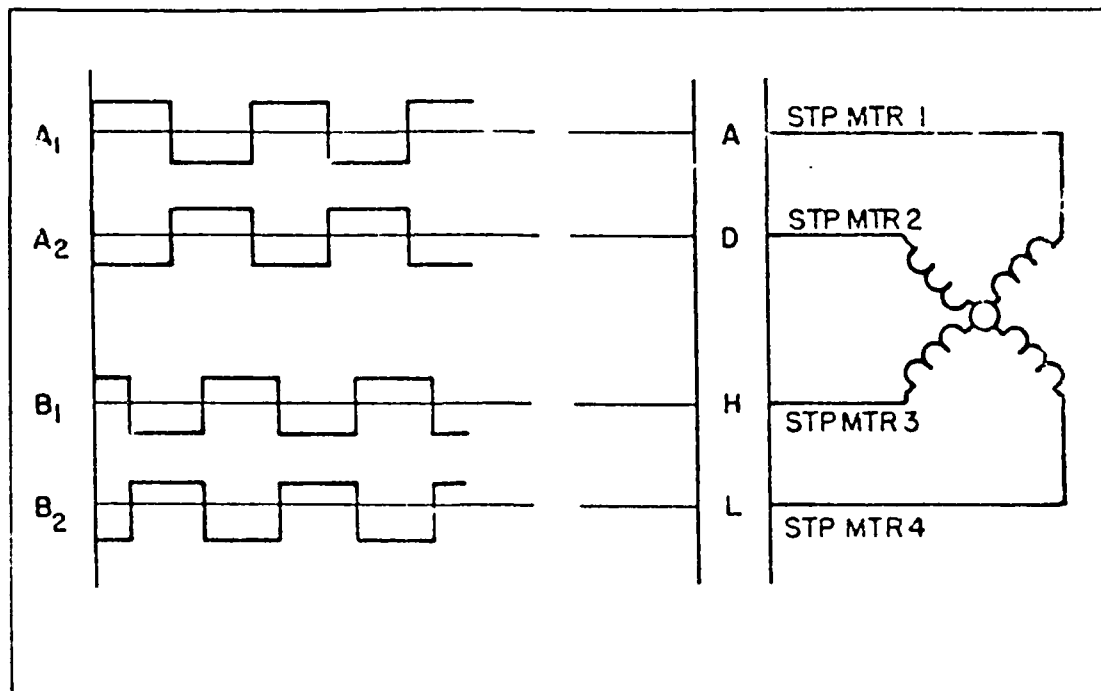


Figure 2.5. Stepper Motor Drive Circuit [2:4-1]

Figure 2.5) change state, the motor shaft moves nine degrees in a direction determined by the previous state of the waveforms. This motion is reduced to a 0.25 degree card rotation by a 36:1 gear train assembly [2:4-1]. Thus, the display card moves in increments of 0.25 degree thereby limiting the resolution of the heading angle to + or - 0.25 degree movement of the indicator display can be tracked by a synchro control transmitter (CX) which is mounted internally on the rear of the compass card shaft (see Figure 2.6). This CX is intended to provide a slaving signal to another display, but can be used to get an absolute electrical representation of heading. This fact is crucial

Syncro Control Transmitter (CX)

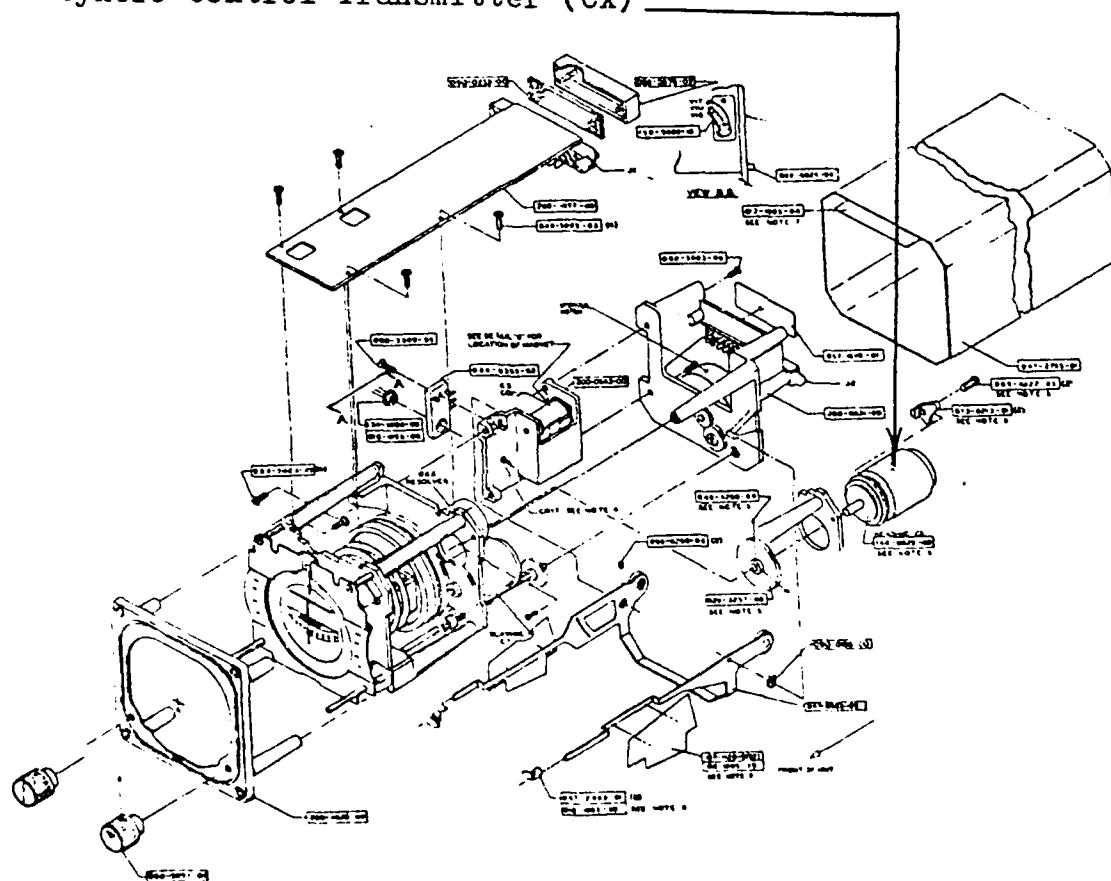


Figure 2.6. Indicator, Exploded View [2:5-5] .

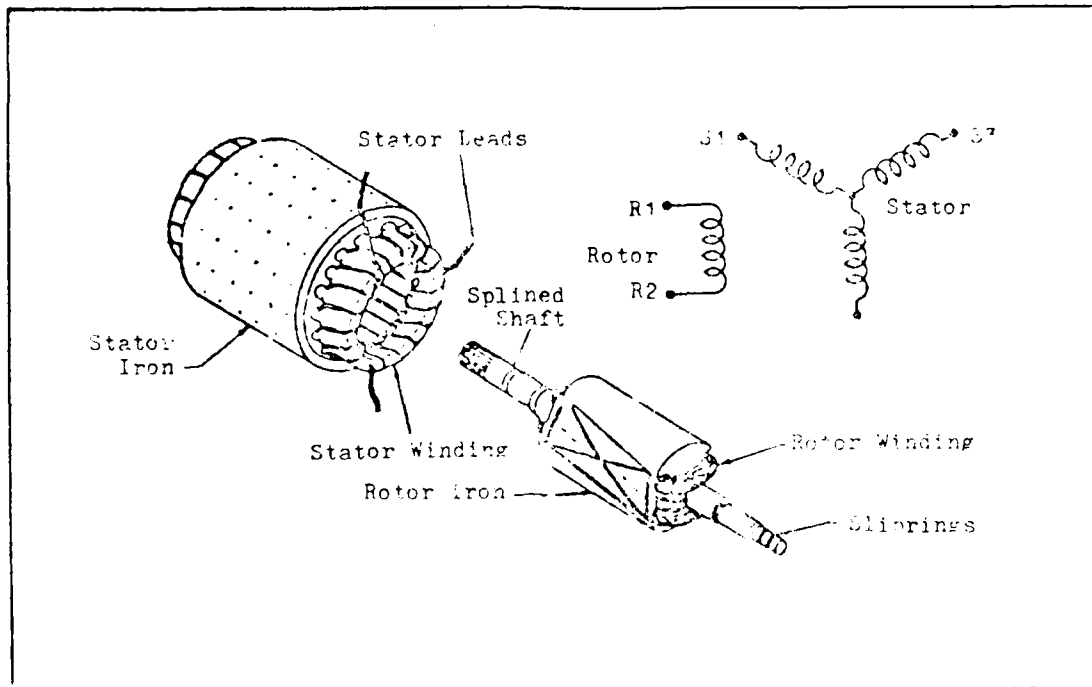


Figure 2.7. Internal Structure of a Syncro Control Transmitter and its electrical representation. [18:2]

to the realization of the GYRAC system. If the rotor of the CX is excited with a reference voltage (AC), then syncro format voltages will appear as output across the S1, S2, and S3 terminals (see Figure 2.7) [18:2]. These voltages are a function of the shaft angle θ . For example, if the rotor (which has a single winding) is excited by a reference voltage across R1 and R2 (see Figure 2.7) of the form:

$$A \sin(\omega t)$$

Then the voltages which will appear across the stator

terminals will be:

$$S1 \text{ to } S3 = A \sin(wt) \sin \theta$$

$$S3 \text{ to } S2 = A \sin(wt) \sin(\theta + 120)$$

$$S2 \text{ to } S1 = A \sin(wt) \sin(\theta + 240)$$

where θ is the shaft angle.

These voltages are known as syncro format voltages [18:2]. A desirable result of the syncro output is that it can be easily converted to a digital signal with a standard syncro-to-digital (S/D) converter. An SDC1700 12 bit S/D converter made by Analog Devices is used (see Appendix A and C) to provide a TTL digital binary representation of the heading angle (shaft angle). The SDC1700 also provides an angular velocity output in analog form which will be converted to digital by an analog-to-digital (A/D) converter. The A/D converter to be used is also produced by Analog Devices and is a 10 bit converter (Part # AD573, see Appendix A and C for detail).

The magnetic flux detector senses the direction of the earth's magnetic field and converts this information to a three-wire syncro format, much like the CX in the indicator. This information is transmitted to the indicator for slaving purposes, see Figure 2.8 for an exploded view of the detector. The flux detector can be oriented so the gyro system displays a heading relative to some artificial North direction. This feature is used to align the GYRAC system

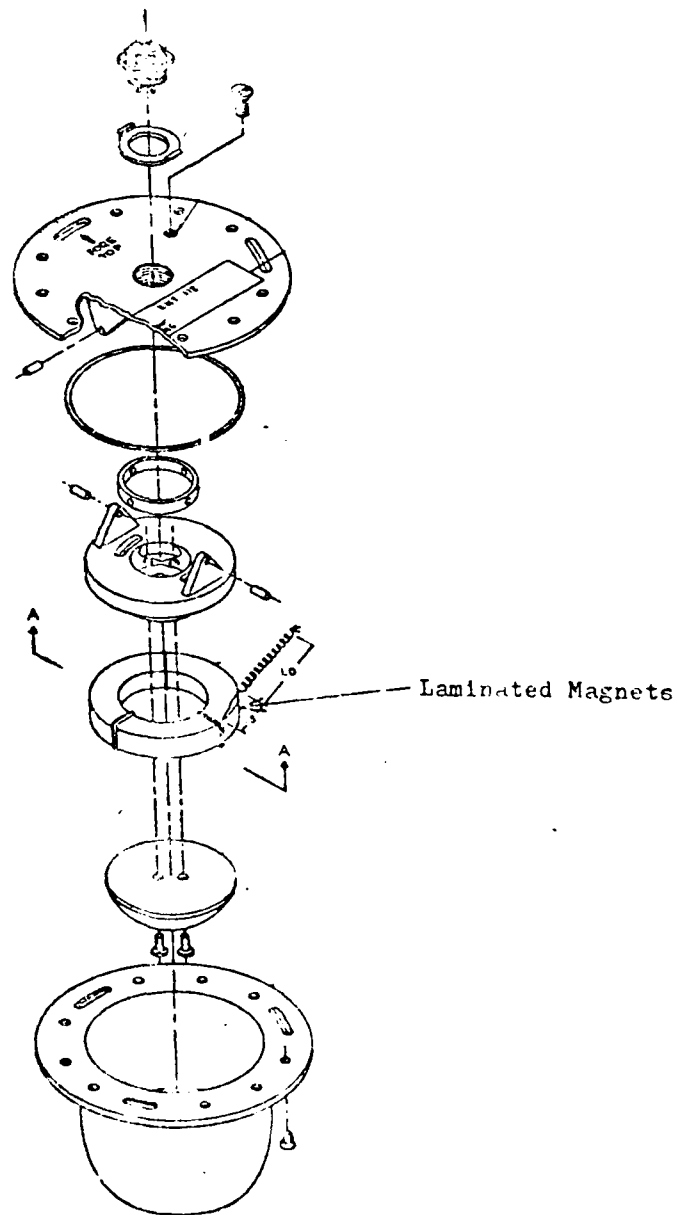


Figure 2.8. Flux Detector, Exploded View [4:5-3] .

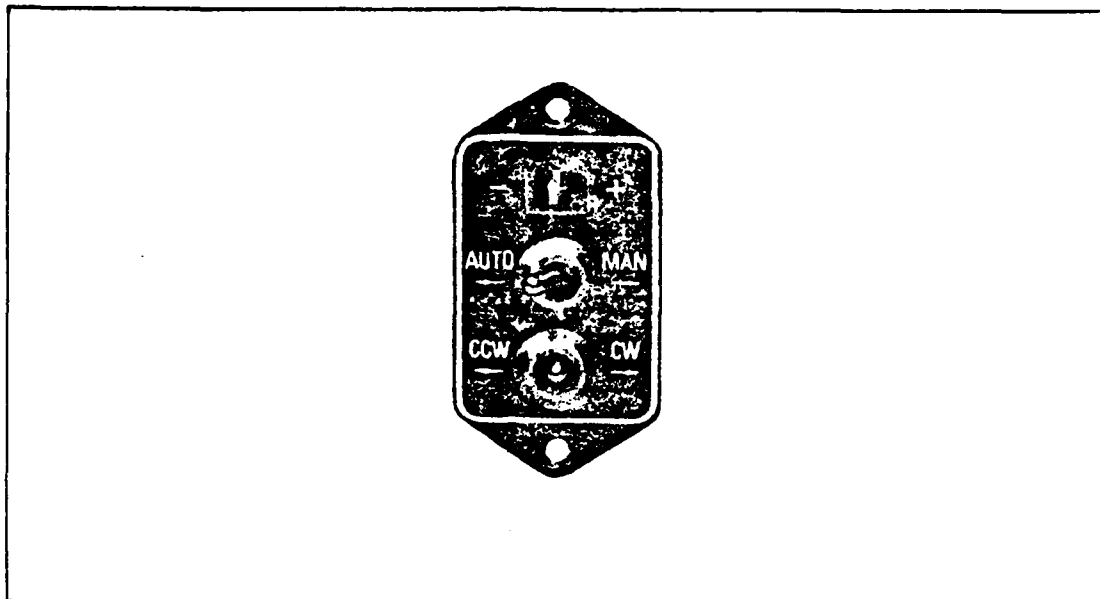


Figure 2.9. Slaving Unit [1:2-19]

with some convenient navigation coordinate system in the test area.

The slaving unit, shown in Figure 2.9, contains a slaving meter, slaving switches, and corrector circuitry which can compensate for the effect of local magnetic disturbances on the flux detector. The meter current is generated in the directional gyro base assembly (slaving logic) and represents the difference between the flux detector sensed heading and the heading displayed on the indicator. The slaving switches allow the gyro system to be operated in either a free-gyro mode (no slaving with the flux detector) or in the slaved mode (automatic slaving with flux detector). There is also a manual slaving switch which

can be used to rotate the display card in the indicator either clockwise or counter-clockwise. In addition to the slaving meter and slave switching functions, the slaving unit also includes a compensation circuit. This circuit causes a shift in the magnetic direction vector and thus can compensate for "hard iron" effects caused by nearby ferrous materials.

The Accelerometer Subsystem.

The accelerometer used in this thesis is a QA-1100 servo-type single-axis accelerometer produced by Sundstrand Data Control Incorporated (see Figure 2.10). The sensor, as

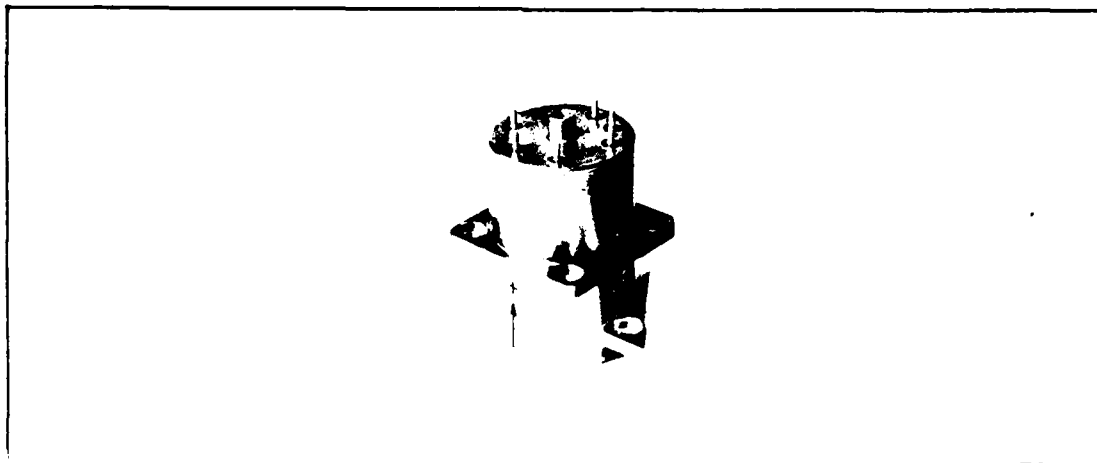


Figure 2.10. The QA-1100 Accelerometer [20]

shown in Figure 2.11, consists of the following key elements [7:1-3]:

1. A proof mass, pendulously supported and ideally constrained so as to allow only one degree of freedom about a well defined axis fixed within the

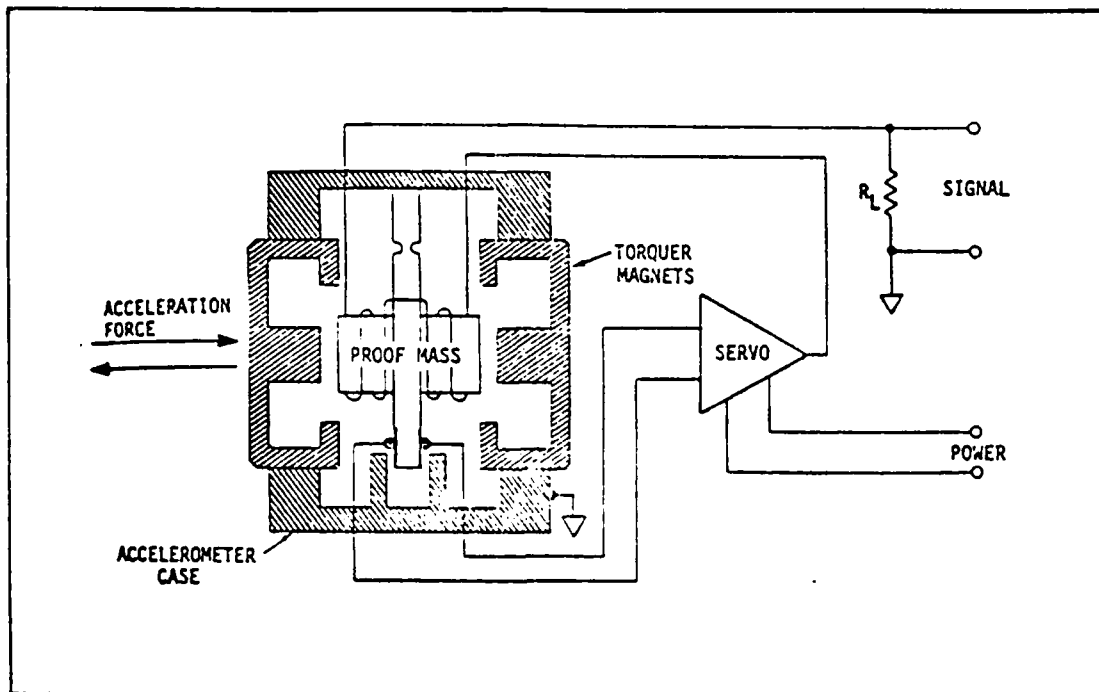


Figure 2.11. Basic Structure of a Servo-Type Accelerometer [7:1-3]

sensor.

2. A pick-off that can sense extremely small displacements of the proof mass.
3. A torquer, which is a coil positioned within a permanent magnetic field and attached to the proof mass, allowing force to be applied to the proof mass in response to a current passed through the coil.
4. A restorer circuit, or servo, that causes an electrical current to flow through the torquer coil in response to a pick-off signal. The resulting electromagnetic force balances the inertial reactive forces.

The basic operation of the accelerometer is that of a linear single axis electro-mechanical device for measuring

acceleration. The operation is based on movement of the proof mass during acceleration. A pickoff senses the displacement of the mass and the servo amplifier develops a current which is supplied through the torque coil to rebalance the proof mass. Thus, the rebalance current is proportional to the sensed acceleration and is a very accurate measure of acceleration. As more acceleration is applied to the accelerometer, the assembly will maintain the proof mass position and rebalance current will increase with increased acceleration until the sensor saturation limit is reached. An exploded view of the actual sensor assembly can be seen in Figure 2.12.

This type of accelerometer does not come with an internal (factory set) load resistor (R_L in Figure 2.11). Thus, an external load resistor must be provided. This is desirable since the ranging, or sensitivity, of the accelerometer output can be chosen to suit specific applications. The value of the external load resistor is determined by the following formula [7:3-6]:

$$R_L = \frac{\text{voltage sensitivity desired}}{\text{current sensitivity of the accelerometer}}$$

The current sensitivity of the QA-1100 is approximately 1.3mA/g (where g is the acceleration due to gravity). For this thesis, the accelerometer is configured to have a

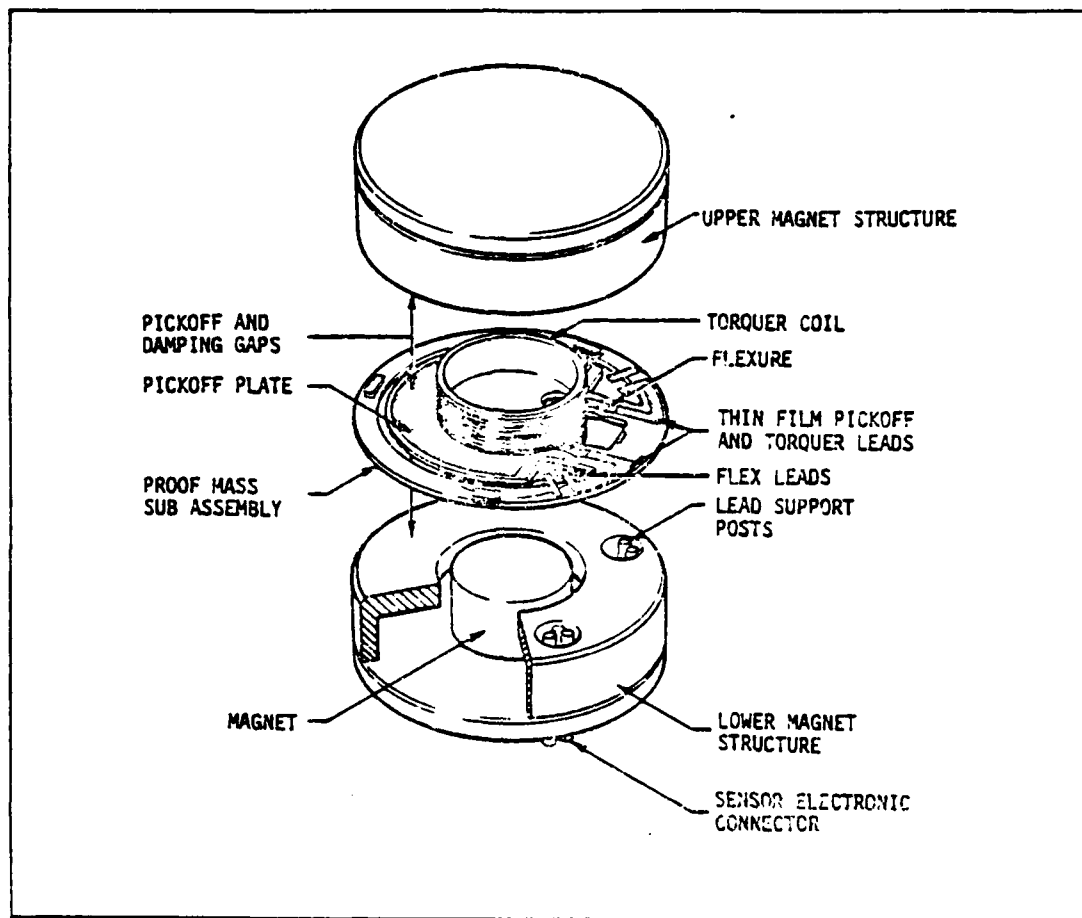


Figure 2.12. QA-1100 Sensor Assembly, Exploded View
[7:1-4]

sensitivity of 2 volts/g. R consists of a single precision resistor in series with a 10 turn trim-pot for fine tuning of the sensitivity (see Appendix B for schematic detail).

As with any measurement device, the accelerometer has a scale factor and bias error. However, without the use of a centrifuge, these values are very hard to determine to a substantial degree of accuracy. Nonetheless, a tumble test can be performed and has been performed. A tumble test

consists of positioning the accelerometer with the input axis exactly vertical, pointing downward and then upward. The sensor will detect the earth's gravity vector. The two measurements (input axis up and input axis down, referred to as V_{OUT}) are then used in the following equation:

$$V_{OUT} = V_{ACT} \times \text{Scale factor} + \text{Bias}$$

Here V_{ACT} is 2 volts (since 2 volts/g is the sensitivity of the accelerometer). Use of this equation results in two equations and two unknowns (scale factor and bias). Preliminary testing of the accelerometer has resulted in a very small value for bias (about 2 millivolts) and a scale factor of very near unity. Thus, for this thesis, the scale factor is assumed to be equal to one and the bias is assumed to be zero. This assumption will be discussed further in Chapter V under Review of Assumptions.

Accelerometers cannot distinguish between gravity and true acceleration. This fact is a major concern in the GYRAC system and must be accounted for. A special mounting platform has been built for the accelerometer allowing for complete leveling. The accelerometer platform will be initially adjusted until a very near zero reading is established from the accelerometer. However, during movement of the GYRAC system it is highly likely that errors will occur in the accelerometer output due to travel over a non-level surface. This problem is discussed further in Chapter

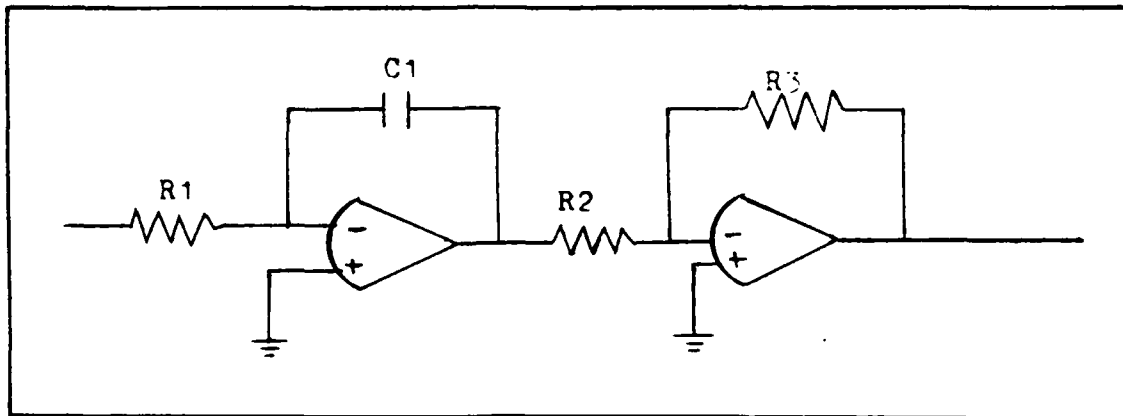


Figure 2.13. Accelerometer Integrator and Scaling Circuit

V under Review of Assumptions.

The output from the accelerometer is connected to an integrator circuit, shown in Figure 2.13. The output, which is velocity, is scaled such that one volt is equal to one foot per second of velocity. This analog voltage is then fed into an A/D converter (another AD573) resulting in a 10 bit binary representation of velocity.

GYRAC Computer and Interfacing Subsystem.

Up to this point, the gyro subsystem and the accelerometer subsystem have been discussed. The resulting output of these systems will be in digital form as mentioned earlier. The remaining task is to transmit these signals out to an external computer where they can be used for navigation purposes. This is accomplished through a bi-directional GYRAC sensor bus, the GYRAC computer, and an RS-232 interface. The sensor bus contains eight data lines,

four address lines, and one read/write line. Since the gyro and accelerometer data is all larger than 8 bits, the data from each of these devices must be gathered in two separate parts. This causes timing problems since the S/D and the A/D converters constantly update themselves with the most current measurement. This means that after a data signal is obtained, the value in the converter will change before the second half of the data signal can be transmitted.

This problem was solved by latching the data into a set of tri-state latches. These latches will hold the data as long as necessary, allowing sufficient time to transmit both bytes of the data signal. See Appendix B for more design detail. Also, since each data signal from each converter is divided into two parts, a separate address is used for each part. Thus, six addresses are needed to obtain all the gyro and accelerometer data. A seventh address is used to reset the integrator constant to zero (by discharging the capacitor over the op-amp). This reset is required to insure no initial condition exists on the integrator and can be used to reset the integrator periodically when the GYRAC system is not moving. See Appendix B for detailed design layout of address decoding.

All computer interface devices, the integrator circuit, the A/D converters, the accelerometer load resistor circuit, and a set of 7-segment LED displays are located on a general purpose wire-wrap card. The LED displays are used to read

the heading information coming from the S/D converter and display it in hexadecimal format. This information is used for initial alignment of the flux detector and troubleshooting. A layout of the wire-wrap card and the LED circuitry can be seen in Appendix B.

The GYRAC computer was originally a custom printer interface card built for AFIT, but was modified to its present state. The processor is a 6802 based microprocessor with 1k of ROM and 128 bytes of RAM. A modification was made to add an additional 2K bytes of static RAM to increase the memory capability. The new memory map is detailed in Appendix D. The computer also contains an asynchronous communication interface adapter (ACIA) which converts eight bit parallel data to RS-232 format serial data and handles all handshaking to an external computer. A parallel interface adapter (PIA) is also resident on the computer card which acts as the interface between the GYRAC sensor bus and the GYRAC computer bus. A power modification was also made to the computer to create its necessary + and - 12 volts and -5 volts from the + and - 15 volts available from the gyro base. The GYRAC computer software is present in the 1k EPROM (see Appendix F for program listing) allowing the computer to receive and respond to commands from an external computer via an RS232 serial link. A schematic diagram of the GYRAC computer showing the RAM and power modifications can be seen in Appendix D. Appendix E contains the edge

connector wiring diagrams for all the GYRAC system circuit boards (S/D converter card, interface card, and computer card) showing all interconnecting plugs.

GYRAC System - Theory of Operation

The purpose of this section is to provide a clear picture of how the GYRAC system functions as a whole. It is intended to supplement the previous subsystem descriptions. This discussion begins by explaining what occurs on system power-up and ends with a description of how the system responds to a command input. The slaving switch is assumed to be placed into the "slaved" position (representing a full up configuration of the GYRAC). The GYRAC must be stationary upon power-up to allow for stabilization of the flux detector.

Once power has been applied to the system, the rotor (spinning mass) in the gyro begins to rotate. Output from the gyro (the two square waves) is paused while the rotor comes up to operational speed (16,000 rpm). During this same time, a red HDG flag (compass warning flag - see Figure 2.4) is displayed on the indicator face. This red flag is a visual indication that the displayed heading is not valid. While the rotor is coming up to speed, the slaving signal from the magnetic flux detector is allowed to pass to the indicator providing the reference signal for magnetic north. The compass card in the indicator is rotated at the fast slaving rate, 360 deg/min [6], until the reading on the

indicator is in agreement with the magnetic flux detector slaving signal. Once the rotor has reached operational speed, the red HDG flag is removed and the indicated heading is valid. This usually occurs about one to two minutes after power-up. Any robot system using the GYRAC must account for this spin-up and alignment time (perhaps through a timed delay before requesting initial GYRAC data).

The absolute heading of the GYRAC system will be accurately shown on the indicator and on the LED display in 12 bit hexadecimal. Any rotational movement of the GYRAC will be sensed by the gyro which provides the signal to keep the indicator accurately positioned. In addition, the indicator will respond to deviations from the flux detector slaving signal at the slow slaving rate, 3 deg/min [6]. This slow rate is used to prevent the indicator from trying to follow an unstable reading from the flux detector. The flux detector is very sensitive to movement, so its output can only be trusted after it has stabilized. At this point (after the initial alignment), the flux detector signal serves primarily to compensate for gyro errors, such as drift. This particular gyro has proven to be a very accurate and stable reference. The drift rate of this gyro is less than 0.25 deg in 12 hours [8]. For this reason, the slaving signal from the flux detector could be turned off (switch to "free gyro" mode) after initial alignment is obtained.

After the heading data becomes valid, the GYRAC is ready to receive a command input. The firmware operating in the GYRAC computer is continuously checking for an input. Once an input arrives, it is compared to a list of acceptable commands. An acceptable command is a single byte of data in ASCII format representing the capitol letters A to O, see Appendix F for command definitions. If it is a valid command, the firmware program sets the appropriate address on the bus to enable the requested data (be it heading, heading rate, or velocity). The desired data is collected over the data bus (one byte at a time), converted to serial format and transmitted out via the RS-232 interface. The RS-232 interface is a simple three wire interface consisting of transmit data, receive data, and ground. See Appendix E for more detail.

It is important to note that the digital heading output is in a right-handed reference system. That is, the heading angle increases with counter-clockwise robot rotation. This is backwards from the visual indicator unit. The indicator displays increasing heading angle for clockwise rotation. Therefore, the digital output from the GYRAC and the LED displayed heading will not agree with the visual indicator except at 0 and 180 degrees. The GYRAC digital heading output was intentionally made to conform to the more conventional right-handed reference system.

III. Integration of the GYRAC System onto MARRS-1

Structural

The entire GYRAC system is contained in a new third body tier which has been added to the top of the existing MARRS-1 physical structure. It is separated and supported from the lower body tiers by eight 10.0 inch by 3/8 inch diameter threaded and tapped aluminum rods. The all aluminum third tier is 12 sided and 20.5 inches by 20.5 inches by 7.0 inches high and contains two swing down removable-pin hinged doors to allow easy access to internal components. An 18.0 inch U-shaped aluminum tower extends above the third tier to provide support and ferro-magnetic isolation for the gyro's magnetic flux detector.

In addition, four aluminum plates were constructed and attached to the first and second body tiers locking them together into a single rigid body. This was done because the original robot design allows for separate body rotation of the first and second tiers. The GYRAC requires a fixed orientation relative to the entire body and can not tolerate rotation without introducing navigation errors.

Electrical

The electrical and mechanical subsystems of the GYRAC are completely isolated and independent of the remainder of MARRS-1. Power for the GYRAC is supplied over an external cable and connects to the body tier via a four pin DIN plug

(see Appendix E for detailed power distribution). All gyro commands and data are passed to and from the GYRAC via a standard three wire RS-232 serial interface. Connection is made on the GYRAC body via a standard RS-232 DB25 cable connector (see Appendix E for pin out details). These are the only two external connections required to operate the GYRAC. It is important to note that both of these connections and system operation is independent of MARRS-1. Therefore the GYRAC could easily be removed from the MARRS-1 structure and mounted on a different platform.

Utilization of the GYRAC system by MARRS-1 for navigation requires communication between three different onboard computers and a single external disk based computer for program transmission and data collection. Figure 3.1 illustrates the required interconnections.

The navigation computer, a Motorola 6802 based system resident in the first body tier, is the navigation system control computer. Its purpose is to collect sensor data from the GYRAC and drive computers, transmit collected data to the external computer, analyze this data and decide how to move, and then issue the appropriate commands to the drive computer.

The GYRAC computer, a Motorola 6802 based system contained in the third body tier, accepts requests for data, formats the data if necessary, and then transmits the requested data.

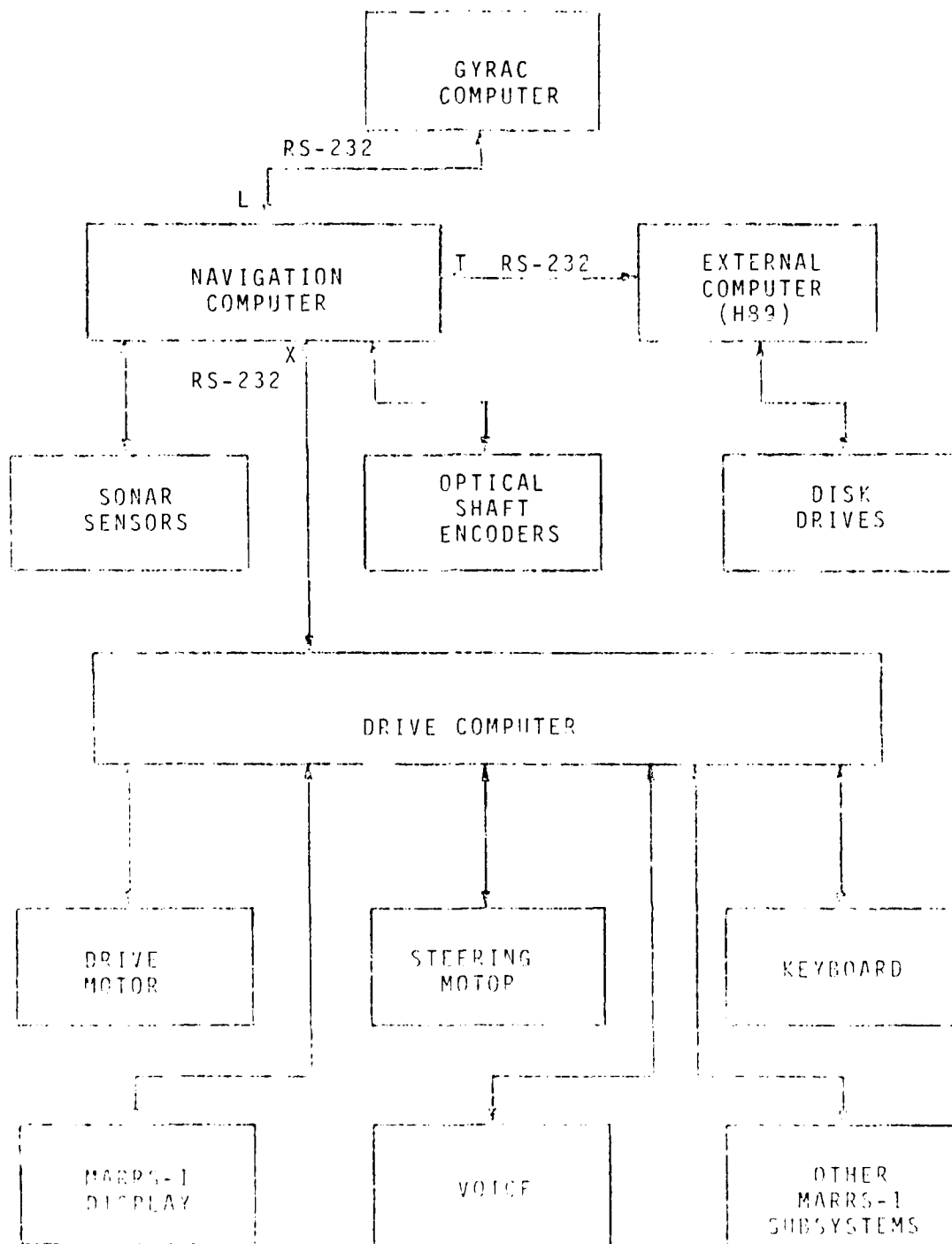


Figure 3.1 MAPRS-1 Computer System

The drive computer, a Motorola 6801 based system by Virtual Devices located in the first body tier, is the main robot computer. It controls all robot sensors and devices except the sonars and optical encoders, which are controlled by the navigation computer and the gyro and accelerometer which are controlled by the GYRAC computer. This computer is able to respond to both requests for data and commands to activate a device. However, as used in this thesis, the drive computer only accepts commands to move the steering wheel and start and stop the main robot drive motor.

The external computer, a Z80 based CP/M system by Heathkit, would not be required in a field deployed operational robot. However, as used in this thesis for data collection, it must be connected in order for the navigation software to function correctly.

All communication between the four computers is done via standard three wire RS-232 serial data links at 9600 baud. A cable is connected between the navigation computer Port X and the drive computer MENOS port. A second cable is connected between the navigation computer Port L and the GYRAC computer. The last cable is connected between the navigation computer Port T and the external computer. All cable connections are made with standard RS-232 DB25 connectors. They are located on the robot's rear lower panel, except the GYRAC connector which is on the back of the third body tier. Notice that all inter-computer

communication must go through the navigation computer.

Port L of the navigation computer was not originally designed to support 9600 baud. Therefore, a modification was made to the navigation computer board to allow Port L to select from one of eight switch selectable baud rates. It is now identical to the layout of Ports X and T [10]. All ports are currently set to 9600 baud.

In addition, the DB25 connector on the lower rear panel was wired in parallel to an existing internal cable to provide both laser barcode communication at 300 baud (original cable) and GYRAC communication at 9600 baud (new connector). Note that both functions can not be used simultaneously.

Software

The MARRS-1 GYRAC system consists of four different custom software programs which can be run in three different system configurations to provide both test data and MARRS-1 navigation.

The first configuration allows direct communication with the GYRAC computer to allow testing, calibration, and checkout of the GYRAC subsystem. It makes use of the GYRAC program resident in read only memory (ROM) on the GYRAC computer board. An RS-232 cable must be connected between the GYRAC and the external computer. The modem 720 program (M72) is executed on the external computer to provide outside communications capability. Commands are typed on

the external computer's terminal and the corresponding data from the GYRAC is displayed. See Appendix F for complete operating instructions, structure charts, and program listings. Note that not all data is displayed since the GYRAC data is transmitted in a raw eight bit serial format which produces occasional non-printable characters.

The second configuration allows for collection and storage of heading, velocity, and angular velocity data at precise 0.1 second intervals. In addition, time mark data and distance moved from all three wheel's optical shaft encoders is provided. All data is reformatted to printable hexadecimal format which may be displayed on the external computer's terminal, saved to disk, or printed on the printer. It makes use of the GYRAC monitor program, in the GYRAC computer, and the GTEST overlay program, in the navigation computer (see Appendix F and G for GYRAC and GTEST program details). An RS-232 cable connection is required between the GYRAC computer and the navigation computer Port L and between the external computer and the navigation computer Port X. The M72 program is executed on the external computer to provide communication with MARRS-1 to send appropriate commands and receive data. See Appendix G for complete operating instructions, structure charts, and program listings. The MBASIC programs CONVERT and POSITION (see Appendix I) may be run on the saved data to produce a data plot.

The third configuration demonstrates limited mobile autonomous robot navigation (using only heading data) and collection and storage of gyro heading data. The heading data is reformatted to printable hexadecimal format which may be displayed on the external computer's terminal, saved to disk, or printed on the printer. It makes use of the GYRAC monitor program, in the GYRAC computer, the MARRS.NAV program in the drive computer, and the NAV program, in the navigation computer. An RS-232 cable connection is required between the GYRAC computer and the navigation computer Port L, between the navigation computer Port X and the drive computer (MENOS), and between the external computer and the navigation computer Port T. The M72 program is executed on the external computer to provide communication with MARRS-1 to send appropriate commands and receive data. See Appendix H for complete operating instructions, structure charts, and program listings. Note that the NAV and MARRS.NAV software demonstrates a very simple method of navigation and inter-computer communication. They are not intended to form the basis of a field application, but to illustrate gyro functionality.

IV. General Robot Navigation Theory

With the recent growth in research in the area of mobile and autonomous robotics, it is only a matter of time before a truly autonomous mobile robot becomes a reality. This robot will possess a navigation system capable of gathering and processing sensory information to accurately determine its location. In addition, the navigation system will also maintain a world model of the robots environment, perform path planning (determine travel routes around known obstructions), and provide for dynamic obstacle avoidance (method of surmounting unknown obstacles). The task of the navigation system will be very complex and its future development is crucial to the realization of a mobile autonomous system.

Two major aspects of the robot navigation problem, world modeling and path planning, will be the topic of this chapter. Dynamic obstacle avoidance is considered beyond the scope of this thesis and will not be covered. First some governing assumptions will be discussed. Second, an overview of several popular approaches to world modeling will be presented. Third, a new world modeling technique will be introduced. Finally, this chapter will conclude with a detailed presentation of path planning based on this new world model.

ASSUMPTIONS

Since the world model is intended for use by a land based robot (MARRS-1 in particular) which can only move in two dimensions, only a two dimensional "floor plan" type world model will be considered. Robots that could extend or shrink themselves vertically would constitute a special category which is beyond the scope of this paper. For more information on three dimensional modeling and path planning see [15]. This section will also be concerned only with a robot which can be modeled in two dimensions as a circle (consistent with the use of MARRS-1). Some techniques for treating robots of other geometries can be found in [15]. Finally, it is assumed that all locations on the world map can be represented directly in an absolute reference frame.

PAST APPROACHES

World modeling can be thought of as providing a description (in essence a map) of the robots known operating environment. This information must be expressed in terms that the robot can easily understand and best utilize. Virtually all models to date represent the physical world of the robot in two dimensions using an outline picture method. Two approaches have been used to describe the robots world. One approach has been to model all the obstacles in the robots world. The other approach has been to model the free space or safe areas of travel for the robot. Basically, the choices are to model where the robot can or cannot go.

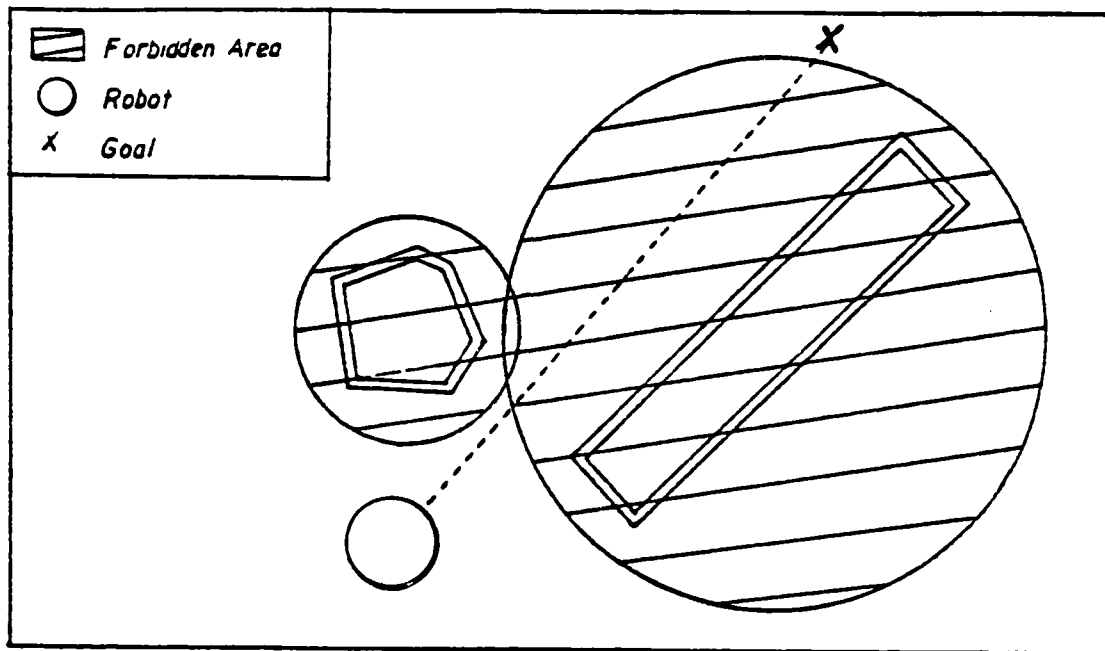


Figure 4.1. Circular approximations of physical objects [16:24]

Moravec [17] proposed modeling all physical obstacles with their enclosing circles. The radii of the enclosing circles could be increased by a small amount to provide a clear area of buffer space surrounding the obstacle. This would help prevent collisions between the robot and the obstacles. The primary drawback of this method is the waste of useful free space (see Figure 4.1).

A better way to model physical objects would be to use straight line polygonal closed surface approximations. The lowest order polygon possible would be the best choice. Lozano-Perez [15] has done considerable work in this area.

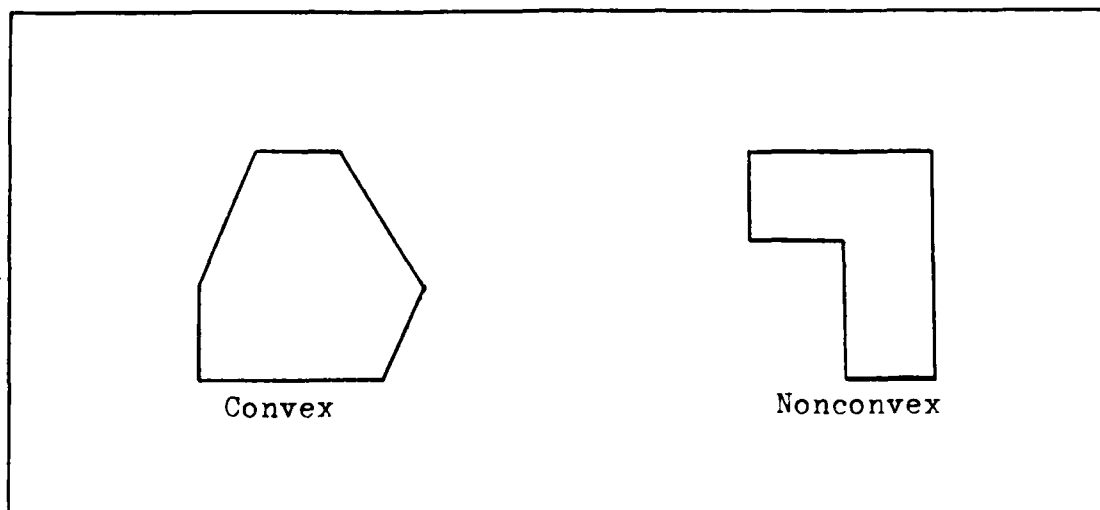


Figure 4.2. Polygon approximations to real world obstacles

He not only chooses to model physical objects as polygons but as convex polygons. A convex polygon is a polygon with no internal angle greater than 180 degrees (see Figure 4.2).

Given that all obstacles are represented as convex polygons, a path can be found around an obstacle by searching for a path around the vertices or corners of the polygon. For example, to go from point A to point B, in Figure 4.3, a path is planned going through each vertice of the polygon obstacle. Only the paths that do not cross the obstacle are considered possible. Either path 1 or 2 could be taken. Both traverse the outside perimeter of the obstacle and result in the shortest paths available. Physical objects such as that shown in Figure 4.2 which are not convex in shape are either modeled as convex anyway or modeled as overlapping convex polygons by Lozano-Perez (see Figure 4.4).

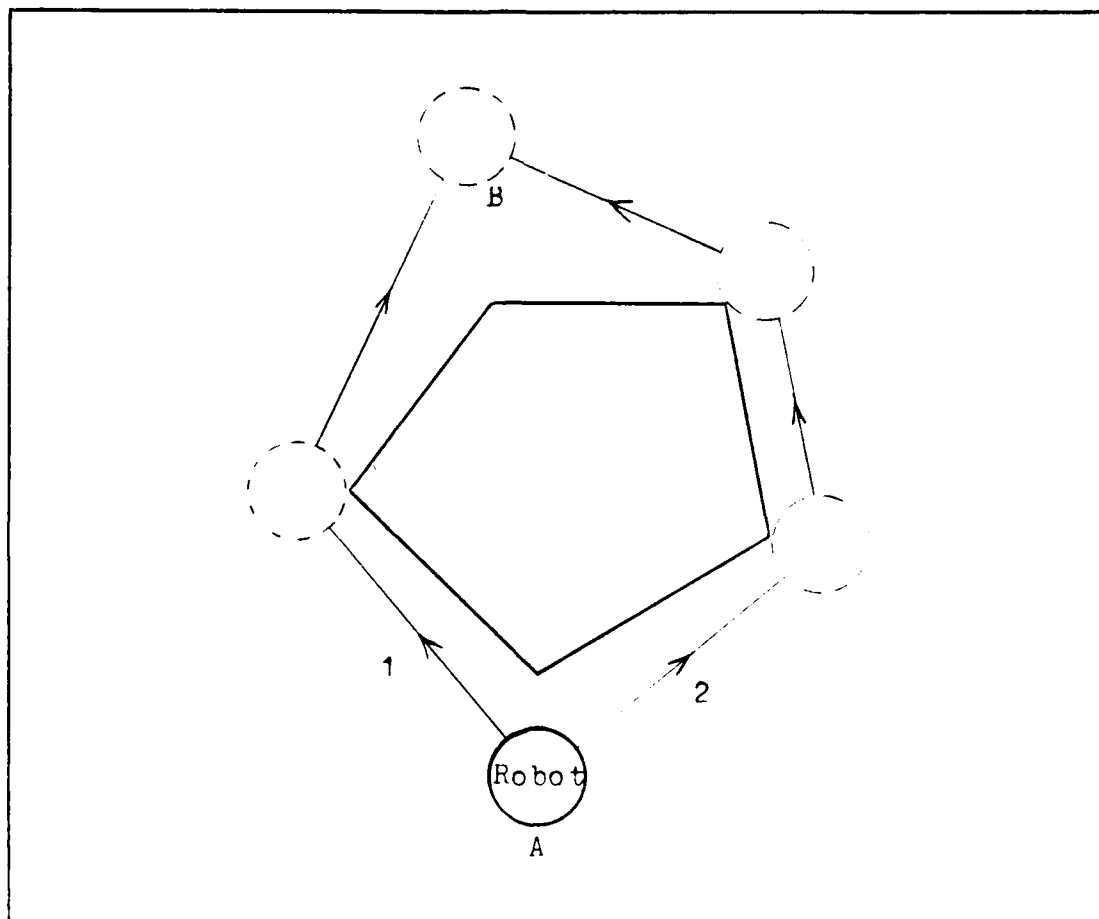


Figure 4.3. Technique of Lozano-Perez for going around an obstacle.

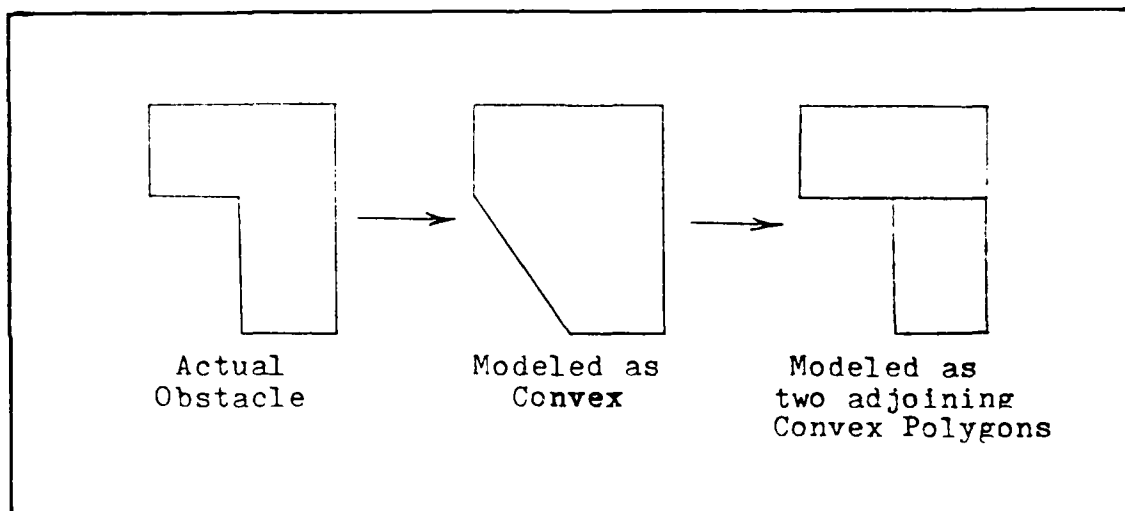


Figure 4.4. Examples of how an obstacle may be modeled using Lozano-Perez technique.

For a circular shaped robot, Lozano-Perez proposes a technique of displacing the vertices of an obstacle by the radius of the robot [15:562]. Thus, the robot can be treated as a point; thereby, greatly simplifying the path finding problem. This technique is illustrated in Figure 4.5. Notice how the robot (now a point) is made to pass through the extended vertices.

The technique of Lozano-Perez has several disadvantages. It can be wasteful of free space and computationally inefficient because physical objects must be modeled as convex polygons. In addition, this technique forces the robot to hug an obstacle as it goes around it. Relatively small errors in the world map or in the navigation data greatly increase the probability of a collision.

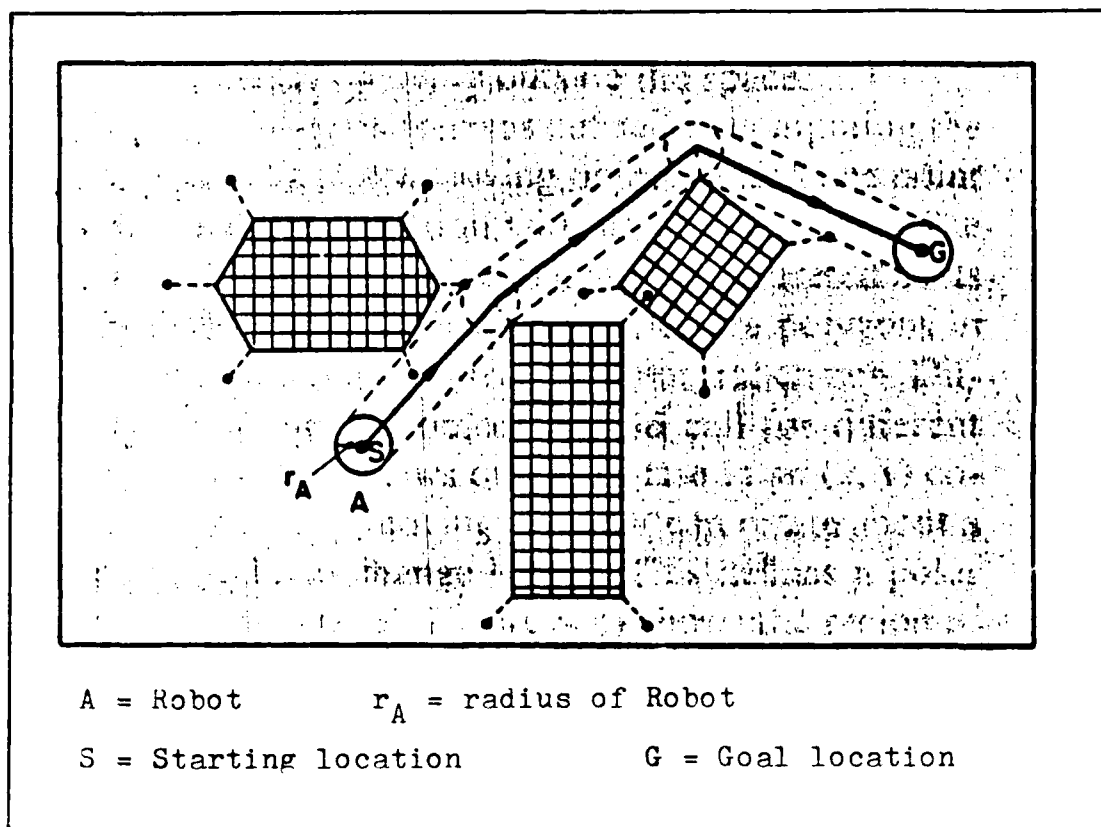


Figure 4.5. Vertices of all obstacles are extended outward so the robot can be treated as a point.
 [15:562]

Monaghan [16] also proposes using polygon approximations for obstacles, but does not restrict the polygons to convex shapes only. This results in a better representation of the actual shape of an obstacle with a minimum number of total vertices. He also shrinks the robot to a point mass and enlarges the obstacles by a likewise amount by extending the sides of the polygons (remember Lozano-Perez extended the vertices). Monaghan's path finding technique is similar to that of Lozano-Perez where a

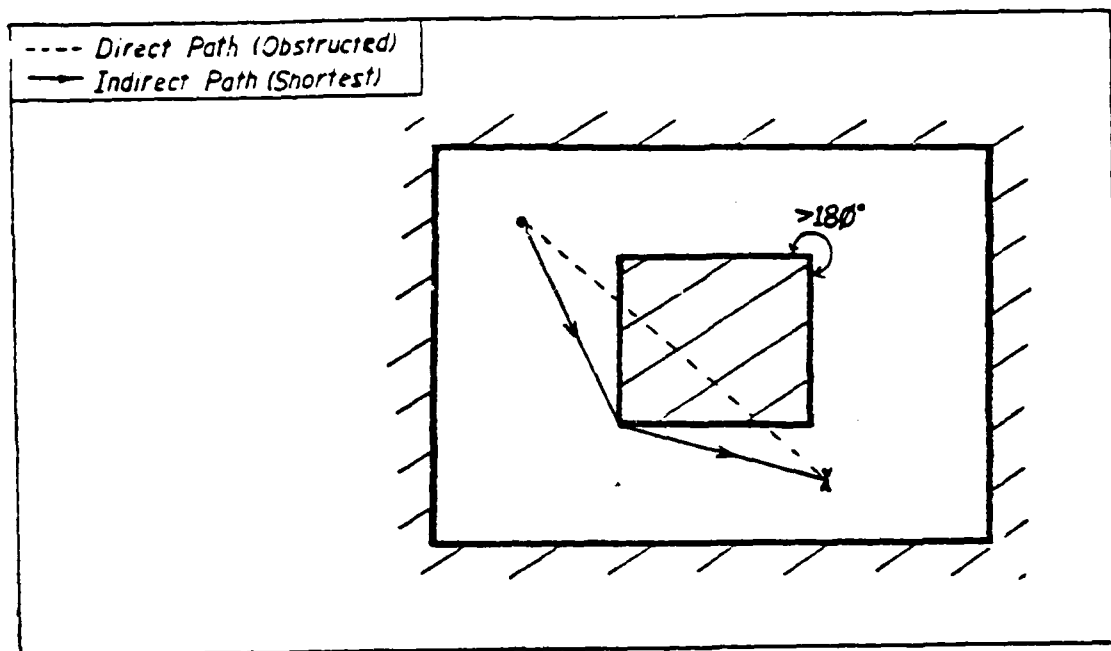


Figure 4.6a. Monaghan's modeling technique [16:47]

search of the vertices of an obstacle is performed to find a way around it. Monaghan's work emphasizes finding the shortest path to the goal point. Thus, a vertex of an obstacle is used as a "way point" as shown in Figure 4.6a. However, an inside corner (resulting from the use of nonconvex obstacles) is never considered as a way point (see Figure 4.6b). This technique does give the shortest path, but it is certainly not the safest (due to possible collision).

So far, the obstacles modeling approach to world modeling has been discussed. We have seen that obstacles may be represented by either their enclosing circle or a polygonal approximation. Another approach to world

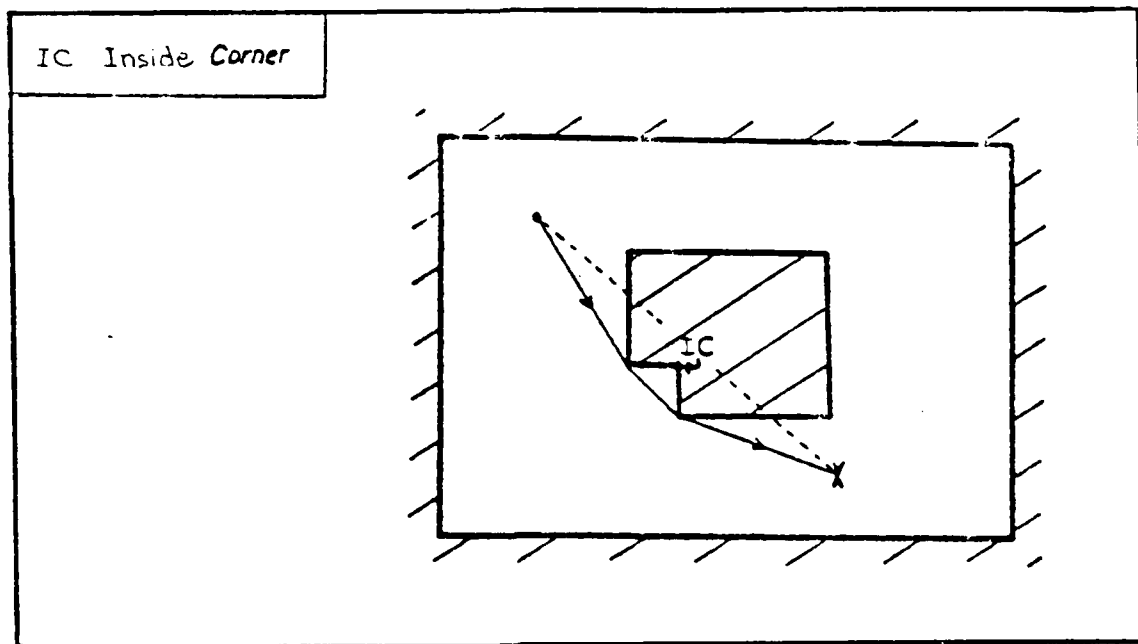


Figure 4.6b. Inside corners are not used as "way points"
[16:48]

modeling is to model the free space which a robot may occupy.

Brooks describes the free space which a robot may travel as a network of cones [16:25]. Obstacles are polygon shaped and the free space between the faces of these polygons can be formed into generalized cones or "freeways" (see Figure 4.7). The robot is restricted to travel along the center or "spine" of these cones. This technique is less prone to collision since the robot is required to remain at the centerline of free space. The major problem with this method, as pointed out by Monaghan [16:28], is "the difficulty of modeling the map to account for movement

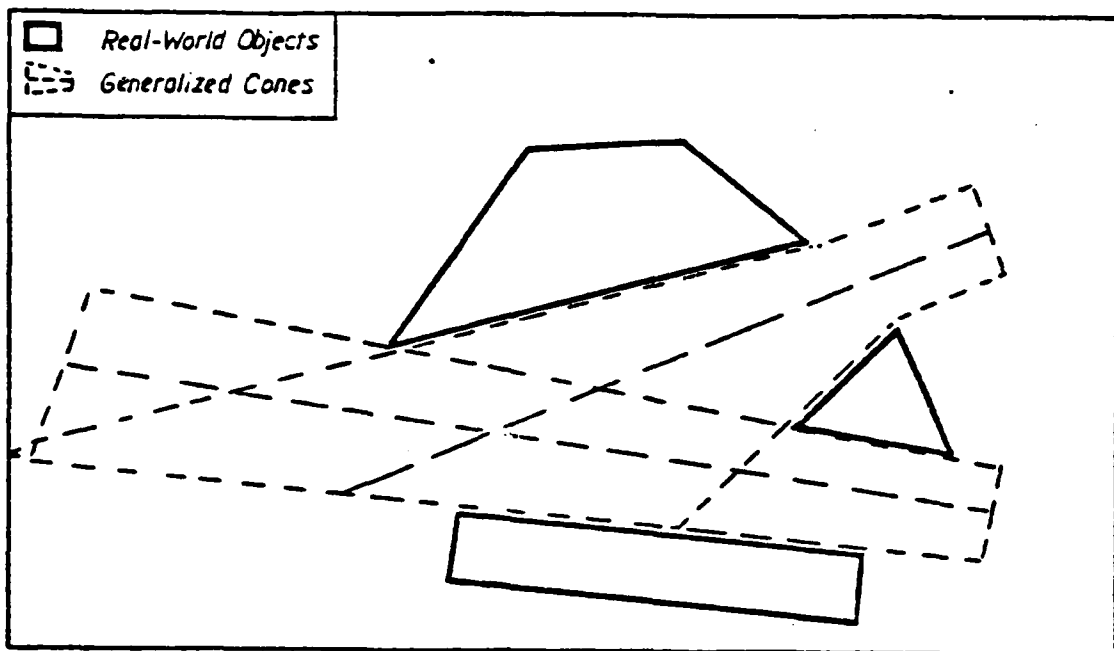


Figure 4.7. Generalized cones form freeways between obstacles [16:27]

of any obstacles. Repositioning a single object could involve comparing each of its faces with all those of every other obstacle to recompute the adjacent free space cones."

Another free space modeling technique directly models the regions through which the robot may travel. This technique is attributed to Crowley [16:28]. Crowley models the free space around objects through a series of convex polygons (see Figure 4.8). It is important to note that any two points within a convex polygon can be connected with an unobstructed line (see Figure 4.9). Thus, movement confined to within the borders of a convex polygon is guaranteed to be collision free. Motion is restricted, however, when it is necessary to travel to other regions (adjacent convex

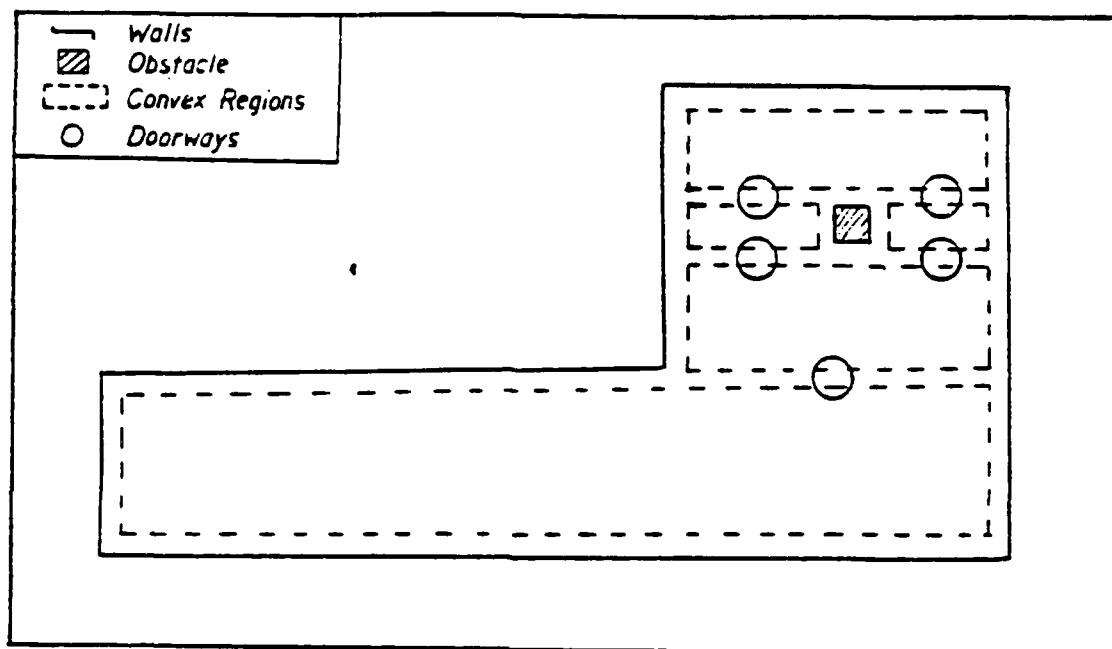


Figure 4.8. Convex regions separated by doorways represent free space [16:29] .

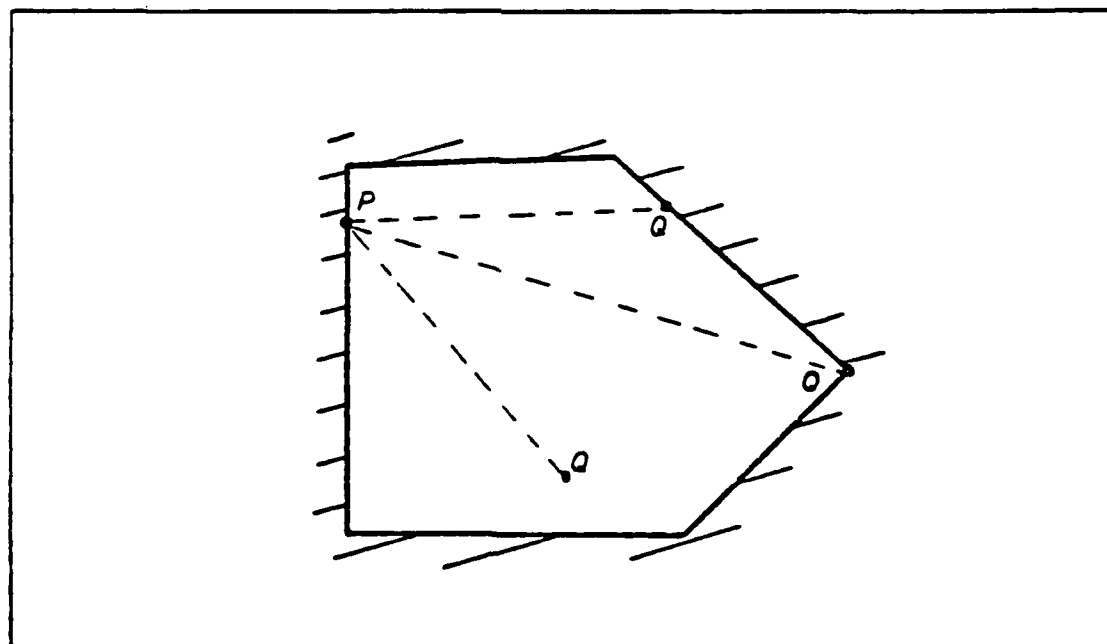


Figure 4.9. Any two points, P and Q, in or on a convex polygon may be connected by an unobstructed straight line [16:45] .

polygons). This can only be done through a defined "doorway" (see Figure 4.8). With this modeling technique, finding a path results in searching the network of doorways and free spaces.

Crowley also treats the robot as a point much the same way Lozano-Perez does. However, while Lozano-Perez enlarges obstacles to account for the robots size, Crowley shrinks his free space by an amount equal to the robot radius. One problem with Crowley's technique can be seen in Figure 4.8. Notice that for just one obstacle, five free space regions must be stored into memory. Also, if an obstacle is moved many free space regions must be recomputed. Crowley's use of doorways, however, is very appealing and will be expounded upon later.

A NEW TECHNIQUE

A brief discussion of the current schools of thought for modeling a robot's world has preceded this section. By combining some of these ideas, a better method can be obtained. Consider the following approach:

1. Obstacles will be modeled as polygons (not just convex).
2. The obstacles will be enlarged so the robot can be treated as a point.
3. Abstract safe points will be established such that at least one safe point can be reached from anywhere in the robots environment.

This method is a combination of obstacle modeling and free

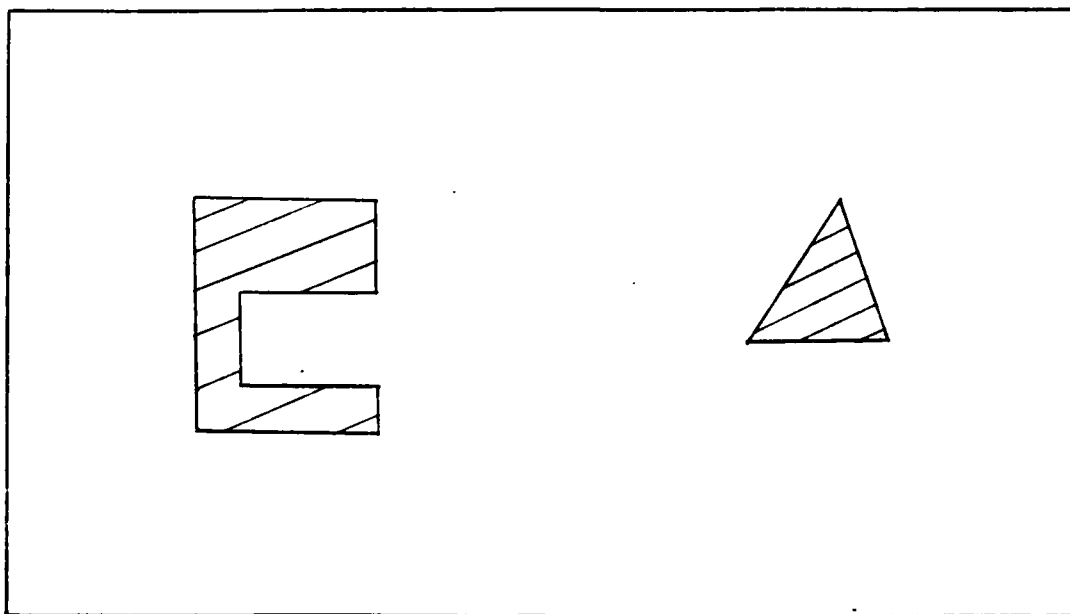


Figure 4.10 Room with two obstacles.

space modeling techniques. For example, in Figure 4.10, a room is depicted with two obstacles. Note that one obstacle is convex in shape and the other is not. Now, a series of doorways can be established much the same way as in Crowley's technique. The free space is divided into adjoining convex polygons as in Figure 4.11. Then, doorways are established between adjacent convex regions. Next, the free space boundaries are removed leaving only the obstacles and the doorways (which are represented as a series of points - see Figure 4.12). These doorway points are called "safe points". If a direct path is obstructed, a search is made of the "safe points" and indirect paths can be obtained as in shown Figure 4.13.

Unlike Crowley's technique, requiring a doorway be used

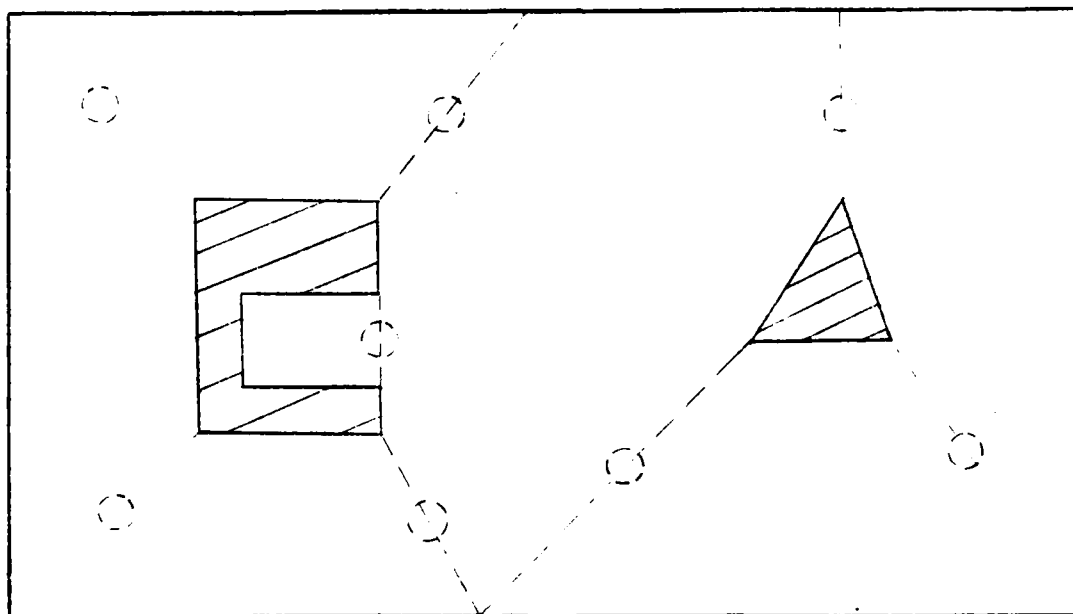


Figure 4.11. Free space is divided into convex regions to define "safe points."

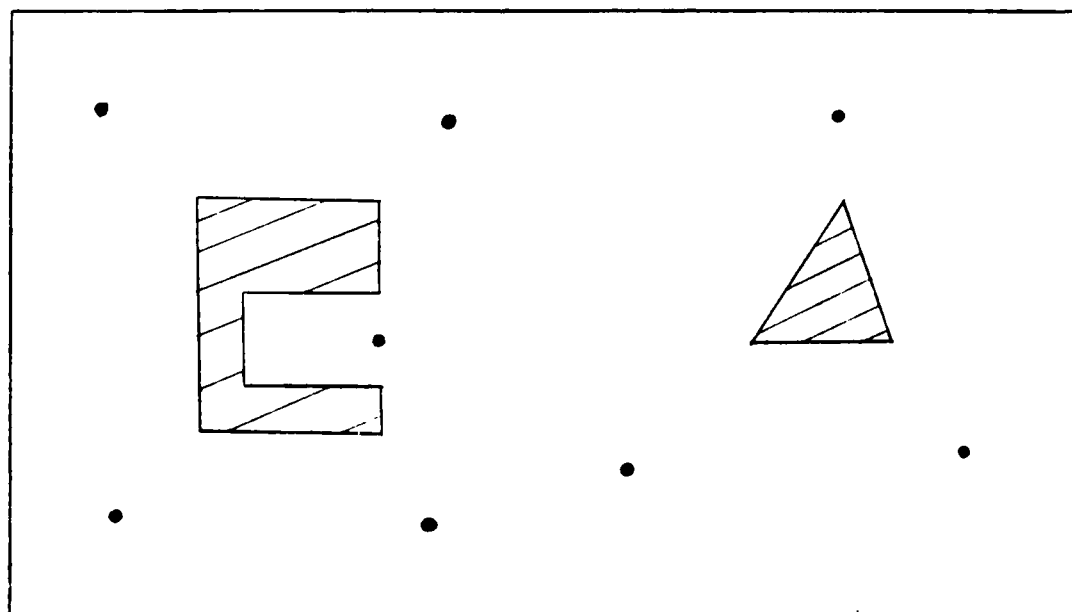


Figure 4.12. Only obstacles and safe points are modeled.

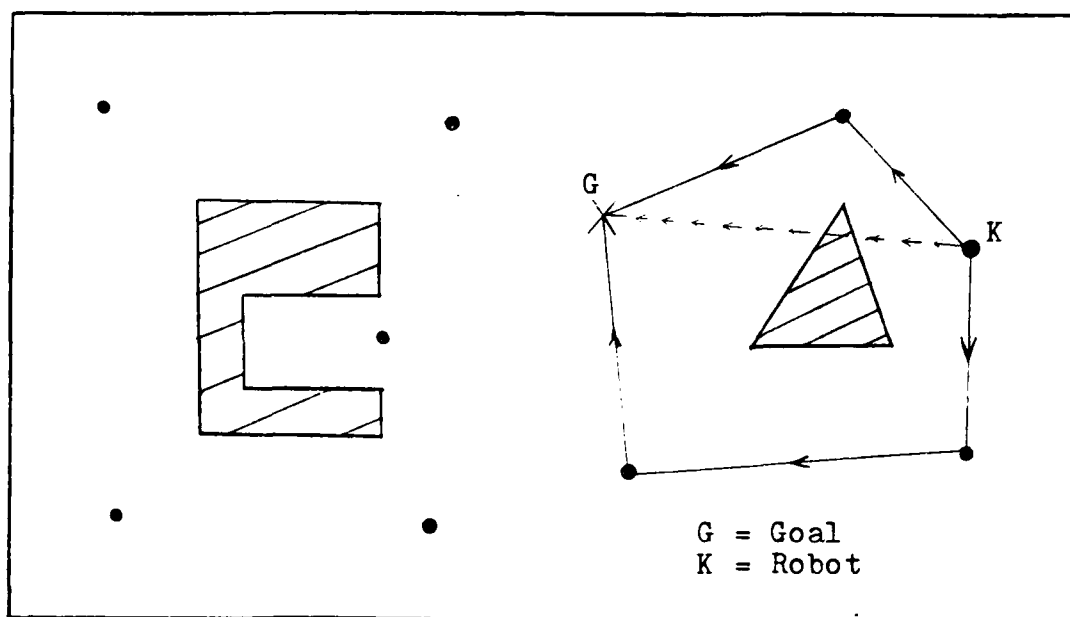


Figure 4.13. Indirect pathways pass through safe points.

as passage between free space regions, this technique uses doorways or safe points only when the goal is obstructed by an obstacle. Direct passage can take place anywhere in the room as long as the pathway is unobstructed.

To avoid having pathways which run very near the side of an obstacle, the free space boundaries must be carefully chosen when establishing safe points. For example, Figure 4.14 shows again the way Crowley separates a room into free space regions. This is a poor choice since it may require sustained travel very near an obstacle or border. Notice how the path to the goal runs parallel to the wall. This increases the chance of collision. As a rule of thumb, free space boundaries should be constructed so they never run parallel to an obstacle face or exterior boundary. Figure

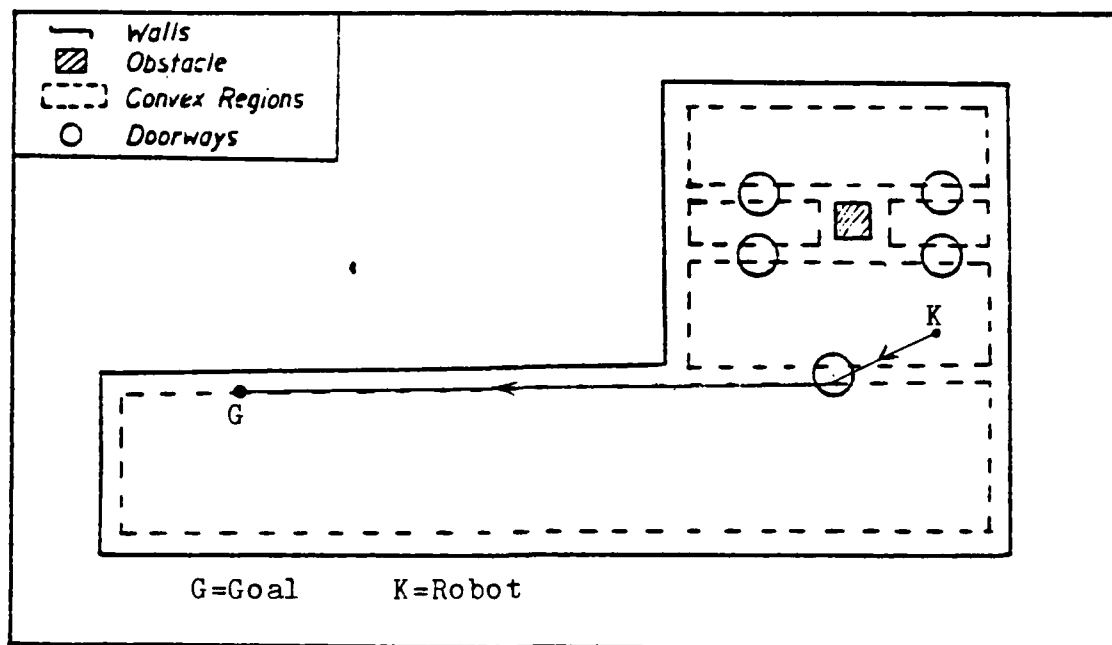


Figure 4.14. Problems occur if free space regions are not chosen correctly.

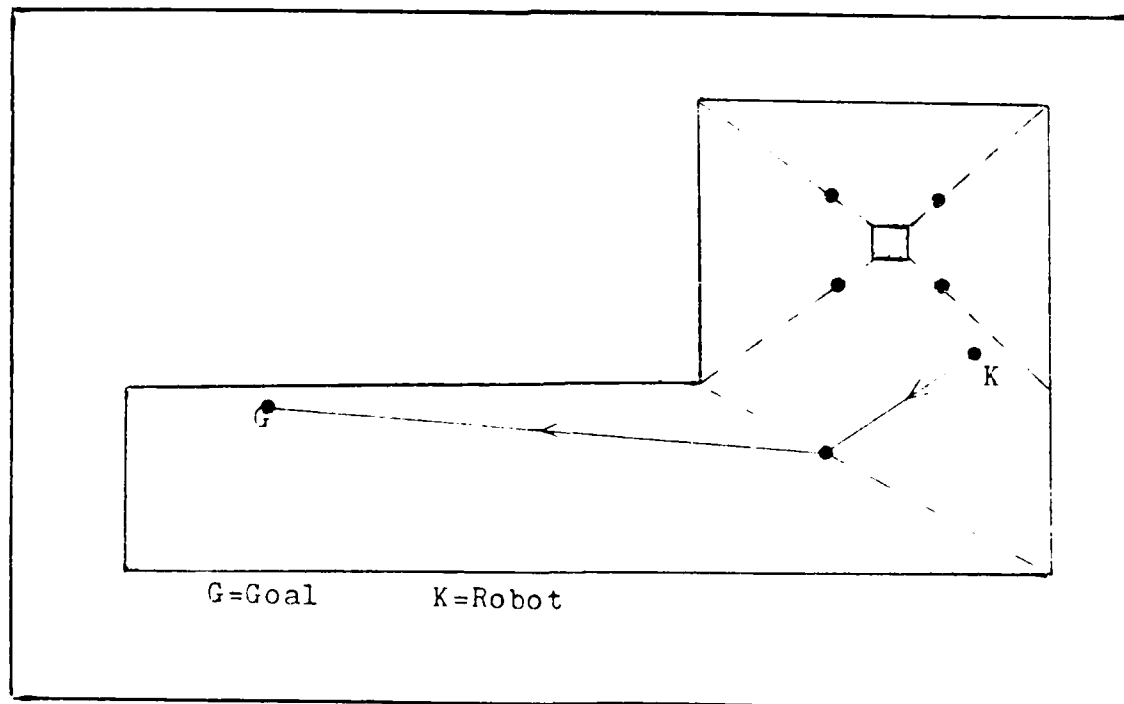


Figure 4.15. Free space regions can be chosen to minimize the probability of collision.

4.15 shows the same room only this time with different free space borders. None of the borders are parallel to an obstacle face or exterior boundary. Figure 4.15 also shows the new path for the same starting and goal points as in Figure 4.14. Notice how the path no longer hugs the wall. By using this rule of thumb, safer pathways can be planned.

This new technique offers several significant advantages. By using regular polygons to model obstacles, an accurate representation of the actual physical object can be obtained, wasting little or no free space. Treating the robot as a point precludes having to consider the volume of space occupied by the robot. Using "safe points" to plan paths around obstacles keeps the robot a safe distance from obstructions. Thus, fewer collisions should occur. Above all, this method is simple and requires minimum computer memory.

This technique also has a few disadvantages. It could be argued that not yielding the shortest path is a disadvantage. However, for a robot not under a tight energy or time constraint, the shortest route is not necessarily the best. Safety may be more important. When the world model becomes very complex, some other disadvantages appear. If an obstacle is moved, several safe points may have to be recomputed. Also, as the number of obstacles increases, the number of safe points goes up almost exponentially resulting in heavy computational loading.

DETAILED PATH PLANNING

World modeling and path planning are highly dependent upon each other. Path planning cannot take place until a world model has been determined and the best world model is one that provides for the best path planning. In the preceding discussion of world modeling, it was necessary to consider path planning in a general sense. For example, the robot must determine if an obstruction lies in its direct path to the goal. How does the robot do this? How does the robot determine the best indirect path if an obstruction exists? Details of path planning will be discussed in the following section which will answer these questions.

The world model is stored in the robots memory as an ordered list of points. All of the points (X,Y) are relative to the same reference system. Each obstacle is described by an ordered list of its vertices. Also, the vertices of all exterior boundaries are stored (to the robot, exterior boundaries are just more obstacles). Safe points are stored as a separate list of points. Thus, a simple room can be represented as in Figure 4.16.

Assume that the robot is located at $(17,10)$ and the goal is located at $(2,11)$ as depicted in Figure 4.16. Notice, that the direct path to the goal is obstructed. This is obvious to us, but how does the robot know this? Before answering this question lets review some geometry. Figure 4.17 shows a line segment connecting the points K and

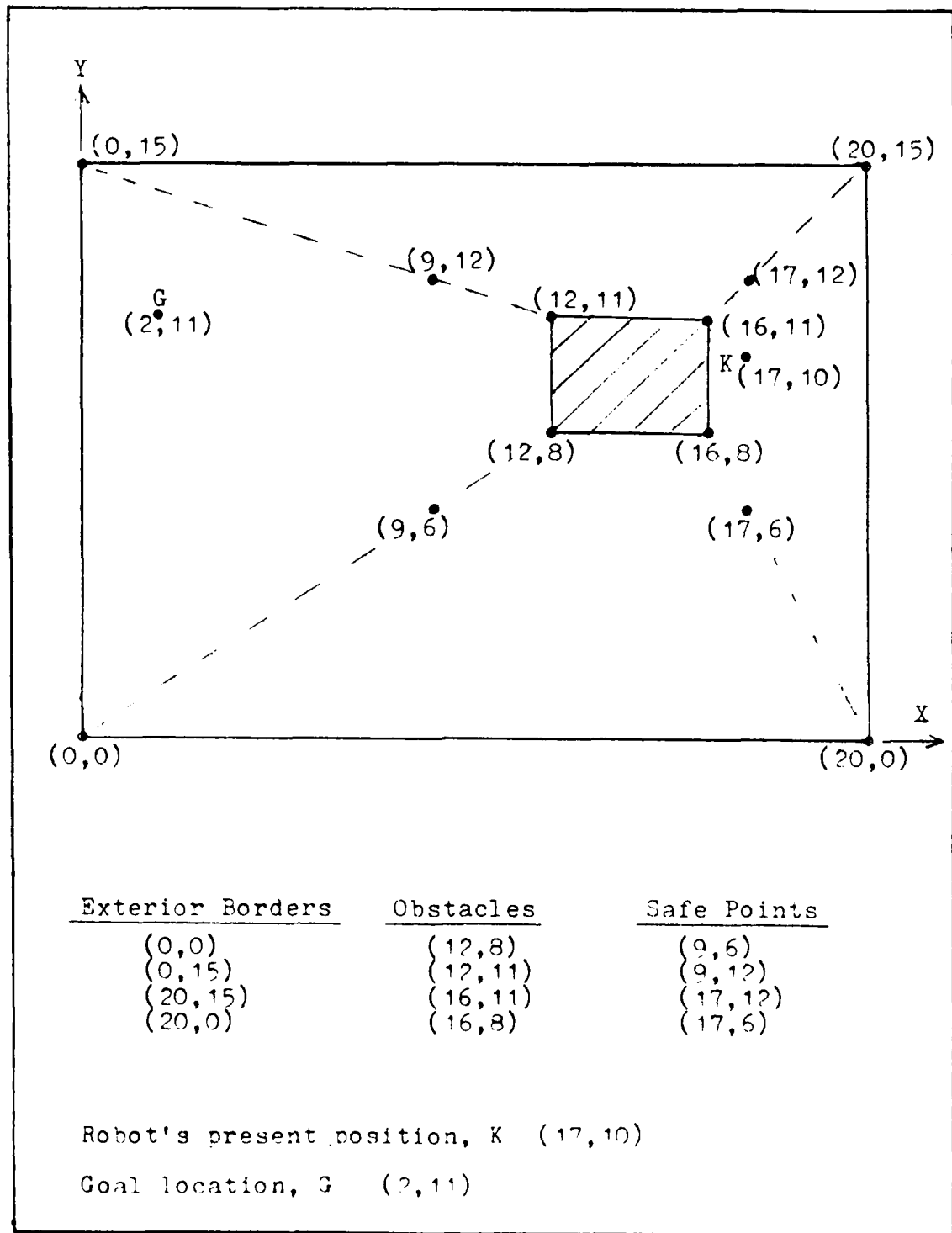


Figure 4.16. Modeling of a room with one obstacle as a set of points.

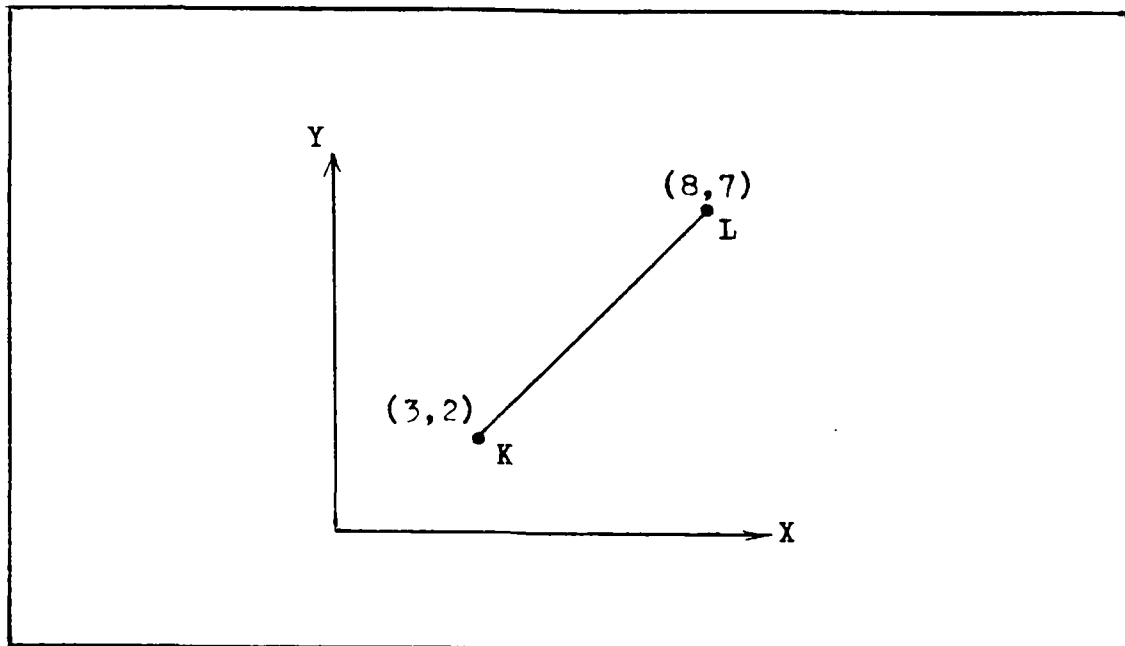


Figure 4.17. Line connecting two points can be represented through parametric equations.

L. This line segment can be represented by the following parametric equations [16:62]:

$$X = X_K + (X_L - X_K)s$$

$$Y = Y_K + (Y_L - Y_K)s \quad (1)$$

Substituting the coordinates of K and L into the parametric equations results in the following expressions:

$$X = 3 + (8 - 3)s$$

$$Y = 2 + (7 - 2)s$$

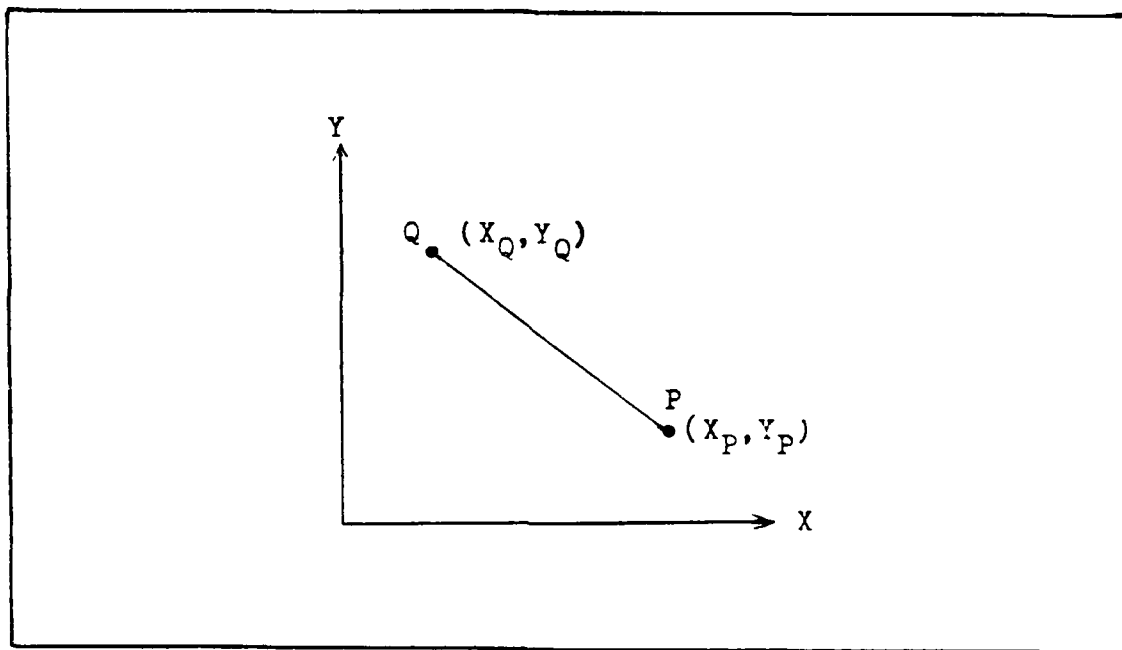


Figure 4.18. Line connecting two arbitrary points.

Simplifying

$$X = 3 + 5s$$

$$Y = 2 + 5s$$

(X,Y) obtained from these equations will always lie on the line segment for s between 0 and 1.

Now consider another set of points P and Q as shown in Figure 4.18. Let this line be represented by the following parametric relations

$$X = X_P + (X_Q - X_P)t$$

$$Y = Y_P + (Y_Q - Y_P)t \quad (2)$$

where t is the parameter in this case. Again, if t lies between 0 and 1 then (X,Y) is on the line joining P and Q . The parametric relations (1) and (2) can be used to develop a test which can determine if two line segments intersect [3]. Solving the set of equations (1) and (2) simultaneously for the parameters s and t results in the following expressions:

$$s = \frac{(X_Q - X_P)(Y_P - Y_K) - (Y_Q - Y_P)(X_P - X_K)}{(X_Q - X_P)(Y_L - Y_K) - (Y_Q - Y_P)(Y_L - X_K)} \quad (3)$$

$$t = \frac{(X_L - X_K)(Y_P - Y_K) - (Y_L - Y_K)(X_P - X_K)}{(X_Q - X_P)(Y_L - Y_K) - (Y_Q - Y_P)(X_L - X_K)}$$

The parameter values, s and t , obtained from the above expressions can be used to determine if two lines intersect. Two lines intersect only if both s and t take on values between 0 and 1. This test will hereafter be referred to as the parameter test.

To determine if an obstacle lies in the direct path of the robot, the parameter test is performed. The robot's location and the goal point form one set of points (K and L). The vertices (P and Q) of each obstacle are then used, one pair at a time, to determine if an intersection exists. All obstacles or obstacle faces may not need to be checked

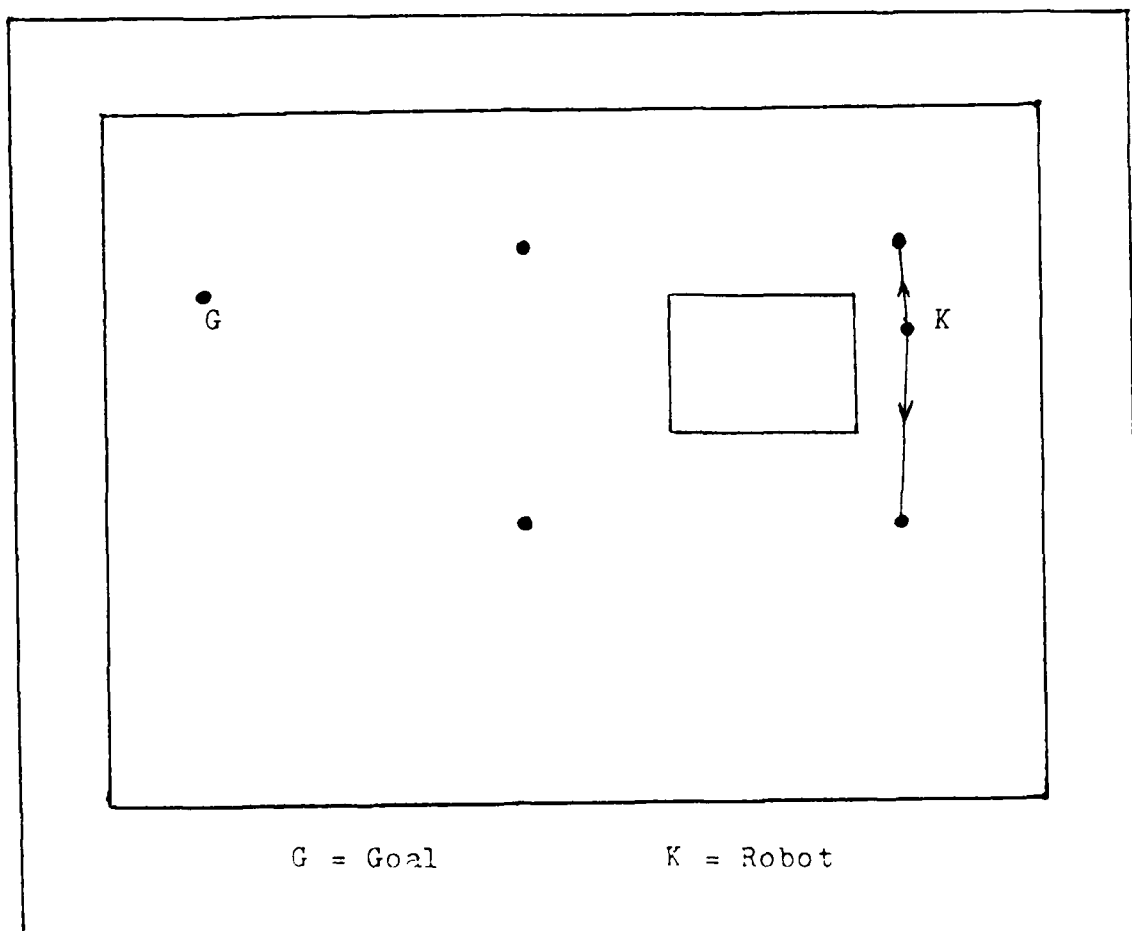


Figure 4.19 Two safe points can be reached through a direct path from the robots present position.

for intersections. No more tests are needed once the first intersection is found. Of course, if no intersections exist then the robot has a clear direct path. If an intersection is found, the robot must determine an indirect path.

To determine an indirect path to the goal point, the robot must perform a search through all the safe points and determine which ones he has direct access to. The parameter test is again used to eliminate the safe points with direct path obstructions. For our example (Figure 4.19), two safe

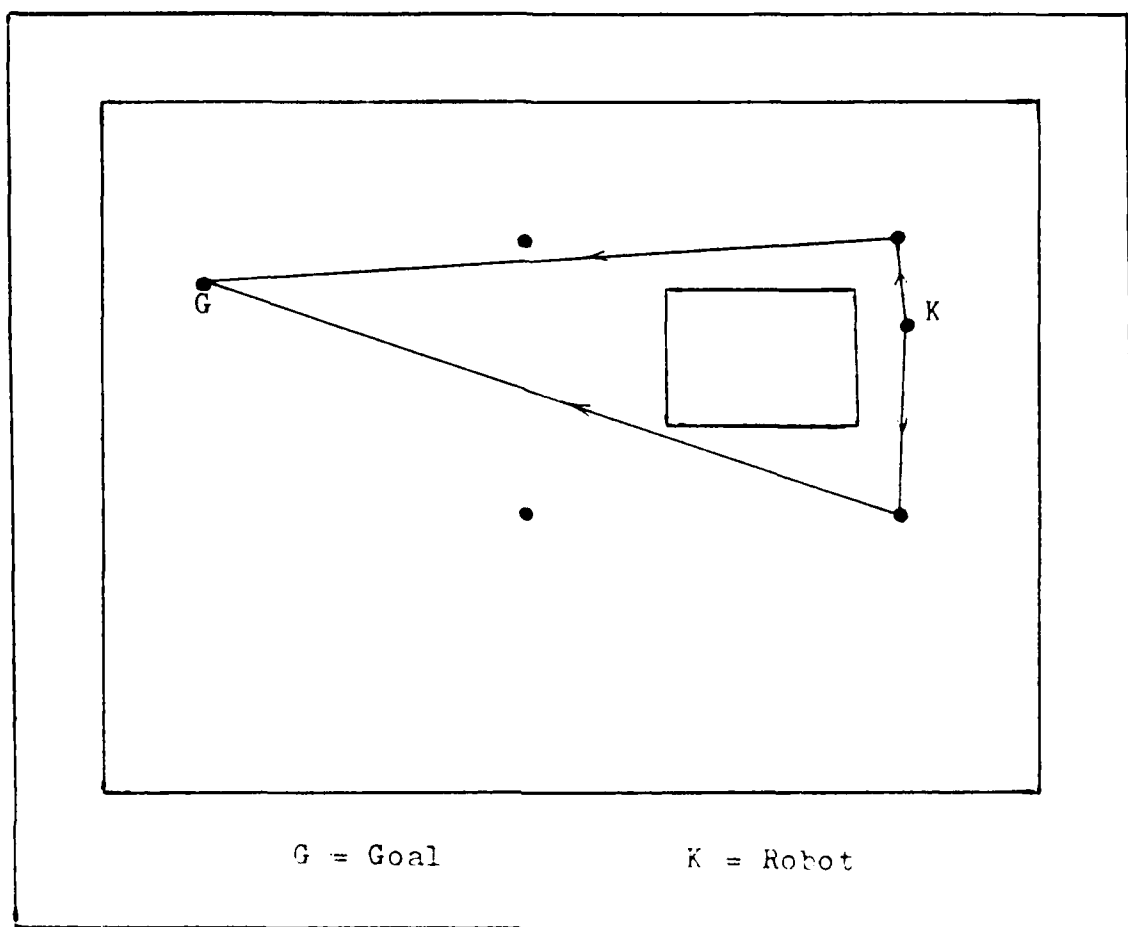


Figure 4.20. Two safe point paths lead to the goal.

points can be reached by a direct path from the robots present location. From each of these reachable safe points, a direct path to the goal is checked for obstacles (again using the parameter test). If no direct paths exist, another safe point must be found. For our example, this is not necessary since the goal can be reached directly from either safe point (see Figure 4.20). However, which path should be taken?

To select the "best" path among several possibilities,

an optimization test is performed. For each possible path, a cost function is maintained. The optimum path is the one with the lowest cost value. The cost function expression is as follows:

$$\begin{aligned} \text{COST} = & \left(\begin{matrix} X \\ K \end{matrix} - \begin{matrix} X \\ P_1 \end{matrix} \right)^2 + \left(\begin{matrix} Y \\ K \end{matrix} - \begin{matrix} Y \\ P_1 \end{matrix} \right)^2 + \left(\begin{matrix} X \\ P_1 \end{matrix} - \begin{matrix} X \\ P_2 \end{matrix} \right)^2 \\ & + \left(\begin{matrix} Y \\ P_1 \end{matrix} - \begin{matrix} Y \\ P_2 \end{matrix} \right)^2 + \dots + \left(\begin{matrix} X \\ P_{n-1} \end{matrix} - \begin{matrix} X \\ P_n \end{matrix} \right)^2 \\ & + \left(\begin{matrix} Y \\ P_{n-1} \end{matrix} - \begin{matrix} Y \\ P_n \end{matrix} \right)^2 + \left(\begin{matrix} X \\ P_n \end{matrix} - \begin{matrix} X \\ G \end{matrix} \right)^2 + \left(\begin{matrix} Y \\ P_n \end{matrix} - \begin{matrix} Y \\ G \end{matrix} \right)^2 \end{aligned}$$

where K = Starting Point
P = Safe Point
G = Goal
n = number of safe points used

This cost function is merely the sum of the distances squared of each leg of the path. Thus, the optimum path is the shortest path.

CONCLUSION

World modeling and path planning represent only a portion of the general robot navigation problem. However, their importance to the realization of an autonomous mobile robot system should not be taken lightly. Before a robot can begin to move, it must have some knowledge of its environment and it must be able to plan out a collision free route through its environment. This problem has received the recent attention of several researchers. Some of the current

techniques of world modeling have been presented along with a new approach. Path planning, under this new approach, consists of finding an unobstructed pathway to the goal point. Safe points are used only if a direct path does not exist. The details of this path planning have been developed through a simple example.

V. Testing, Analysis, and Results

The testing of the GYRAC system was divided into three primary phases. The goal of each of these phases is listed below:

- Phase I. Verify the functionality of the GYRAC system.
- Phase II. Determine if the data from the GYRAC can be used to accurately track the location of the robot (MARRS-1) as it moves about the test area.
- Phase III. Demonstrate the capability of MARRS-1 to use GYRAC heading data to follow a programmed heading exercising closed loop steering control.

Phase I

The primary thrust of this phase was to verify that every part of the GYRAC system operated properly. This turned out to be a tremendous task consuming a substantial portion of the allotted thesis time.

For Phase I testing, the GYRAC was connected to an H89 computer through an RS-232 interface and interrogated via M72 modem software. A logic state analyzer, and oscilloscope (see Appendix L) were used to troubleshoot the GYRAC hardware, firmware and verify correct operation of the GYRAC computer. Excluding the accelerometer subsystem, all hardware and software was eventually verified to function exactly as planned. The accelerometer subsystem could not be completely verified until subjected to motion. However, under static test several problems were encountered.

Output from the accelerometer integrator circuit was continually changing. Within a few minutes after power-up of the GYRAC system, the integrator output would become saturated. Operational amplifier (op amp) integrator circuits of this type, operating normally, eventually integrate into saturation under a constant input. However, the rate at which the output from the GYRAC integrator increased was much faster than anticipated. The input to the integrator (output from the accelerometer) was not constant, due to the extreme sensitivity of the accelerometer to movement, but it was very small (about 0.1mv). Such a small input should not cause saturation of the integrator so rapidly. An identical integrator circuit was breadboarded for testing.

The breadboarded integrator circuit was tested without an input (zero input voltage) and within a few minutes after power up it would integrate into saturation (just like the actual circuit). This was unexpected. After consulting with Analog Devices Corporation, it was discovered that the observed drift rate could be modeled mathematically through the following equation:

$$R = \frac{I \quad B}{C}$$

where R = drift rate
 I = current bias of operational amplifier
 B
 C = capacitance of integrator feedback

It can be seen from the above equation that in order to decrease the drift rate, it is necessary to decrease the current bias of the op amp or increase the capacitance in the circuit, or both. The AD544 op amp (see Appendix A), manufactured by Analog Devices, was selected as a replacement due to its low current bias of 10 picoamps. Also, the integrator circuit was redesigned to contain a higher capacitance. Both a 200 microfarad and a 2000 microfarad capacitor were ordered. After obtaining the capacitors, they were measured for actual capacitance and resistors were chosen to achieve the appropriate gain for the integrator circuit. The 2000 microfarad capacitor was selected for installation into the GYRAC due to the very low drift rate achieved in the test circuit with this capacitor. The lowest possible drift rate was desired since the input signal to the integrator circuit is also very small.

After installation into the GYRAC, the accelerometer circuit was again tested. The results were much better than originally obtained. However, the output from the integrator was erratic and inconsistent. Through a process of elimination, another problem was found. The active CMOS switch used to reset the integrator (through software) was leaking current into the integrator circuit and charging the

capacitor thereby causing the inconsistent and erratic output. The switch is presently disconnected from the circuit and a manual reset of the integrator must be performed by physically shorting across the capacitor.

At the end of Phase I testing, the accelerometer subsystem appeared to be functioning correctly. The GYRAC system was ready for Phase II.

Phase II

The objective of Phase II was to collect GYRAC heading and velocity data while the GYRAC was under motion and post process the data to determine the location of the robot (MARRS-1) in the test area. The computed location of the robot could then be compared with the actual location to test the performance of the GYRAC system.

For this phase of testing, the GYRAC was fully integrated with the MARRS-1 test bed. A memory overlay program, GTEST.A, was created to take advantage of firmware already operating inside the NAV computer. Clifford and Schneider had produced NAV computer firmware for collecting data from the various optical shaft encoders and sonars onboard the MARRS-1. [10:B-1] The overlay program replaces Clifford and Schneider routines for gathering sonar data with routines for gathering GYRAC data. See Appendix G for GTEST.A structure charts, program listing, and operating instructions.

All data are received through an H89 computer via M72

modem software and are then stored on floppy disk. The collected data are in exactly the same format as Clifford/Schneider data [10:IV-11] with the GYRAC data in place of the sonar data.

The raw GYRAC data is in hexadecimal format so an MBASIC program called CONVERT (see Appendix I for listing) was created to convert the raw data to integer format. The integer data is then used in another MBASIC program called POSITION (see Appendix I for listing) which computes the position of the robot in the test area based on the GYRAC heading and velocity data and a given initial position. See Appendix J for a sample output from the POSITION program. The computed position is in terms of a cartesian coordinate system centered in a corner of the test area as shown in Figure 5.1. The GYRAC is aligned such that heading is referenced to zero degrees along the x-axis and increases in the counter-clockwise direction, right handed system.

After several test runs with consistent but unusual results, the accelerometer subsystem was again suspect. The computed position of the robot indicated almost no movement, see Figure 5.2. The velocity levels gathered from the GYRAC were much too small. Eventually, it was discovered that the integrator circuit was loading the internal accelerometer restorer circuit (servo), see Figure 2.11. The result was a total changing of the characteristics of the accelerometer. The voltage sensitivity of the accelerometer, 2 volts/g, was

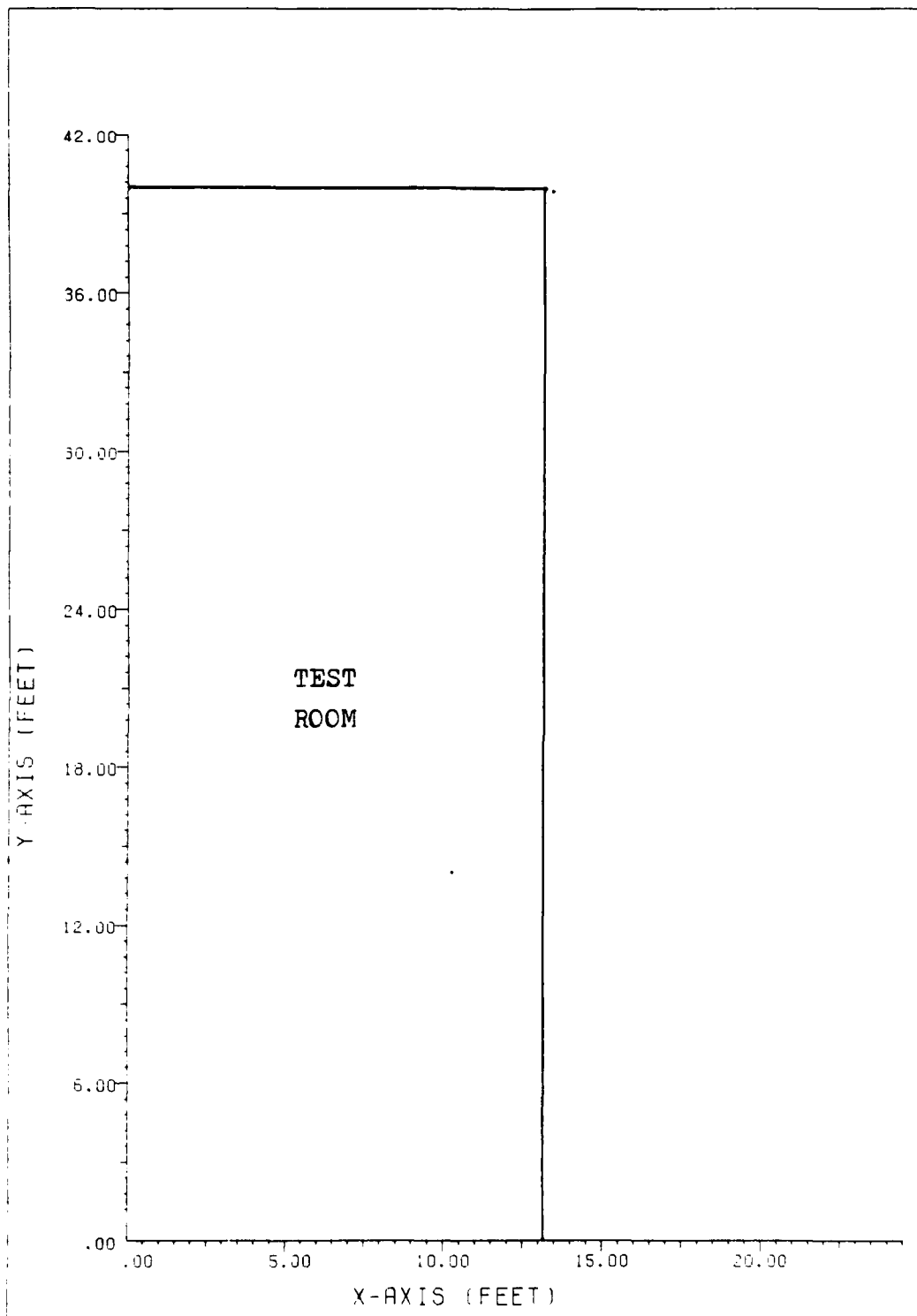


Figure 5.1. Cartesian coordinate system is centered in corner of test area.

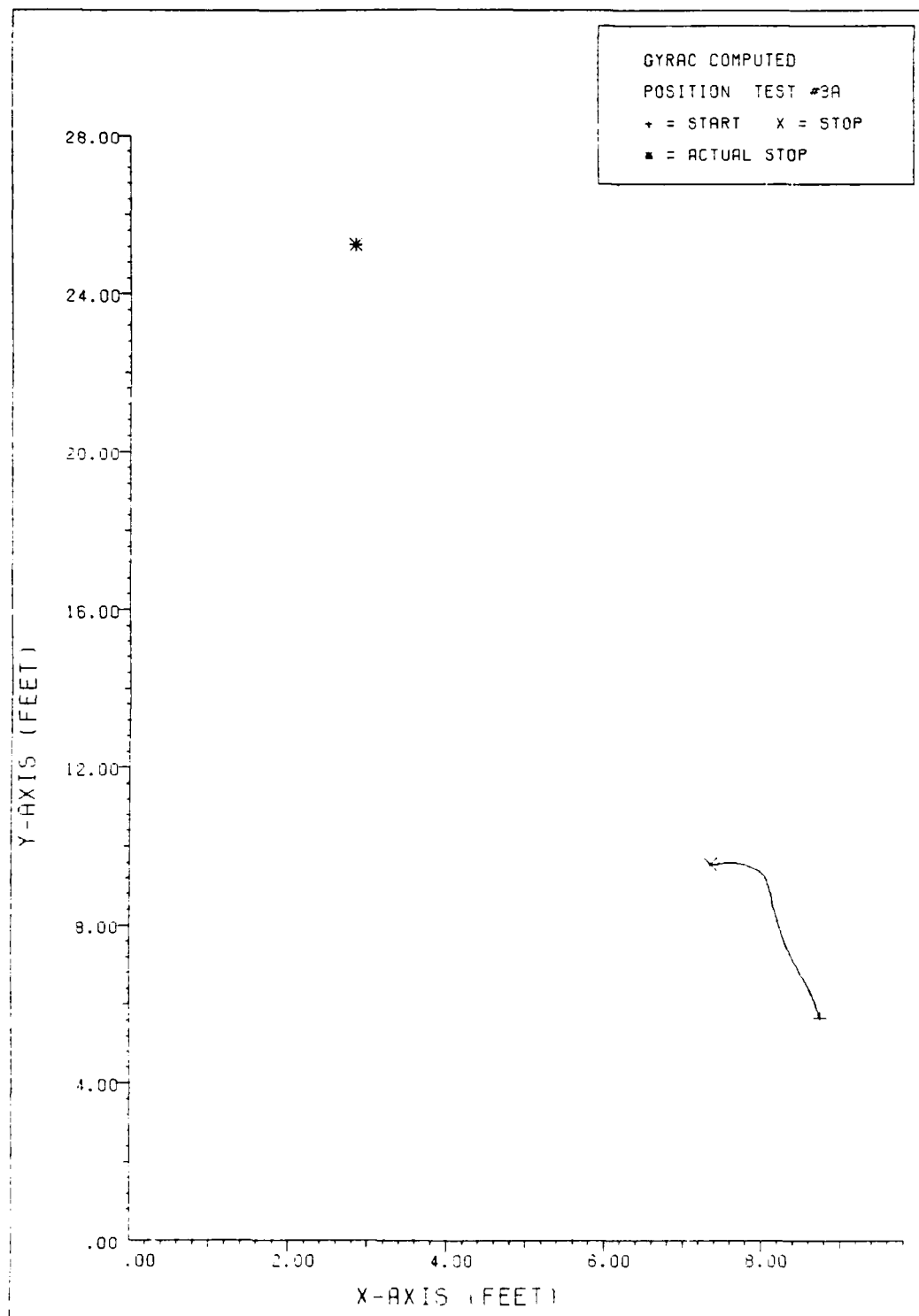


Figure 5.2. GYRAC computed position test using accelerometer sensitivity of $2v/g$.

no longer valid. A tumble test of the accelerometer was performed with the accelerometer completely connected to the the rest of the GYRAC, under full electrical load. The new voltage sensitivity was measured to be 0.393 volts/g instead of the 2 volts/g desired. This explained the low velocity levels. However, the accuracy of this newly measured sensitivity was questionable since the total loading effect of the integrator circuit could not be determined. The 0.393volts/g was measured at the load resistor R_L (see Figure 2.11). Using this new sensitivity, further testing resulted in computed positions that were in error on the high side. The computed location of the robot was always downrange from the actual location, see Figure 5.3. This indicated that the actual sensitivity of the accelerometer must be higher. An average sensitivity value of 0.6volts/g was obtained by comparing test runs using the 2v/g sensitivity with those using the 0.393v/g sensitivity. This 0.6v/g sensitivity resulted in computed positions much closer to the actual positions but still only with "ball park" accuracy, see Figure 5.4. In addition, the results were not consistent, sometimes high and sometimes low, almost random. Another problem had been around from the beginning; the output from the integrator (velocity) would not go back to zero after stopping MARRS-1. These problems indicated a possible error in sensed acceleration.

Several tests were performed with only the

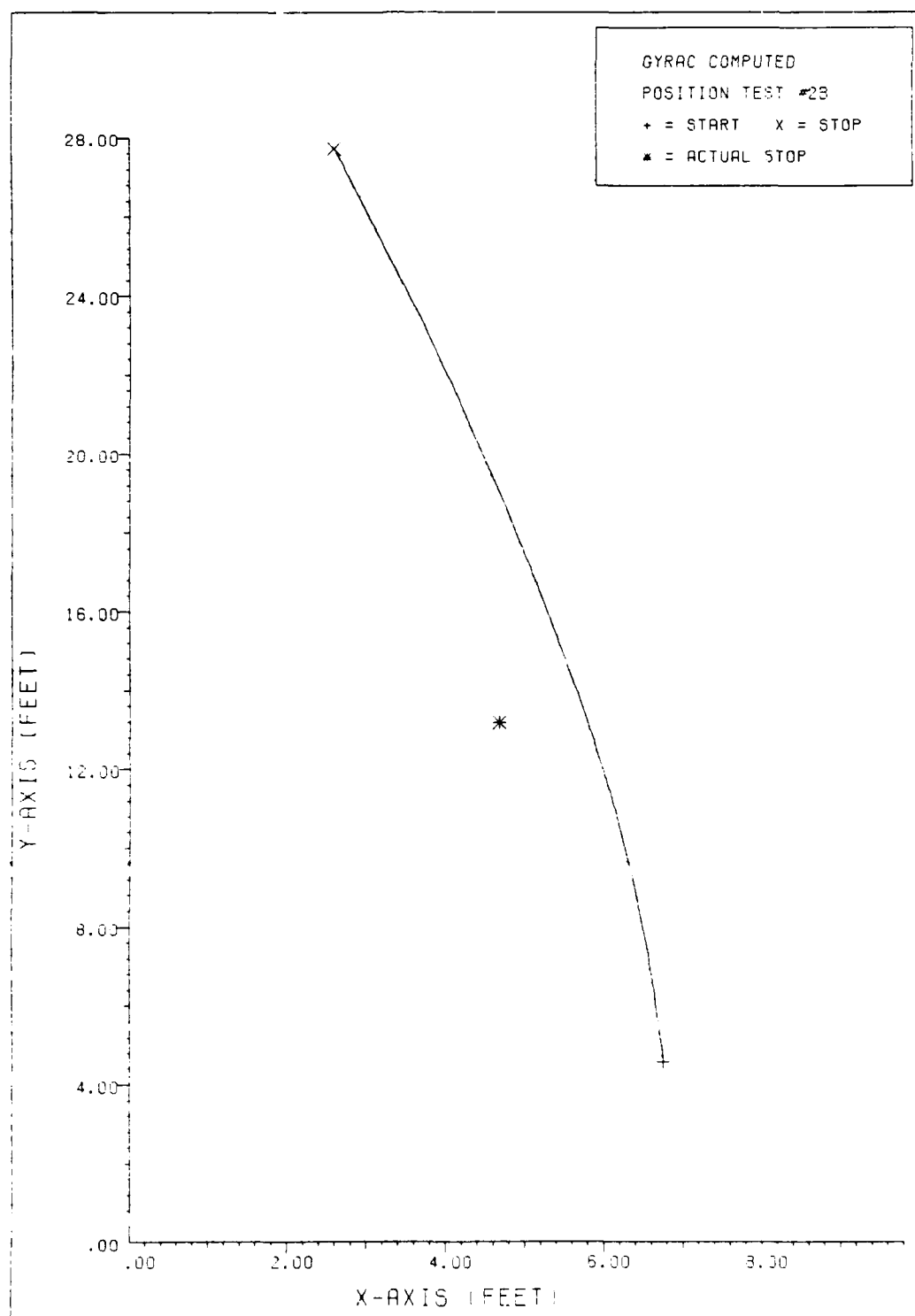


Figure 5.3. GYRAC computed position test using accelerometer sensitivity of 0.393v/g .

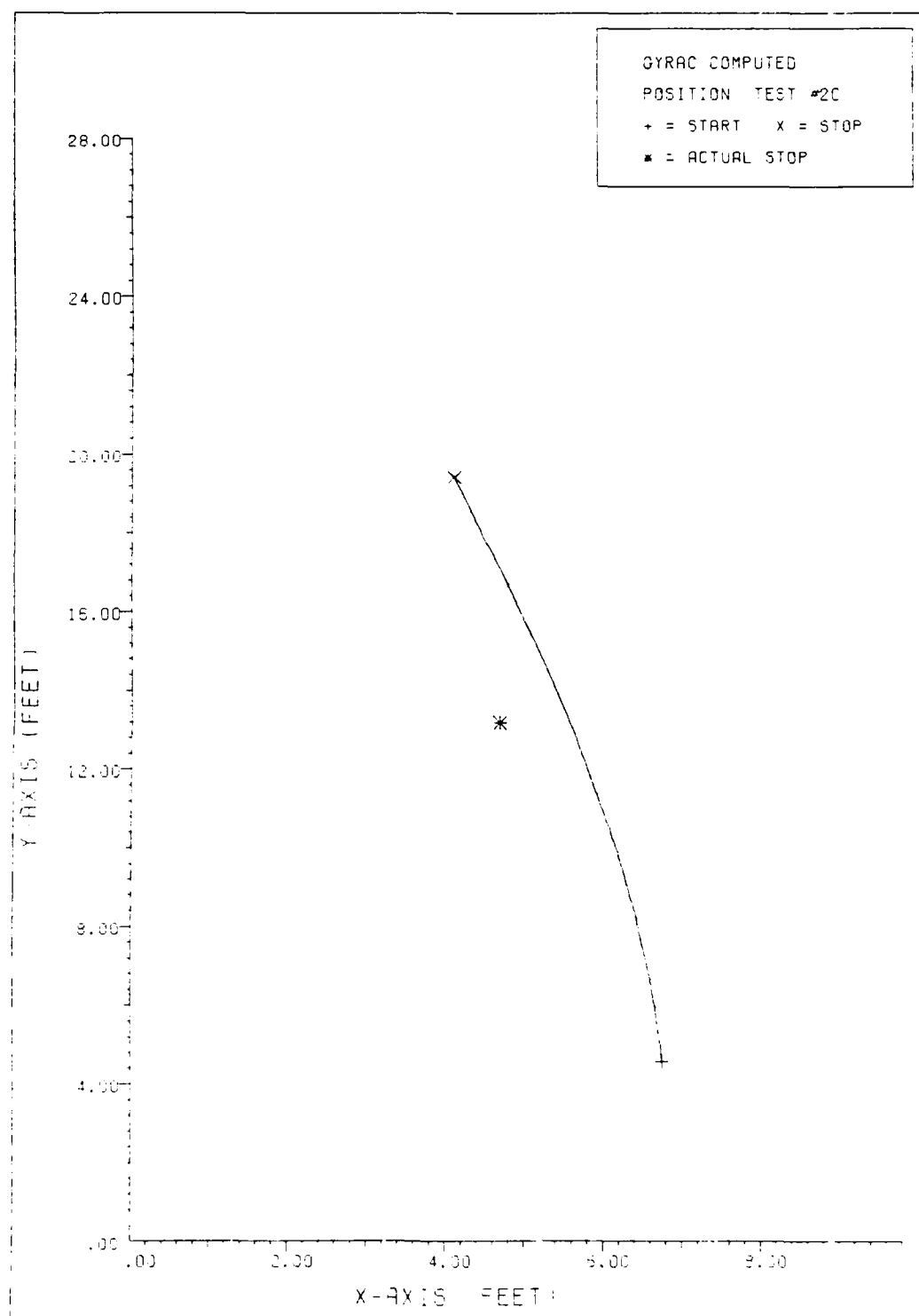


Figure 5.4. GYRAC computed position test using accelerometer sensitivity of 0.6v/g.

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MOBILE AUTONOMOUS RO. (U) AIR FORCE INST OF TECH
WRIGHT-PATTERSON AFB OH SCHOOL OF ENGI.

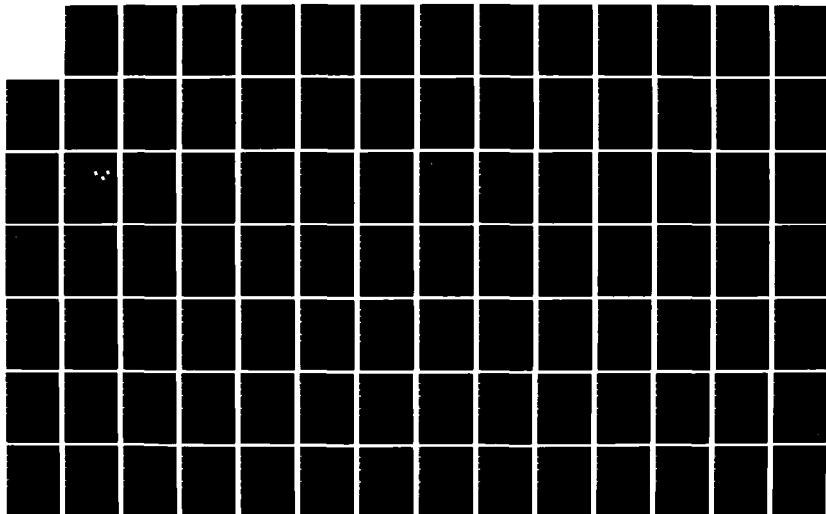
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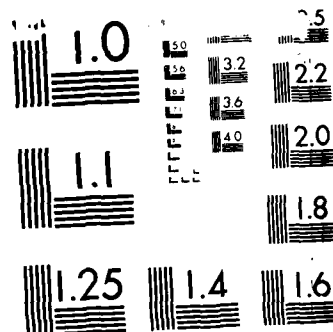
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MICROCOPY RESOLUTION TEST CHART
NATIONAL BUREAU OF STANDARDS-1963-A

accelerometer in the system. The integrator circuit was completely disconnected and the accelerometer output was connected to a gain circuit which was connected to the A/D converter. Pure acceleration data was obtained to determine the levels of acceleration achieved by the MARRS-1 under normal movement about the test area. These tests were very revealing. Figure 5.5 is a plot of acceleration vs time and illustrates the random nature of the sensed acceleration. It shows that the actual acceleration due to motion sensed by the GYRAC is on the same order of magnitude as the sensed acceleration due to tilt error (sensed gravity). In essence, the signal to noise ratio of the system is about one. The MARRS-1 moves at such a slow speed that the actual acceleration never gets much over the noise level. For example, in Figure 5.5, it is not obvious when the robot began movement and when it stopped. In Figure 5.5, the robot actually started forward movement at 2.3 seconds and was at a complete stop at 13.8 seconds. Thus, the acceleration data from the GYRAC and likewise the velocity data cannot be relied upon without some type of error compensation or a stable platform. The assumption of a perfectly smooth and level surface had been violated.

Due to the results from the acceleration tests, it was not necessary to continue Phase II testing. The accelerometer subsystem could never perform adequately without major modifications. Therefore, the third phase of

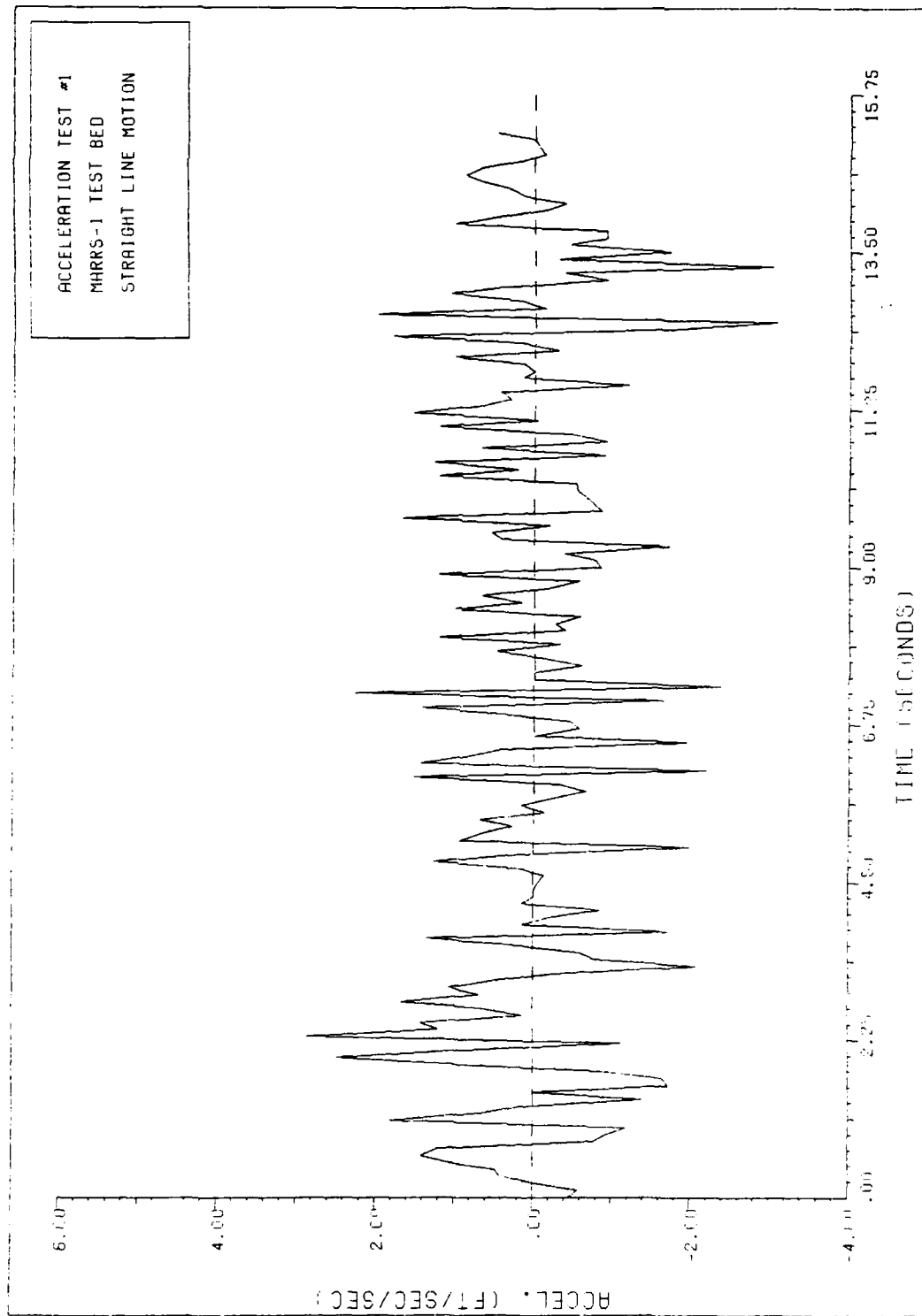


Figure 5.5 Acceleration Test #1.

testing was initiated since it did not require the use of the accelerometer subsystem.

Phase III

The purpose of Phase III testing was to demonstrate the feasibility of using gyro heading data for closed loop control of the MARRS-1 steering motor. This effort produced a navigation program for the Nav computer that requests heading data from the GYRAC system and issues commands to a control program in the drive computer (see Appendix H for listings, structure charts, and operating instructions for both programs). The robot will rotate in place until locked on the specified heading. It then follows the given heading correcting for course errors as it moves until manually stopped. In addition, at each point where a course correction is considered the heading data is transmitted to an external computer for storage and off line analysis (see Appendix K).

Three problems surfaced during the design and testing of this system. First, a communication execution speed problem; second, a steering motor response problem; and third, a steering over correction problem.

Implementation of this system of navigation routines required communication between four different computers: the Nav, GYRAC, Drive, and external computers. The manner in which these communication links and interfaces were

implemented have a significant impact on the navigation performance.

The link between the GYRAC and the Nav computer is an RS-232 line operating at 9600 baud. As used in this application, one byte commands are issued by the Nav computer and two bytes of heading data are returned by the GYRAC computer. The communication programs at both ends of the link are written in assembly language to make the link perform efficiently and quickly. This link performed without error and did not significantly slow down the navigation process.

The link between the Nav computer and the external computer is very similar to the GYRAC-Nav computer link. It also performed well and did not slow down the navigation process.

However, the link between the Nav computer and the Drive computer, as implemented, slowed down the course correction process. this resulted in impaired navigation performance. Once again, a 9600 baud RS-232 link was used. However, communication over the link does not use single byte commands and is only driven by assembly language communication routines at the Nav computer end.

To simplify implementation, the decision was made to use the existing Drive computer communication interface and assembly language control routines for the steering and drive motors. The Nav computer controls the operation by

sending six bytes of data representing a jump to subroutine command and a specific memory address (ASCII format). Execution of these subroutines by the Drive computer controls the steering and drive motors. Unfortunately, the Drive computer communications software interface requires a small time delay between bytes of data. In addition, each Drive computer motor control subroutine executes a voice command, READY, before returning control of the system back to the communications routine. These two unnecessary time delays limit the Drive computer to at most one command per second which limits the rate at which course corrections can be made.

The command communication problem is further compounded by a slow steering motor response. The steering motor does not move instantly from one position to another. It takes as long as four to five seconds to move 180 degrees. In addition, once the wheel is turned to the desired angle it takes a finite amount of time for this change to produce a measurable course correction. Small changes in wheel direction can produce large changes in robot heading if given sufficient time for movement, but the robot will be off course for this entire time period.

The solution used to alleviate these problems is time delay. Time delays are executed for each steering command to allow the wheel sufficient time to move to the directed position. Small time delays were also added after each

course correction to allow time for the wheel direction change to take effect.

A steering over correction problem occurs when the steering wheel is turned for course corrections. The Nav computer is not able to straighten the steering wheel onto the correct heading before the robot has overshoot the desired course. This causes the robot to oscillate around the given course resulting in an unstable system where the overcorrections become increasingly large.

This problem has several causes. First, the gyro heading data is measured in increments of approximately 0.088 degrees, but the robot can not set a course to this accuracy since the steering stepper motor moves in one degree increments. Therefore, it must alternately switch between two adjacent steering stepper motor settings to follow most headings.

Second, due to irregularities in the floor and an unbalanced weight distribution of the robot platform over its wheels, the robot drifts from a "straight course" even if the steering wheel is locked in the center position.

Third, course corrections are made in one degree increments each time the heading is sampled and found to deviate from the desired course. If a large course correction must be made, many wheel turn commands will be issued causing a sharp wheel angle to be present when the desired heading is detected. The wheel can only be

straightened by many more wheel turn commands in the opposite direction. However, during this time the robot will continue moving in the wrong direction incurring a large overcorrection error.

The overcorrection problem was solved by defining a heading window formed by dropping the least significant four bits of the twelve bit heading reading. This makes the window approximately 1.5 degrees wide with the desired heading located on one of the sixteen possible headings inside the window (unfortunately not centered in the window). Any heading inside the window is defined to be the "correct course". The unstable oscillations are damped by moving the steering wheel to the center position (straight ahead) with a single command as soon as the edge of the heading window is detected. Detected headings within the heading window do not produce a course correction, but allow the robot to continue moving straight ahead (steering wheel centered).

Additional time delays have been provided after each course correction to allow small steering changes more time to take effect. This works well only if the robot begins its movement within or near the heading window. To insure that this always occurs, a rotate-robot-to-heading-window routine is executed before following the directed heading. This aligns the robot's heading within the heading window before forward motion is started.

Figure 5.6 graphically shows the heading data for a 33 foot robot run where the robot's initial heading was within the heading window. No rotation occurred since the initial heading was 88.15 degrees which is inside the heading window. The heading samples are not evenly distributed in time, but occur at course correction decision points. Notice that the robot still oscillates around the given heading, but the oscillations are damped making the navigation system stable. This figure also shows that few detected headings are in the heading window which produces many course corrections and therefore small oscillations around the window.

Figure 5.7 shows a 33 foot robot run where the initial heading was not within the heading window. An initial heading of 66.53 degrees is not shown on the graph since the robot rotated without forward motion to 88.15 degrees which is the first point shown in the figure. Notice that the rotate routine aligned the robot's heading within the heading window. Figure 5.7 also shows the course tracking accuracy that can be obtained when the navigation system "locks on" to a course inside the window.

As in the previous figure, Figure 5.8 does not show the initial heading of 112.67 degrees. The robot automatically aligned itself inside the heading window by rotating to a heading of 87.54 degrees. Notice that a large number of the heading points are again outside the window resulting in

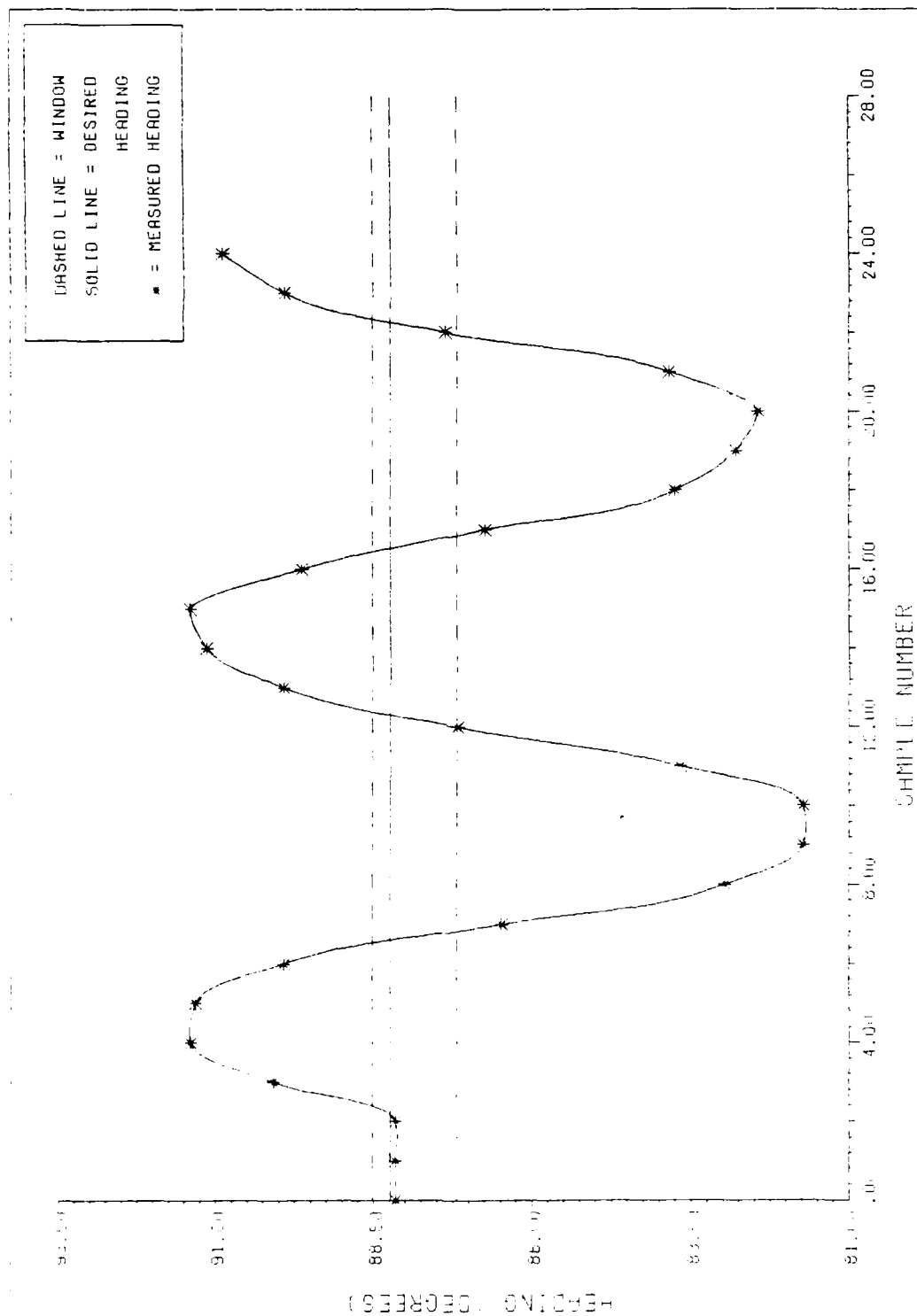


Figure 5.6. GYRAC Navigation Test #1, Automatic control of MARRS-1 Heading.

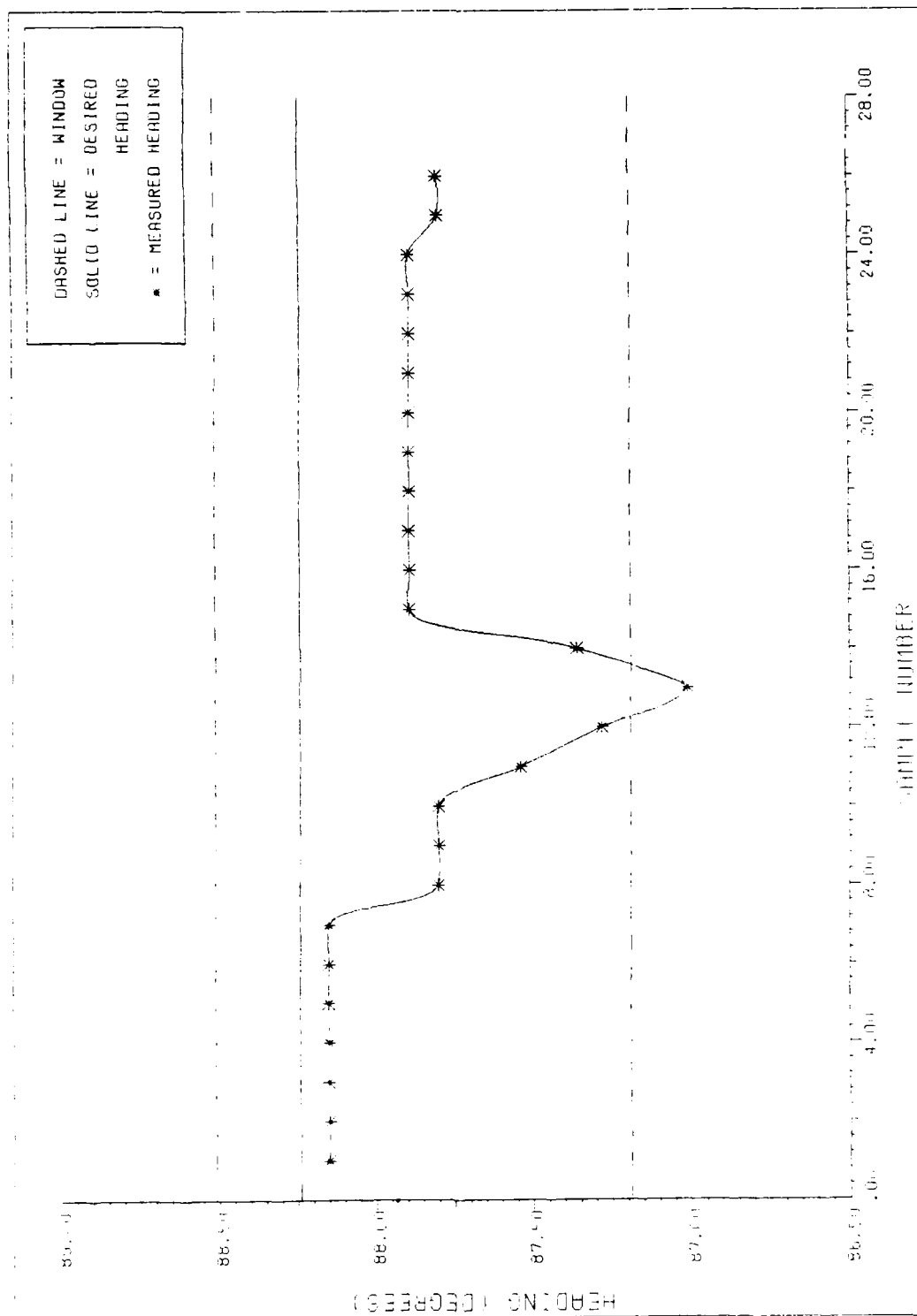


Figure 5.7. GYRAC Navigation Test #2, Automatic Control of MARRS-1 Heading.

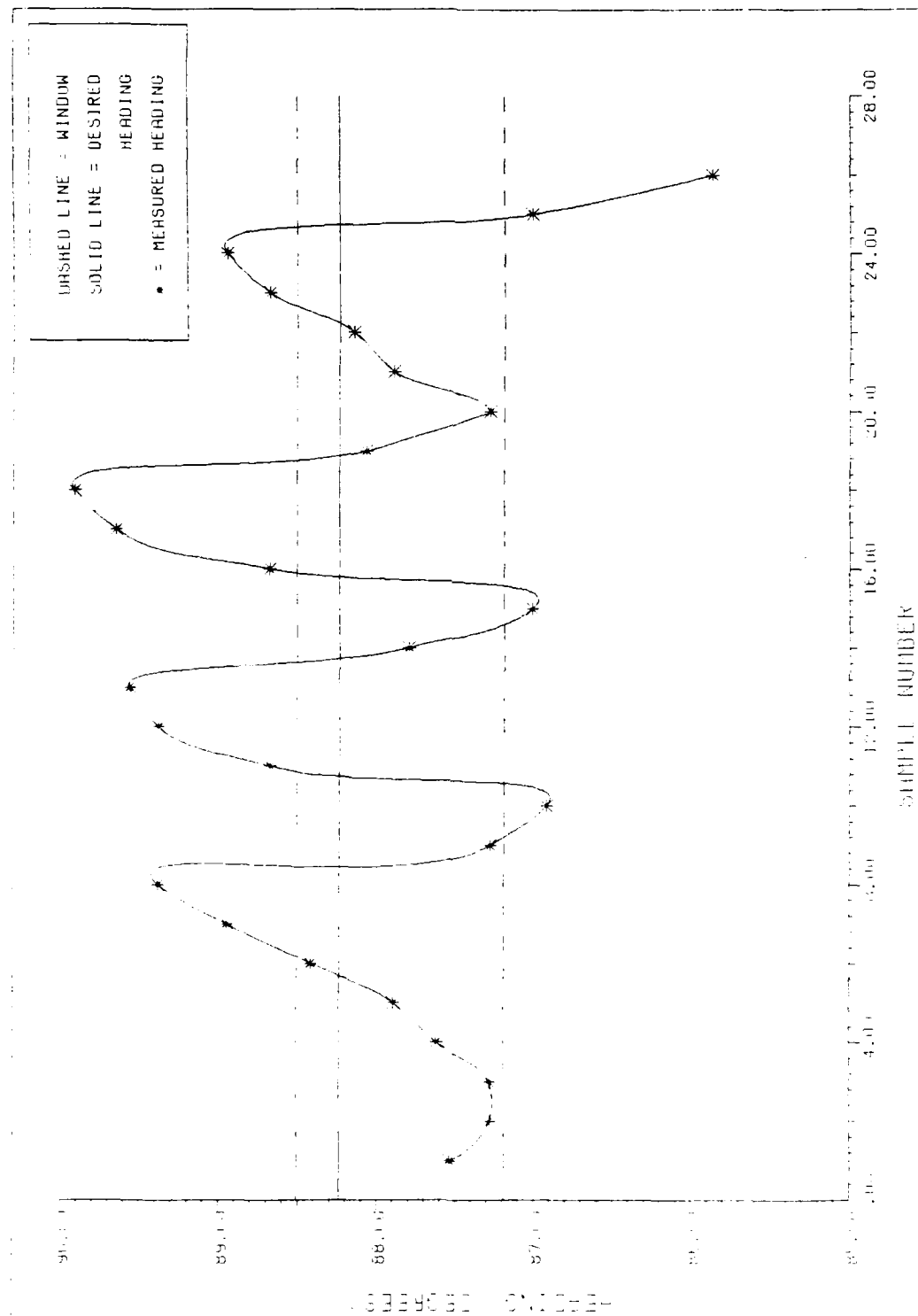


Figure 5.8. GYRAC Navigation Test #3, Automatic Control of MARRS-1 Heading.

course oscillation. However, the oscillations are not as severe as in Figure 5.6.

Careful study of all three figures indicate large oscillations occur when larger wheel angles (from the center position) are used. This causes frequent course changes to be executed because detected headings are not within the heading window. However, because course deviations oscillate around the actual course the mean course was very close to the desired course. The worst case oscillations resulted in movement of plus or minus five inches around the desired heading. No course drift was observed which is supported by a worst case final destination error of five inches. Therefore, feasibility of gyro based robot navigation has been demonstrated.

Review of Assumptions

The purpose of this section is to address the validity and impact of each assumption made in Chapter I and Chapter II.

Five governing assumptions were set forth in Chapter I. The first assumption, concerning local magnetic disturbances, proved to be valid. The magnetic flux detector was aligned only once and throughout the testing of the GYRAC, consistent heading information was obtained at all points in the test area.

The second assumption, that of a perfectly smooth and level operating surface, was the nemesis of this thesis. As

mentioned under Phase II testing, the effect of accelerometer tilt error was far greater than anticipated. As a result, the GYRAC velocity data can not be used for navigation.

The validity of the remaining three assumptions (a perfect integrator, constant velocity over sample period, and precisely known sample time), all dealing with the accelerometer subsystem, could not be determined due to the inaccuracy in sensed acceleration. The effect of each of these assumptions is expected to be small given an accurate acceleration measure.

In Chapter II, it was assumed that the bias and scale factor errors would be negligible. The effect of this assumption could not be determined. However, due to the very small acceleration levels involved with the movement of MARRS-1, the bias and scale factor errors could have a significant impact on the accuracy of the sensed acceleration. In this case, they would have to be compensated for.

Miscellaneous

The gyro base assembly, which serves as the power source for the entire GYRAC system, was noticed to get extremely warm during operation of the GYRAC. To avoid damage to the base assembly, a separate source of +5 volt power was instituted. A separate external power supply is

used to source the +5 volts and is provided through the same cable as is used for the system +28 volts external supply. This modification greatly reduced the load on the gyro base assembly and corrected the heating problem. The GYRAC power panel diagram in Appendix E has been updated to reflect this change.

VI. Summary, Conclusions, and Recommendations

Summary and Conclusions

The purpose of this thesis was to design and fabricate a real time, point to point, closed loop, mobile autonomous robot navigation system for the MARRS-1 robot. Specifically, the thesis was limited to three primary tasks. The first task was to develop the GYRAC system which would be capable of providing heading and velocity data. The second task was to integrate the GYRAC onto the MARRS-1 robot for verification testing, and the third task was to demonstrate autonomous navigation with the MARRS-1/GYRAC system.

Each of the these three parts was completed with the last task being completed to a limited extent. The GYRAC system is a complete and functional unit. However, the velocity data from the GYRAC is not usable for navigation. As mentioned in Chapter V, the true acceleration due to motion rarely, if ever, gets above the tilt error sensed by the accelerometer. This results in an ambiguous acceleration signal and thus an inaccurate velocity signal.

This problem has two major causes; the acceleration actually experienced by the MARRS-1 as it travels across the floor is very small in magnitude and short in duration; and, there is no error compensation in the accelerometer subsystem for gravity induced (tilt) errors. Nothing can be done about the small accelerations experienced by the

MARRS-1, since it is an inherently slow moving robot. Furthermore, speeding up the movement of MARRS-1 would not solve the problem since any flexible robot navigation system must be able to perform well at all reasonable speeds. This means that to make the GYRAC a completely usable system, a method of isolating the accelerometer from gravity tilt error must be incorporated. There are several methods for overcoming this tilt error problem. Several of these are presented in the Recommendations section.

The problems encountered with the accelerometer subsystem should not overshadow the success with the remainder of the GYRAC system. The GYRAC has proven to be a very reliable and accurate source of heading data. This heading data can be used by any robot system with an RS-232 serial interface. In addition, the heading data available from the GYRAC can be with respect to any reference direction and only one alignment along this reference is necessary. Subsequently, the GYRAC need only be powered up and will automatically provide accurate heading data with respect to the aligned reference. This GYRAC heading data could be combined with a separate source of distance measurement, such as wheel optical shaft encoders, to produce a viable navigation system. This could be accomplished on the MARRS-1 through software alone and is discussed further in the Recommendations section.

The GYRAC is completely integrated onto the MARRS-1 and

several tests of the integrated system have been completed verifying the compatibility of the two systems. Numerous software routines were produced allowing for communication between the MARRS-1 and the GYRAC and for data gathering and processing purposes. Complete detail of these programs can be found in Appendices G, H, and I.

The MARRS-1 is not yet capable of autonomous navigation, but a step was made toward that goal. The MARRS-1 can follow a given heading under self control of the steering stepper motor. MARRS-1 can be initially positioned at any heading and under self control it will rotate in place until it is aligned along a programmed heading, straighten out the front wheel, and begin forward movement making steering corrections as it travels in order to maintain its heading. Currently, the MARRS-1 will follow the programmed heading until manually stopped. With the addition of a distance measurement to the control algorithms, the MARRS-1 could be programmed to move autonomously about the test area.

Finally, the importance of robot world modelling and path planning to autonomous navigation should not be taken lightly. Before a robot can begin to move it must have some knowledge of its surroundings and it must be able to plan out collision free and efficient pathways through its environment. Some of the current techniques of world modelling have been presented along with a new approach.

Path planning, under this new approach consists of finding an unobstructed pathway to the goal point. Safe points are used only if a direct path does not exist. The details of this path planning have been developed through a simple example.

Recommendations

There was not time to accomplish many of the goals optimistically set forth at the beginning of this thesis effort. In addition, throughout the development of the GYRAC system and while working with the MARRS-1 robot, many problems were identified too late to correct and many new ideas were conceived too late to implement. Therefore, the following recommendations are offered as possible extensions of, or improvements to, this thesis:

1. To correct the tilt error problem with the accelerometer subsystem in the GYRAC, some type of error compensation is necessary. For example, two or more accelerometers could be added to the system. These accelerometers would be perpendicular to each other and perpendicular to the existing accelerometer. By aligning one accelerometer along the vertical and the other along the horizontal (but perpendicular to the existing accelerometer), a signal could be generated which would be proportional to the amount of tilt experienced by the platform. This signal could be subtracted from the original

accelerometer signal; thereby, nulling out the tilt error. Perpendicular accelerometer pairs are commercially available through Sundstrand Data Control Incorporated and other manufacturers. In addition, due to the small amount of acceleration actually experienced by the MARRS-1, another single-axis accelerometer should be purchased with much greater sensitivity. This accelerometer would replace the current QA-1100 in the GYRAC. A full scale range of plus or minus one "g" would be sufficient (the QA-1100 has a range of plus or minus 13 "g's") and would result in accelerometer readings which would be less susceptible to bias and scale factor errors. Also, a new integrator circuit should be designed with a much higher impedance to avoid loading the accelerometer internal servo circuit. This is necessary so the sensitivity of the accelerometer will not be effected by the integrator circuit. The new integrator circuit must also be designed with drift rate in mind, as described in Chapter V under Phase I testing.

2. Another possibility for correcting the tilt error problem would be to incorporate a displacement gyro. This gyro could be used to sense small displacement angles of the accelerometer into the vertical. This displacement, or tilt angle, could be used to generate a signal proportional to the amount of accelerometer tilt. This signal could then be used to null out the tilt error.

3. The tilt error problem could also be corrected by

using a stable platform, such as those used in inertial navigation systems (INS), to mount the accelerometer. Only a single axis platform would be required to maintain the accelerometer input axis in the horizontal plane. Taking this suggestion even further, a complete INS could be incorporated on the MARRS-1 or another robot instead of the GYRAC. An INS would provide velocity and direction information sufficient for navigation.

4. The GYRAC heading data could be combined with an external source of displacement data, such as the wheel shaft encoder data on MARRS-1, to produce data suitable for position determination.

5. Tests need to be conducted to compare the accuracy of computing the position of MARRS-1 based on pure wheel shaft data with computed position based on both GYRAC heading and wheel shaft data.

6. More work is necessary to refine the control of the MARRS-1 allowing it to follow a given heading. Reasonably accurate navigation was observed using the relatively simple approached outlined in Chapter V. However, several improvements can be made that should greatly improve performance.

First, the drive computer control programs and communication software should be rewritten in assembly language using single byte ASCII commands. This will eliminate command and communication time delays allowing

faster steering response (hundreds per second as opposed to one per second). This would also allow the heading window to be narrowed which would help reduce oscillations.

This change also requires the command routines of the Nav program to be changed, but nav.a (see Appendix H) has been designed in a three level structure to make changes and additions easy. The bottom level consists of hardware and device dependent routines such as transmit a byte of data to the Drive computer. The middle level of intermediate routines uses the lower level routines to define function primitives such as turn on drive motor at high speed. The top level of commands use the function primitives to form navigation commands such as rotate until locked on the heading window. Therefore, each level is independent of the way lower levels are implemented which limits the effects of changes to within a module.

Second, more intelligent steering control routines should be developed. They should anticipate when to start straightening out the front wheel instead of trying to do it all at once as was done in this thesis. In addition, they should be able to move the steering wheel in increments proportional to the amount of correction needed as opposed to single degree increments per correction. These additions will flatten out the oscillations and provide better navigation accuracy.

7. Once an accurate method of position determination

and of controlling the MARRS-1 is verified, the next step would be to develop the math routines necessary for MARRS-1 to perform the onboard processing required for navigation. The full world model as described in Chapter IV could then be implemented to provide the MARRS-1 with path planning capability.

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Vita

Captain Roland J. Bloom was born on 2 December 1958 in Grants, New Mexico. After graduating from White Pine High School, Ely, Nevada in 1977 he entered the United States Air Force Academy in Colorado Springs, Colorado.

Upon graduation from the Academy in May 1981, he received the degree of Bachelor of Science in Astronautical Engineering and was commissioned as a Regular Second Lieutenant of the United States Air Force.

He was assigned to the Peacekeeper Division of the 6595th Missile Test Group at Vandenberg AFB, California. While at Vandenberg AFB, Captain Bloom served as primary Test Controller for the assembly, check-out, and launch preparation of the first four Peacekeeper flight-test missiles.

In may 1984, he entered the Masters's program in Astronautical Engineering at the Air Force Institute of Technology.

Captain Bloom is married to the former Raylene A. Burgess of East Ely, Nevada. They have two children: Jessica and Brandon.

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Vita

Captain William J. Ramey Jr. was born December 25, 1952 in Great Falls, Montana. He graduated as class Valedictorian in 1971 from Jefferson High School, Boulder, Montana and later that year attended Montana State University. He enlisted in the Air Force in 1974 and became an Honor Graduate of the Cryptographic Equipment Maintenance School at Lackland AFB, Texas, where he was later assigned as an Instructor. He attended San Antonio College pursuing a degree in Electrical Engineering. In 1977, he was selected for the Airman Education and Commissioning Program and attended Texas A&M University where he received a Bachelor of Science degree in Electrical Engineering in 1979. In 1980, he graduated as a Distinguished Graduate from Officer Training School. Upon graduation, he was assigned to the National Security Agency (NSA) as a Computer Engineer and was later certified by NSA as a Senior Cryptologic Engineer. He attended the University of Maryland pursuing a Masters degree in Electrical Engineering. In 1984, Captain Ramey began a Masters program in Electrical Engineering at the Air Force Institute of Technology, Wright Patterson AFB, Ohio.

Captain Ramey is married to the former Elizabeth A. West of Farmingdale, New York. They have three children: Matthew, Joshua, and Katherine.

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APPENDIX A

SDC 1700 S/D Converter Data Sheet	A-2
AD573 A/D Converter Data Sheet	A-8
AD544 Operational Amplifier Data Sheet	A-15

SDC 1700 S/D Converter Data Sheet



Low Profile Synchro/Resolver-to-Digital Converter

SDC1700/1702/1704 SERIES

FEATURES

Internal Microtransformers for 60Hz, 400Hz and 2.6kHz

References

Low Profile (0.4")

10-, 12- or 14-Bit Resolution for 360°

High Tracking Rates (75 revs/sec)

Voltage Scaling with External Resistors (Unique Feature)

DC Voltage Output Proportional to Angular Velocity

Low Cost

Lightweight 3oz. (85 grams)

MIL Spec/Hi Rel Options Available

APPLICATIONS

Servo Mechanisms

Retransmission Systems

Coordinate Conversion

Antenna Monitoring

Simulation

Industrial Controls

Fire Control Systems

Machine Tool Control Systems

GENERAL DESCRIPTION

The SDC1700, SDC1702 and SDC1704 are modular, continuous tracking Synchro/Resolver-to-Digital Converters which employ a type 2 servo loop.

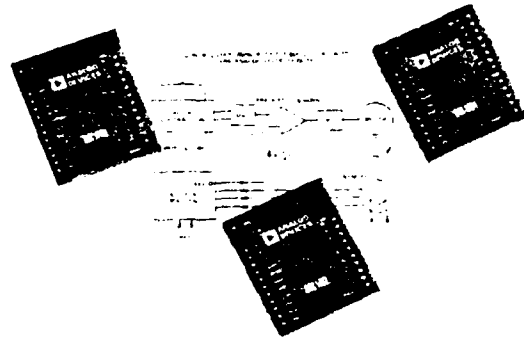
They are intended for use in both Industrial and Military applications.

The input signals can be either 3 wire synchro plus reference or 4 wire resolver plus reference, depending on the option. The outputs will be presented in TTL compatible, parallel natural binary.

One of the outstanding features of the converters is the use of precision Scott T and reference microtransformers. This has made it possible to include the transformers within the module, even on the 60Hz option, and yet still maintain the profile height of 0.4"

Particular attention has been paid in the design, to achieving the highest tracking rates and accelerations possible, compatible with the resolution and carrier frequency used, while at the same time obtaining a high overall accuracy.

When SDC's are used in control loops, it is often useful to have a voltage which is proportional to angular velocity. This voltage is available and has been brought out on all the SDC1700 converters.



Extended temperature range versions of all the converters are available.

MODELS AVAILABLE

The three Synchro-to-Digital Converters described in this data sheet differ primarily in the areas of resolution, accuracy and dynamic performance as follows:

Model SDC1702XYZ is a 10-bit converter which has an overall accuracy of ± 22 arc-minutes and a resolution of 21 arc-minutes.

Model SDC1700XYZ is a 12-bit converter with an overall accuracy of ± 8.5 arc-minutes and a resolution of 5.3 arc-minutes.

Model SDC1704XYZ is a 14-bit converter with an overall accuracy of ± 2.2 arc-minutes ± 1 LSB and a resolution of 1.3 arc-minutes.

The XYZ code defines the option thus: (X) signifies the operating temperature range, (Y) signifies the reference frequency, (Z) signifies the input voltage and range, and whether it will accept synchro or resolver format.

More information about the option code is given under the heading of "Ordering Information".

NOTE:

For all the standard options, no external transformers are needed with these converters.

SDC 1700 S/D Converter Data Sheet (continued)

SPECIFICATIONS (typical @ +25°C unless otherwise noted)

MODEL	SDC1702	SDC1700	SDC1704
ACCURACY ¹ (max error)			
50Hz	±22 arc minutes	±8.5 arc minutes	±2.2 arc minutes ±1LSB
400Hz	±22 arc minutes	±8.5 arc minutes	±2.2 arc minutes ±1LSB
2 kHz	±22 arc minutes	±8.5 arc minutes	±2.2 arc minutes ±1LSB
RESOLUTION	10 Bits (1LSB = 21 arc mins)	12 Bits (1LSB = 5.3 arc mins)	14 Bits (1LSB = 1.3 arc mins)
OUTPUT (in Parallel)	10 Bits (Natural Binary)	12 Bits (Natural Binary)	14 Bits (Natural Binary)
SIGNAL AND REFERENCE FREQUENCY	50Hz, 400Hz, 2 kHz	*	*
SIGNAL VOLTAGE (Line-to-Line)			
Low Level	11.8V rms	*	*
High Level	90V rms	*	*
SIGNAL IMPEDANCES			
Low Level	20kΩ (Resistive)	*	*
High Level	200kΩ (Resistive)	*	*
REFERENCE VOLTAGE			
Low Level	26V (11.8V Signal)	*	*
High Level	115V (90V Signal)	*	*
REFERENCE IMPEDANCE			
	270kΩ (115V Signal)	*	*
	56kΩ (26V Reference)	*	*
	(Impedance is Resistive)	*	*
TRANSFORMER ISOLATION	500V dc	*	*
TRACKING RATE (min)			
50Hz	5 Revolutions Per Second	*	500 ² sec
400Hz	30 Revolutions Per Second	*	12 Revolutions Per Second
2 kHz	75 Revolutions Per Second	*	25 Revolutions Per Second
Accel ²			
Constant K _A			
50Hz	1980/sec ²	*	520/sec ²
400Hz	110,000/sec ²	*	30,000/sec ²
2 kHz	518,000/sec ²	*	170,000/sec ²
STEP RESPONSE (179° Step) (For 1LSB Error)			
50Hz	1.5sec	*	*
400Hz	125ms	*	*
2 kHz	50ms	*	*
POWER LINES	±15V @ 25mA ±5% ±5V @ 20mA	*	±15V @ 30mA ±5% ±5V @ 85mA
POWER DISSIPATION	1.1 Watts	*	1.3 Watts
DATA LOGIC OUTPUT ³ (TTL Compatible)	2 TTL Loads SDC17026VZ 4 TTL Loads SDC17025VZ	2 TTL Loads SDC17006VZ 4 TTL Loads SDC17005VZ	2 TTL Loads on All Options
BUSY LOGIC OUTPUT POSITIVE PULSE (1 TTL Load)			
50Hz	9.0μs	*	9.0μs
400Hz	2.0μs	*	2.0μs
2 kHz	2.0μs	*	1.5μs
MAX DATA TRANSFER TIME			
50Hz	40μs	*	15μs
400Hz	5.0μs	*	1.0μs
2 kHz	1.8μs	*	0.8μs
INHIBIT INPUT (For Inhibit)	Logic 0 ⁴ (1 TTL Load)	*	Logic 0 ⁴ (2 TTL Loads)
WARM UP TIME	1 sec to Rated Accuracy	*	*
TEMPERATURE RANGE			
Operating	0 to +70°C Standard -55 to +125°C Extended	*	*
Storage	-55 to +125°C	*	*
DIMENSIONS	0.125" x 2.025" x 0.4"	*	*
	7.4 x 51.6 x 10.2mm	*	*
WEIGHT	1.05 g (0.5 ounce)	*	*

NOTES

¹ Accuracy same as SDC1702

² Test over the appropriate operating temperature range of the option and for 40% rms signal and reference amplitude variation, 50% rms signal and reference harmonic distortion, 0.15% power supply variation and 10% variation in reference frequency.

³ It is recommended that buffers should be used if the above converters

are required to drive over a distance greater than 6'

⁴ The output is correct over the range without notice.

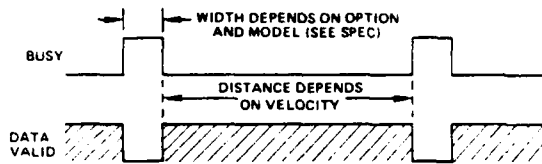
SDC 1700 S/D Converter Data Sheet (continued)

DATA TRANSFER (All Models)

The readiness of the converters for data transfer is indicated by the state of the BUSY pin.

The voltage appearing on the BUSY pin consists of a train of pulses, at TTL levels, of length according to the model and option (see specification table). The converter is busy when the BUSY pin is at a TTL "High" level. These pulses correspond to those delivered by the VCO to increment or decrement the up-down counter (see schematic diagram). Thus the pulses will occur for increasing and decreasing counts.

The most suitable time for transferring data is when the BUSY is at a logic "Lo" state, and the times allowable for data transfer are shown in the specification. Even at the maximum speed of the option, these times will be sufficient to transfer data before the next BUSY pulse occurs.



Data Transfer Diagram

DATA TRANSFER DIAGRAM

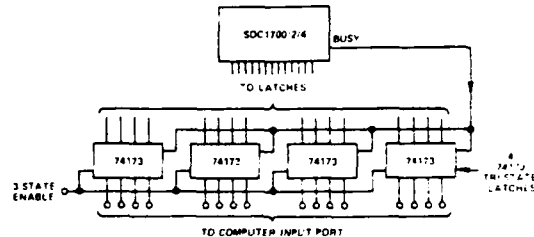
Taking the INHIBIT to a logic "Lo" state prevents the VCO (BUSY) pulses from updating the up-down counter. However, if applied during a BUSY pulse, the INHIBIT will not become effective until the end of the BUSY pulse.

The best method of transferring the data is by applying the INHIBIT (taking it to a logic "Lo" state), waiting for at least the width of a BUSY pulse, transferring the data and releasing the INHIBIT.

Note that sustained application of the INHIBIT opens the internal control loop and the converter may take on appreciable time to recover to full accuracy when the loop is restored.

INTERFACING WITH A COMPUTER

It is recommended that external latches are used to enable data to be transferred onto a computer data bus. One method is shown in the diagram. Using this method will mean that the latches are constantly updated by the BUSY signal, while at the same time enabling inputs to be made to the computer by means of normal data transfer procedures. The AC1755 mounting card contains these external components.



Suggested External Computer Interface Circuitry

THEORY OF OPERATION

If the unit is a Synchro-to-Digital Converter, then the 3 wire synchro output will be connected to S1, S2 and S3 on the module and the Scott T transformer pair will convert these signals into resolver format.

$$\begin{aligned} \text{i.e., } V_1 &= K E_0 \sin \omega t \sin \theta \\ V_2 &= K E_0 \sin \omega t \cos \theta \end{aligned}$$

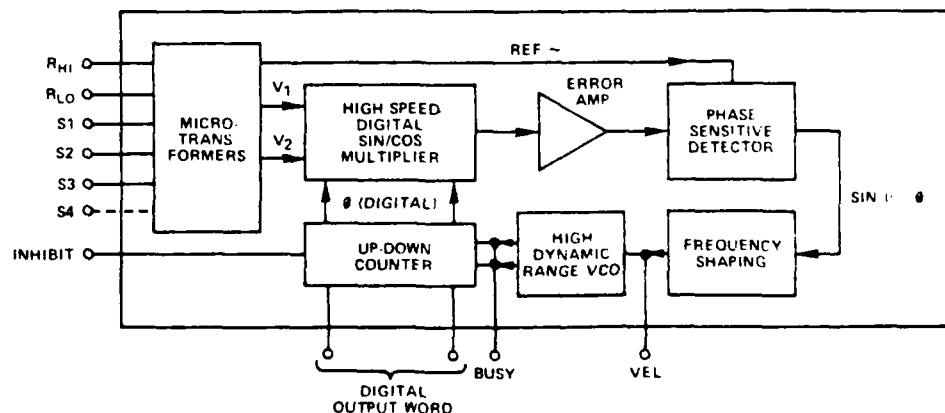
Where θ is the angle of the Synchro Shaft

If the unit is a Resolver-to-Digital Converter, then the 4 wire resolver output will be connected to S1, S2, S3 and S4 on the module and the microtransformer will act purely as an isolator.

To understand the conversion process, then assume that the current word state of the up-down counter is ϕ

The V_1 is multiplied by $\cos \phi$ and V_2 is multiplied by $\sin \phi$ to give

$$\begin{aligned} &K E_0 \sin \omega t \sin \theta \cos \phi \\ \text{and } &K E_0 \sin \omega t \cos \theta \sin \phi \end{aligned}$$



Functional Diagram of the SDC1700.2/4 Converters

SDC 1700 S/D Converter Data Sheet (continued)

These signals are subtracted by the error amplifier to give:

$$K E_O \sin \omega t (\sin \theta \cos \phi + \cos \theta \sin \phi) \\ \text{or } K E_O \sin \omega t \sin (\theta - \phi)$$

A phase sensitive detector, integrator and Voltage Controlled Oscillator (VCO) form a closed loop system which seeks to null $\sin (\theta - \phi)$.

When this is accomplished, the word state of the up-down counter (ϕ), equals within the rated accuracy of the converter, the synchro shaft angle θ .

CONNECTING THE CONVERTER

The electrical connections to the converter are straightforward. The power lines, which must not be reversed, are $\pm 15V$ and $5V$. They must be connected to the " $\pm 15V$ " and " $5V$ " pins with the common connection to the ground pin GND.

It is suggested that $0.1\mu F$ and $0.8\mu F$ capacitors be placed in parallel from $+15V$ to GND, from $-15V$ to GND and from $+5V$ to GND.

The digital output is taken from pins:

- 1 through 10 for the SDC1702
- 1 through 12 for the SDC1700
- 1 through 14 for the SDC1704

Pin 1 represents the MSB in each case. The reference connections are made to pins " R_{HI} " and " R_{LO} ".

In the case of a Synchro, the signals are connected to " $S1$ ", " $S2$ " and " $S3$ " according to the following convention:

$$E_{S1 - S3} = E_{RLO} - R_{HI} \sin \omega t \sin \theta \\ E_{S3 - S2} = E_{RLO} - R_{HI} \sin \omega t \sin (\theta + 120^\circ) \\ E_{S2 - S1} = E_{RLO} - R_{HI} \sin \omega t \sin (\theta + 240^\circ)$$

For a resolver, the signals are connected to " $S1$ ", " $S2$ ", " $S3$ " and " $S4$ " according to the following convention:

$$E_{S1 - S3} = E_{RLO} - R_{HI} \sin \omega t \sin \theta \\ E_{S2 - S4} = E_{RHI} - R_{LO} \sin \omega t \cos \theta$$

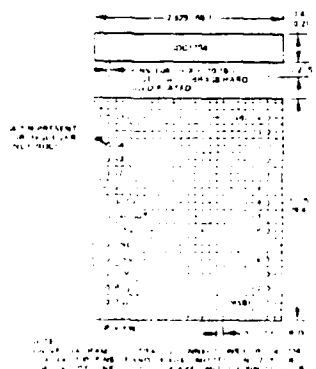
The analog voltage representing velocity is available between " VEL " and " GND ".

The " $BCSY$ " and " $INHIBIT$ " pin (if used), should be connected as described under the heading "Data Transfer".

NOTE: If the INHIBIT pin is used (i.e., driven to 0 volts), the control loop will be opened and a finite time will be required (see spec) for the converter to recover.

OUTLINE DIMENSIONS AND PIN CONNECTION DIAGRAM

Dimensions are shown in inches and (mm).



MATING SOCKET: CAMBION 450-3388-01-03

RESISTIVE SCALING OF INPUTS

A unique feature of the SDC1700 series of converters is that the inputs can be resistively scaled to accommodate any range of input signal and reference voltages.

This means that a standard converter can be used with a personality card in systems where a wide range of input and reference voltages are encountered. In addition it should be noted that a 400Hz unit will operate from a 2.6kHz reference. It will however have the velocity and acceleration characteristics as specified for the 400Hz converter. A 60Hz converter will operate from a 400Hz reference and will have the velocity and acceleration characteristics as specified for the 60Hz converter.

To calculate the values of the external scaling resistors for a synchro converter, add $1.11k\Omega$ in series with $S1$, $S2$ and $S3$ per extra volt in the case of the signal, and $2.2k\Omega$ in the case of the reference. In the case of a resolver converter add $2.22k\Omega$ per extra volt in series with $S1$ and $S2$ for the signal and $2.2k\Omega$ per extra volt in series with R_{HI} for the reference.

For example, assume that we have an 11.8 volt line to line signal/26.0 volt reference converter, and we wish to use a 60 volt line to line signal with a 115 volt reference.

Thus in each signal input line, the extra voltage capability required is:

$$60 - 11.8 = 48.2 \text{ volts}$$

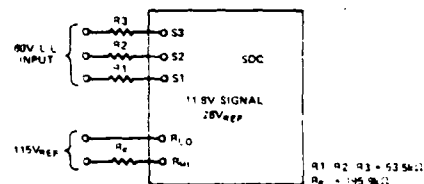
Therefore each resistor needs to have a value of $48.2 \times 1.11 = 53.5k\Omega$. In the case of the reference, the extra voltage capability required is:

$$115 - 26.0 = 89 \text{ volts}$$

Therefore the resistor needs to have a value of:

$$89.0 \times 2.2 = 195.8k\Omega$$

Thus the inputs can be scaled as in the diagram below.



NOTE: IN THE CASE OF $R1$, $R2$ AND $R3$ THE RATIO ERRORS BETWEEN THE RESISTANCES IS MORE IMPORTANT THAN THE ABSOLUTE RESISTANCE VALUES.

IN GENERAL A 1% RATIO ERROR WILL GIVE RISE TO AN EXTRA INACCURACY OF 17 ARC MINUTES WHILE A RATIO ERROR OF 0.1% WILL GIVE RISE TO AN EXTRA INACCURACY OF 1.7 ARC MINUTES.

THE ABSOLUTE VALUE OF $R6$ IS NOT CRITICAL.

BIT WEIGHT TABLE

Bit Number	Weight in Degrees
1 (MSB)	150.0000
2	75.0000
3	45.0000
4	22.5000
5	11.2500
6	5.6250
7	2.8125
8	1.4063
9	0.7031
10 (LSB for SDC1702)	0.3516
11	0.1758
12 (LSB for SDC1700)	0.0879
13	0.0439
14 (LSB for SDC1704)	0.0220

SDC 1700 S/D Converter Data Sheet (continued)

VELOCITY PIN

This pin provides a voltage output which is proportional to the angular velocity of the input. The voltage goes negative for an increasing digital angle and goes positive for a decreasing digital angle.

The characteristics of the velocity pin output are given in the table below.

Scaling of Output Voltage for One Full max. Velocity	2Volts (Nominal)
Output Voltage Temp. Coeff.	0.05%/°C of Output
Output Voltage Drift (All Models)	0 to +70°C ±50µV/°C -55°C to +105°C ±100µV/°C
Linearity	0 sec to 800/sec SDC1704 400Hz 1% 0 sec to 100/sec SDC1704 60Hz 1% 0 sec to 800/sec SDC1700/2 400Hz 2% 0 sec to 100/sec SDC1700/2 60Hz 1.5%
Noise (0 to 20Hz)	1600/sec SDC1700/2/4 400Hz 2mV rms 200/sec SDC1700/2/4 60Hz 2mV rms
Impedance (Output)	1Ω
max Current Available	1mA

The velocity voltage can be used in closed loop servo systems for stabilization instead of a tachometer.

The SDC1700/2/4 velocity outputs do not have the disadvantages of being inefficient at low speeds and do not need gearing required by tachometers. In addition, the output is available at no extra cost.

For other velocity output scaling and linearity consult the factory.

Two examples of the use of the velocity pin are shown in the diagram below.

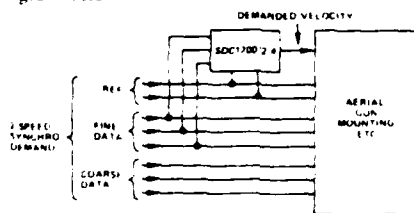


Diagram showing a velocity feed forward application. The SDC is used to produce the demanded velocity from Synchro form inputs.

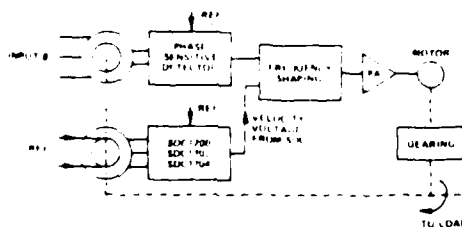
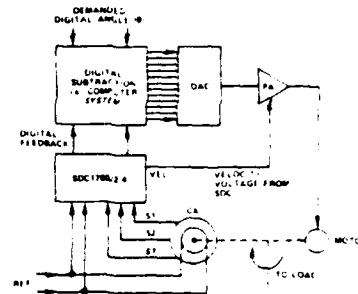


Diagram showing the velocity voltage being used to stabilize an electro-mechanical control loop.

APPLICATIONS OF SYNCHRO-TO-DIGITAL CONVERTERS

SDCs can be used in a variety of ways in control loops as well as for the conversion of angular data into a form which is readily acceptable to digital displays or computers.

The diagram below shows an SDC being used in a digitally controlled feedback loop.



An SDC Being Used in a Digitally Controlled Feedback Loop

Such loops as shown in the diagram above require the high dynamic performance of the SDC1700 series converters. It should be noted that in this application, the SDC1700 series will replace conventional tachometers and phase sensitive detectors while at the same time provide digital position feedback.

Many synchro systems employ a two speed, geared arrangement utilizing one synchro for the fine shaft and one for the coarse. An example of this type is shown below.

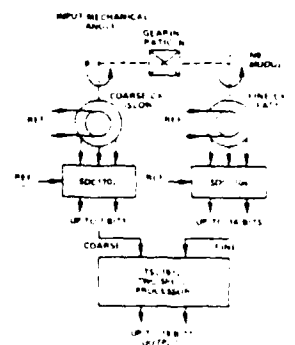


Diagram Showing Coarse-Fine Synchro Processor System

In the above example, two tracking SDC's are being used to provide data for coarse/fine (two speed) data transmission systems.

The TSL1612 is a processor which combines the outputs of two SDC's to provide one output word of up to 19 bits in length.

The TSL1612 is available for any ratio between 2:1 and 36:1 and provides automatic compensation for misalignment of the coarse synchro relative to its shaft. It also corrects for any overlap between the digits of the coarse and fine shafts.

SDC 1700 S/D Converter Data Sheet (continued)

MEAN TIME BETWEEN FAILURES (M.T.B.F.)

The estimated mean time between failures is given as follows:

SDC1700/2	174,000 Hours
SDC1704	167,000 Hours

Further information relating to M.T.B.F. and to the quality control and test procedures employed by us can be obtained from the factory on request.

TRANSFER FUNCTION

The transfer function of the SDC1700/2 and SDC1704, 400Hz versions, is given below.

For the transfer functions of the other models or for a detailed analysis of those given here, please contact us.

SDC1700/2 400Hz

$$\frac{\theta_0}{\theta_1} = \frac{8.8 \times 10^7 (1 + 0.8 \times 10^{-3} s)}{s^3 + 8.04 \times 10^2 s^2 + 6.1 \times 10^5 s + 8.8 \times 10^7}$$

SDC1704 400Hz

$$\frac{\theta_0}{\theta_1} = \frac{2.95 \times 10^7 (1 + 8.2 \times 10^{-3} s)}{s^3 + 8.05 \times 10^2 s^2 + 1.95 \times 10^5 s + 2.95 \times 10^7}$$

CARD MOUNTING

All the converters can be mounted on an AC1755 mounting card. This card contains the latches described under the "Data Transfer" heading, which are necessary to transfer the data on to a computer bus system, and sockets for the converter.

The latches have a tri-state output to facilitate ease of use.

The AC1755 also contains facilities for the inclusion of input signal and reference scaling resistors as described under the heading "Resistive Scaling of Inputs".

The card uses a 22/22 0.156" pitch edge connector. The pin out is shown below. If it is not required to use the external latches, they can be jumpered on the board.

AC1755 MOUNTING CARD

Dimensions shown in inches and (mm).

First Angle
Projection

EDGE CONNECTIONS AC1755

Pin Number	Function	Pin Number	Function
1	REF 1	17	REF 2
2	REF 1	18	REF 2
3	REF 1	19	REF 2
4	REF 1	20	REF 2
5	REF 1	21	REF 2
6	REF 1	22	REF 2
7	REF 1	23	REF 2
8	REF 1	24	REF 2
9	REF 1	25	REF 2
10	REF 1	26	REF 2
11	REF 1	27	REF 2
12	REF 1	28	REF 2

ORDERING INFORMATION

Parts should be ordered by the appropriate part number (i.e.,

VOL. II, 13-54 SYNCHRO & RESOLVER CONVERTERS

SDC1700, SDC1702, SDC1704) followed by the appropriate XYZ option code.

If the unit is to be a Resolver-to-Digital Converter, the SDC should be replaced by RDC in the part number.

The XYZ options are as follows:

X signifies the operating temperature range and the options are:

- X = 5 signifies 0 to +70°C (commercial) temperature.
- X = 6 signifies -55°C to +105°C (extended) temperature.

Y signifies the reference frequency and the options are:

- Y = 1 signifies 400Hz
- Y = 2 signifies 60Hz *
- Y = 4 signifies 2.6kHz

Z signifies the input signal and reference voltages and whether the converter is an SDC or an RDC. The options are:

- Z = 1 signifies synchro, signal 11.8V rms, reference 26V rms
- Z = 2 signifies synchro, signal 90V rms, reference 115V rms
- Z = 3 signifies resolver, signal 11.8V rms, reference 11.8V rms
- Z = 4 signifies resolver, signal 26V rms, reference 26V rms
- Z = 8 signifies resolver, signal 11.8V rms, reference 26V rms

Thus, for example, an SDC1704 with a commercial (0 to +70°C) operating range, using a 400Hz, 26V reference with an 11.8V signal would be ordered as an SDC1704511.

For other than these options, consult the factory.

CAUTIONS

Do not reverse the power supplies.

Do not connect signal and/or reference inputs to other than S1, S2, S3, S4, R_{HI} or R_{LO}.

Do not connect signals and/or references to a lower voltage rated converter. (Such as a 115V Synchro into a 26V Converter).

Misconnections as per the above will damage the units and void the warranty.

OTHER PRODUCTS

The SDC1700/2/4 converters are just a few of the modules and instruments concerned with Synchro and Resolver conversion manufactured by us.

Other products are listed below and technical data is available. If you have any questions about our products or require advice about the use of them for a particular application, please contact our Applications Engineering Department.

TWO SPEED PROCESSORS

Which utilize the digital outputs of two SDCs in a 2 speed coarse/fine system to produce one combined digital word of up to 19 bits in length. The TSL1612 in particular is available for any ratio between 2:1 and 36:1.

DIGITAL-TO-SYNCHRO CONVERTERS

Resolutions of between 10 and 14 bits are available.

BCD OUTPUT SYNCHRO-TO-DIGITAL CONVERTERS

The SBDC1752 and SBDC1753 are converters with a BCD instead of a binary output based upon the SDC1700. They have outputs of 0-360.0 degrees and 0 to 360.0 degrees respectively.

*60Hz Operation

For 60Hz operation, a 60Hz converter can be used with no reduction in accuracy.



Fast, Complete 10-Bit A/D Converter with Microprocessor Interface

AD573*

FEATURES

Complete 10-Bit A/D Converter with Reference, Clock and Comparator

Full 8- or 16-Bit Microprocessor Bus Interface

Fast Successive Approximation Conversion – 20 μ s typ

No Missing Codes Over Temperature

Operates on +5V and -12V to -15V Supplies

Low Cost Monolithic Construction

PRODUCT DESCRIPTION

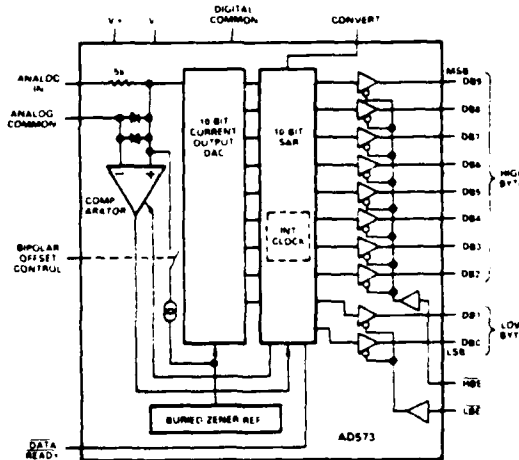
The AD573 is a complete 10-bit successive approximation analog to digital converter consisting of a DAC, voltage reference, clock, comparator, successive approximation register (SAR) and 3 state output buffers—all fabricated on a single chip. No external components are required to perform a full accuracy 10-bit conversion in 20 μ s.

The AD573 incorporates the most advanced integrated circuit design and processing technology available today. The successive approximation function is implemented with 1^{1/2} integrated injection logic. Laser trimming of the high stability SiCr thin film resistor ladder network at the wafer stage, LWT, insures high accuracy, which is maintained with a temperature compensated sub-surface Zener reference.

Operating on supplies of $-5V$ and $-12V$ to $-15V$, the AD573 will accept analog inputs of 0 to $-10V$ or $-5V$ to $-5V$. The trailing edge of a positive pulse on the CONVERT line initiates the 20 μs conversion cycle. DATA READY indicates completion of the conversion. HIGH BYTE ENABLE HBE and LOW BYTE ENABLE LBE control the 8-bit and 2-bit three state output buffers.

The AD573 is available in two versions for the 0 to +70°C temperature range, the AD573J and AD573K. The AD573S guarantees ± 1 LSB relative accuracy and no missing codes from -55°C to +125°C.

AD573 FUNCTIONAL BLOCK DIAGRAM



Two package configurations are offered. All versions are also offered in a 20-pin hermetically sealed ceramic DIP. The AD573J and AD573K are also available in a 20-pin plastic DIP.

PRODUCT HIGHLIGHTS

1. The AD573 is a complete 10-bit A/D converter. No external components are required to perform a conversion.
2. The AD573 interfaces to many popular microprocessors without external buffers or peripheral interface adapters. The 10 bits of output data can be read as a 10-bit word or as 8- and 2-bit words.
3. The device offers true 10-bit accuracy and exhibits no missing codes over its entire operating temperature range.
4. The AD573 adapts to either unipolar (0 to +10V) or bipolar (-5V to +5V) analog inputs by simply grounding or opening a single pin.
5. Performance is guaranteed with +5V and -12V or -15V supplies.

*Protected by U.S. Patent Nos. 3,940,760; 4,213,806; 4,136,349; 4,400,689, and 4,400,690

AD573 A/D Converter Data Sheet (continued)

SPECIFICATIONS

($T_A = 25^\circ\text{C}$, $V_+ = +5\text{V}$, $V_- = -12\text{V}$ or -15V , all voltages measured with respect to digital common, unless otherwise indicated)

Model	AD573J			AD573K			AD573S			Units
	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
RESOLUTION	10			10			10			Bits
RELATIVE ACCURACY ¹	± 1			± 1.2			± 1			LSB
$T_A = T_{min}$ to T_{max}	± 1			± 1.2			± 1			LSB
FULL SCALE CALIBRATION ²	± 2			± 2			± 2			LSB
UNIPOlar OFFSET	± 1			± 1.2			± 1			LSB
BIPOLAR OFFSET	± 1			± 1.2			± 1			LSB
DIFFERENTIAL NONLINEARITY ³	10			10			10			Bits
$T_A = T_{min}$ to T_{max}	9			10			10			Bits
TEMPERATURE RANGE	0		+70	0		+70	-55		+125	$^\circ\text{C}$
TEMPERATURE COEFFICIENTS ⁴										
Unipolar Offset	± 2			± 1			± 2			LSB
Bipolar Offset	± 2			± 1			± 2			LSB
Full Scale Calibration ²	± 4			± 2			± 5			LSB
POWER SUPPLY REJECTION										
Positive Supply										
+4.5V $V_+ = +5.5\text{V}$	± 2			± 1			± 2			LSB
Negative Supply										
-12.5V $V_- = -14.25\text{V}$	± 2			± 1			± 2			LSB
-12.5V $V_- = -11.4\text{V}$	± 2			± 1			± 2			LSB
ANALOG INPUT IMPEDANCE	3.0	5.0	7.0	3.0	5.0	7.0	3.0	5.0	7.0	k Ω
ANALOG INPUT RANGES										
Unipolar	0		+10	0		+10	0		+10	V
Bipolar	-5		+5	-5		+5	-5		+5	V
OUTPUT CODING										
Unipolar	Positive True Binary			Positive True Binary			Positive True Binary			
Bipolar	Positive True Offset Binary			Positive True Offset Binary			Positive True Offset Binary			
LOGIC OUTPUT										
Output Sink Current										
$V_{OL} = 0.4\text{V max}$, T_{min} to T_{max}	3.2			3.2			3.2			mA
Output Source Current ⁵										
$V_{OH} = 2.4\text{V max}$, T_{min} to T_{max}	0.5			0.5			0.5			mA
Output Leakage	± 40			± 40			± 40			μA
LOGIC INPUTS										
Input Current	± 100			± 100			± 100			μA
Logic "1"	2.0			2.0			2.0			V
Logic "0"	0.8			0.8			0.8			V
CONVERSION TIME										
$T_A = T_{min}$ to T_{max}	10	20	30	10	20	30	10	20	30	μs
POWER SUPPLY										
V_+	+4.5	+5.0	+7.0	+4.5	+5.0	+7.0	+4.5	+5.0	+7.0	V
V_-	-11.4	-15	-16.5	-11.4	-15	-16.5	-11.4	-15	-16.5	V
OPERATING CURRENT										
V_+	15	25		15	25		15	25		mA
V_-	9	15		9	15		9	15		mA
PACKAGE ⁶										
Ceramic DIP	D20A			D20A			D20A			
Plastic DIP	N20A			N20A						

NOTES

Relative accuracy is defined as the deviation of the code transition points from the ideal transfer point on a straight line from the zero to the full scale of the device.

Full scale calibration is guaranteed trimmable to zero with an external 50k potentiometer in place of the 15k fixed resistor.

Full scale is defined as 10 volts minus 1 LSB, or 9.999 volts.

Defined as the resolution for which no missing codes will occur.

Range from -15V to +25V, $T_A = 25^\circ\text{C}$ to T_{max} or T_{min} .

The data output lines have active pull-ups to source 0.5mA. The DATA READY line is open collector with a nominal 6k Ω internal pull-up resistor.

See Section 19 for package outline information.

Specifications subject to change without notice.

Specifications shown in boldface are tested on all production units at final electrical test. Results from those tests are used to calculate outgoing quality levels. All min and max specifications are guaranteed, although only those shown in boldface are tested on all production units.

AD573 A/D Converter Data Sheet (continued)

FUNCTIONAL DESCRIPTION

A block diagram of the AD573 is shown in Figure 1. The positive CONVERT pulse must be at least 500ns wide. DR goes high within 1.5µs after the leading edge of the convert pulse indicating that the internal logic has been reset. The negative edge of the CONVERT pulse initiates the conversion. The internal 10-bit current output DAC is sequenced by the integrated injection logic (I²L) successive approximation register (SAR) from its most significant bit to least significant bit to provide an output current which accurately balances the input signal current through the 5kΩ resistor. The comparator determines whether the addition of each successively weighted bit current causes the DAC current sum to be greater or less than the input current; if the sum is more, the bit is turned off. After testing all bits, the SAR contains a 10-bit binary code which accurately represents the input signal to within $\pm \text{LSB} = 0.05\%$ of full scale.

The SAR drives \overline{DR} low to indicate that the conversion is complete and that the data is available to the output buffers. \overline{HBE} and \overline{LBE} can then be activated to enable the upper 8-bit and lower 2-bit buffers as desired. \overline{HBE} and \overline{LBE} should be brought high prior to the next conversion to place the output buffers in the high-impedance state.

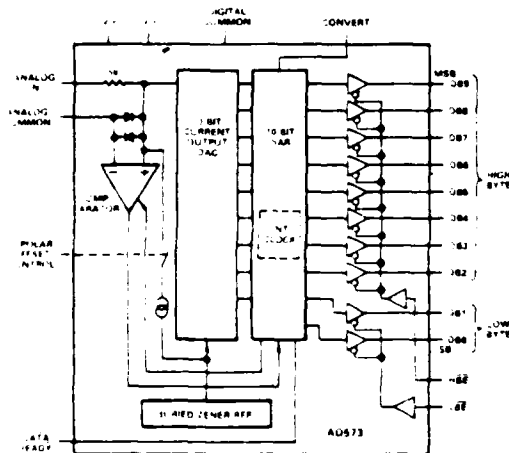


Figure 1 AD573 Functional Block Diagram

The temperature compensated buried Zener reference provides the primary voltage reference to the DAC and ensures excellent stability with both time and temperature. The bipolar offset input controls a switch which allows the positive bipolar offset current, exactly equal to the value of the MSB less \pm LSB, to be injected into the summing \pm node of the comparator to offset the DAC output. Thus the nominal 0 to \pm LSB unipolar input range becomes a \pm SV to \pm SV range. The 10 Ω thin film input resistor is trimmed so that with a full scale input signal, an input current will be generated which exactly matches the DAC input with all bias on.

UNIPOLAR CONNECTION

The AD573 contains all the active components required to perform a complete A/D conversion. Thus, for many applications, all that is necessary is connection of the power supplies, $+5V$ and $+12V$ to $+15V$, the analog input and the convert pulse. However, there are some features and special connections which should be considered for achieving optimum performance. The functional pin-out is shown in Figure 2.

The standard unipolar 0 to +10V range is obtained by shorting the bipolar offset control pin (pin 16) to digital common (pin 17).

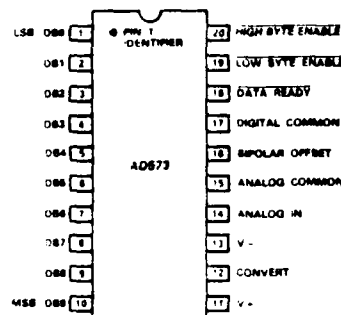


Figure 2. AD573 Pin Connections

Full Scale Calibration

The 5k Ω thin film input resistor is laser trimmed to produce a current which matches the full scale current of the internal DAC—plus about 0.3%—when an analog input voltage of 9.990 volts \pm 10 volts \pm 1LSB is applied at the input. The input resistor is trimmed in this way so that if a fine trimming potentiometer is inserted in series with the input signal, the input current at the full scale input voltage can be trimmed down to match the DAC full scale current as precisely as desired. However, for many applications the nominal 9.99 volt full scale can be achieved to sufficient accuracy by simply inserting a 15 Ω resistor in series with the analog input to pin 14. Typical full scale calibration error will then be within \pm 2LSB or \pm 0.2%. If more precise calibration is desired, a 50k Ω trimmer should be used instead. Set the analog input at 9.990 volts, and set the trimmer so that the output code is just at the transition between 11111111 10 and 11111111 11. Each LSB will then have a weight of 9.76mV. If a nominal full scale of 10.24 volts is desired (which makes the LSB have a weight of exactly 10.00mV), a 100k Ω resistor and a 100k Ω trimmer (or a 200k Ω trimmer with good resolution) should be used. Of course, larger full scale ranges can be arranged by using a larger input resistor, but linearity and full scale temperature coefficient may be compromised if the external resistor becomes a sizeable percentage of 5k Ω . Figure 3 illustrates the connections required for full scale calibration.

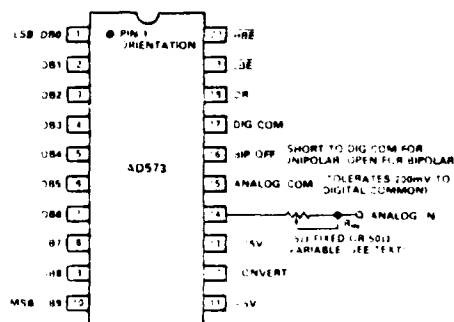


Figure 3 Standard AD573 Connections

Unipolar Offset Calibration

Since the Unipolar Offset is less than ± 1 LSB for all versions of the AD573, most applications will not require trimming. Figure 4 illustrates two trimming methods which can be used if greater accuracy is necessary.

Applying the AD573

Figure 4a shows how the converter zero may be offset by up to ± 3 bits to correct the device initial offset and/or input signal offsets. As shown, the circuit gives approximately symmetrical adjustment in unipolar mode.

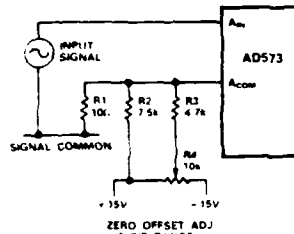


Figure 4a.

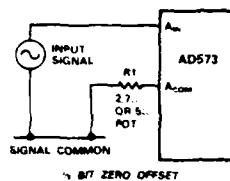


Figure 4b.

Figure 5 shows the nominal transfer curve near zero for an AD573 in unipolar mode. The code transitions are at the edges of the nominal bit weights. In some applications it will be preferable to offset the code transitions so that they fall between the nominal bit weights, as shown in the offset characteristics.

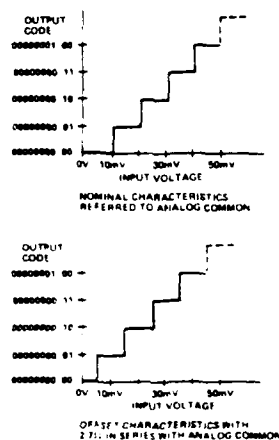


Figure 5. AD573 Transfer Curve - Unipolar Operation (Approximate Bit Weights Shown for Illustration, Nominal Bit Weights = 9.766mV)

This offset can easily be accomplished as shown in Figure 4b. At balance, after a conversion, approximately 2mA flows into the Analog Common terminal. A 2.7k resistor in series with the terminal will result in approximately the desired 1-bit offset of the transfer characteristics. The nominal 2mA Analog Common current is not closely controlled in manufacture. If high accuracy is required, a 50k potentiometer, connected as a rheostat, can be used as R1. Additional negative offset range may be obtained by using larger values of R1. Of course, if the zero transition point is changed, the full scale transition point will also move. Thus, if an offset of 1LSB is introduced, full scale trimming as described on the previous page should be done with an analog input of 9.985 volts.

NOTE: During a conversion, transient currents from the Analog Common terminal will disturb the offset voltage. Capacitive decoupling should not be used around the offset network. These transients will settle appropriately during a conversion. Capacitive

decoupling will "pump up" and fail to settle resulting in conversion errors. Power supply decoupling, which returns to analog signal common, should go to the signal input side of the resistive offset network.

BIPOLAR CONNECTION

To obtain the bipolar $-5V$ to $+5V$ range with an offset binary output code, the bipolar offset control pin is left open.

A -5.000 volt signal will give a 10-bit code of 00000000 00; an input of 0.000 volts results in an output code of 10000000 00 and $+4.99$ volts at the input yields the 11111111 11 code. The nominal transfer curve is shown in Figure 6.

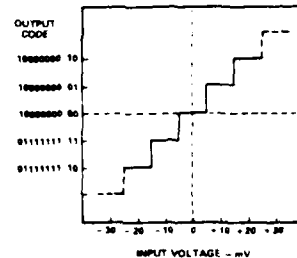


Figure 6. AD573 Transfer Curve - Bipolar Operation

Note that in the bipolar mode, the code transitions are offset 1/2LSB such that an input voltage of 0 volts $\pm 5mV$ yields the code representing zero (10000000 00). Each output code is then centered on its nominal input voltage.

Full Scale Calibration

Full Scale Calibration is accomplished in the same manner as in Unipolar operation except the full scale input voltage is $+4.985$ volts.

Negative Full Scale Calibration

The circuit in Figure 4a can also be used in Bipolar operation to offset the input voltage (nominally $-5V$) which results in the 00000000 00 code. R2 should be omitted to obtain a symmetrical range.

The bipolar offset control input is not directly TTL compatible but a TTL interface for logic control can be constructed as shown in Figure 7.

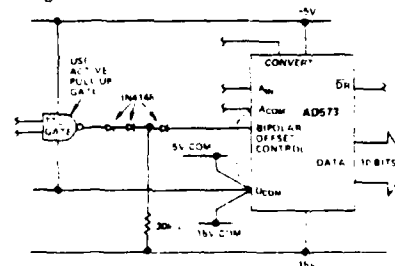


Figure 7. Bipolar Offset Controlled by Logic Gate Gate Output = 1 Unipolar 0 - 10V Input Range Gate Output = 0 Bipolar $\pm 5V$ Input Range

SAMPLE-HOLD AMPLIFIER CONNECTION TO THE AD573

Many situations in high-speed acquisition systems or digitizing rapidly changing signals require a sample-and-hold amplifier (SHA) in front of the A-D converter. The SHA can acquire and hold a

AD573 A/D Converter Data Sheet (continued)

signal faster than the converter can perform a conversion. A SHA can also be used to accurately define the exact point in time at which the signal is sampled. For the AD573, a SHA can also serve as a high input impedance buffer.

Figure 8 shows the AD573 connected to the AD582 monolithic SHA for high speed signal acquisition. In this configuration, the AD582 will acquire a 10 volt signal in less than 10 μ s with a droop rate less than 100 μ V/ms.

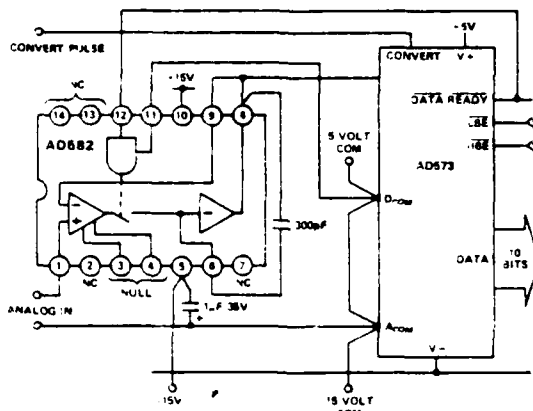


Figure 8. Sample-Hold Interface to the AD573

\overline{DR} goes high after the conversion is initiated to indicate that reset of the SAR is complete. In Figure 8 it is also used to put the AD582 into the hold mode while the AD573 begins its conversion cycle. The AD582 settles to final value well in advance of the first comparator decision inside the AD573.

\overline{DR} goes low when the conversion is complete placing the AD582 back in the sample mode. Configured as shown in Figure 8, the next conversion can be initiated after a 10 μ s delay to allow for signal acquisition by the AD582.

Observe carefully the ground, supply, and bypass capacitor connections between the two devices. This will minimize ground noise and interference during the conversion cycle.

GROUNDING CONSIDERATIONS

The AD573 provides separate Analog and Digital Common connections. The circuit will operate properly with as much as ± 200 mV of common mode voltage between the two commons. This permits more flexible control of system common bussing and digital and analog returns.

In normal operation, the Analog Common terminal may generate transient currents of up to 2mA during a conversion. In addition a static current of about 2mA will flow into Analog Common in the unipolar mode after a conversion is complete. The Analog Common current will be modulated by the variations in input signal.

The absolute maximum voltage rating between the two commons is ± 1 volt. It is recommended that a parallel pair of back-to-back protection diodes be connected between the commons if they are not connected locally.

CONTROL AND TIMING OF THE AD573

The operation of the AD573 is controlled by three inputs: CONVERT, \overline{HBE} and \overline{LBE} .

Starting a Conversion

The conversion cycle is initiated by a positive-going CONVERT

pulse at least 500ns wide. The rising edge of this pulse resets the internal logic, clears the result of the previous conversion, and sets \overline{DR} high. The falling edge of CONVERT begins the conversion cycle. When conversion is completed \overline{DR} returns low. During the conversion cycle, \overline{HBE} and \overline{LBE} should be held high. If \overline{HBE} or \overline{LBE} goes low during a conversion, the data output buffers will be enabled and intermediate conversion results will be present on the data output pins. This may cause bus conflicts if other devices in a system are trying to use the bus.

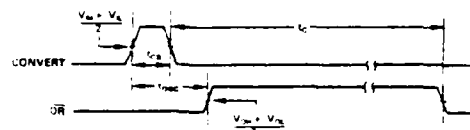


Figure 9. Convert Timing

Reading the Data

The three-state data output buffers are enabled by \overline{HBE} and \overline{LBE} . Access time of these buffers is typically 150ns (250 maximum). The Data outputs remain valid until 50ns after the enable signal returns high, and are completely into the high-impedance state 100ns later.

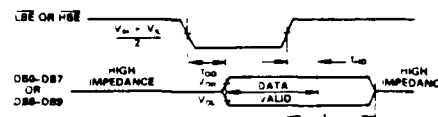


Figure 10. Read Timing

TIMING SPECIFICATIONS (All grades, $T_A = T_{max} - T_{min}$)

Parameter	Symbol	Min	Typ	Max	Units
CONVERT Pulse Width	t_{CS}	500	—	—	ns
\overline{DR} Delay from CONVERT	t_{DSC}	—	1	1.5	μ s
Conversion Time	t_C	10	20	30	μ s
Data Access Time	t_{OD}	0	150	250	ns
Data Valid after \overline{HBE} , \overline{LBE}					
High	t_{HD}	50	—	—	ns
Output Float Delay	t_{FL}	—	100	200	ns

MICROPROCESSOR INTERFACE CONSIDERATIONS - GENERAL

When an analog-to-digital converter like the AD573 is interfaced to a microprocessor, several details of the interface must be considered. First, a signal to start the converter must be generated; then an appropriate delay period must be allowed to pass before valid conversion data may be read. In most applications, the AD573 can interface to a microprocessor system with little or no external logic.

The most popular control signal configuration consists of decoding the address assigned to the AD573, then gating this signal with the system's \overline{WR} signal to generate the CONVERT pulse, and gating it with \overline{RD} to enable the output buffers. The use of a memory address and memory \overline{WR} and \overline{RD} signals denotes "memory-mapped" I/O interfacing, while the use of a separate I/O address space denotes "isolated I/O" interfacing. In 8-bit bus systems, the 10-bit AD573 will occupy two locations when data is to be read, therefore, two usually consecutive addresses must be decoded. One of the addresses can also be used as the address which produces the CONVERT signal during \overline{WR} operations.

Figure 11 shows a generalized diagram of the control logic for

AD573 A/D Converter Data Sheet (continued)

Interfacing to the AD573

an AD573 interfaced to an 8-bit data bus, where two addresses (ADC ADDR and ADC ADDR + 1) have been decoded. ADC ADDR starts the converter when written to; the actual data being written to the converter does not matter and contains the high byte data during read operations. ADC ADDR + 1 performs no function during write operations, but contains the low byte data during read operations.

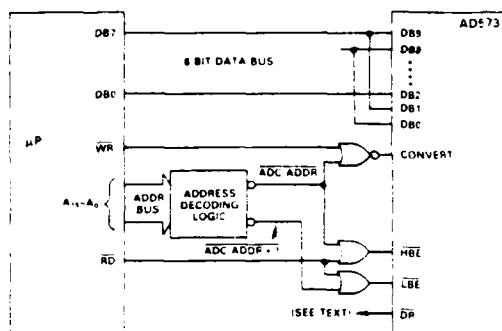


Figure 11. General AD573 Interface to 8-Bit Microprocessor

In systems where this read-write interface is used, at least 30 microseconds (the maximum conversion time) must be allowed to pass between starting a conversion and reading the results. This delay or "timeout" period can be implemented in a short software routine such as a countdown loop, enough dummy instructions to consume 30 microseconds, or enough actual useful instructions to consume the required time. In tightly-timed systems, the DR line may be read through an external three-state buffer to determine precisely when a conversion is complete. Higher-speed systems may choose to use DR to signal an interrupt to the processor at the end of a conversion.

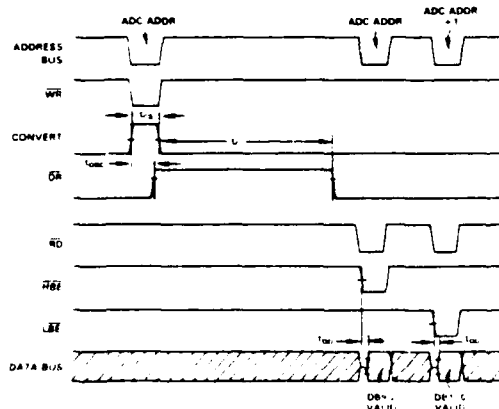


Figure 12. Typical AD573 Interface Timing Diagram

CONVERT Pulse Generation

The AD573 is tested with a CONVERT pulse width of 500ns and will typically operate with a pulse as short as 300ns. However, some microprocessors produce active WR pulses which are shorter than this. Either of the circuits shown in Figure 13 can be used to generate an adequate CONVERT pulse for the AD573.

In both circuits, the short low-going WR pulse sets the CONVERT line high through a flip-flop. The rising edge of DR, which signifies that the internal logic has been reset, resets the flip-flop and brings CONVERT low, which starts the conversion.

Note that t_{DSE} is slightly longer when the result of the previous conversion contains a logic 1 on the LSB. This means that the actual CONVERT pulse generated by the circuits in Figure 13 will vary slightly in width.

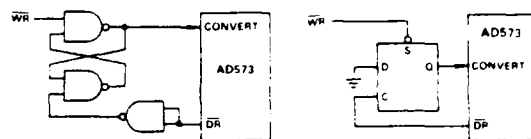


Figure 13a. Using 74LS00 Figure 13b. Using 1 74LS74

Output Data Format

The AD573 output data is presented in a left-justified format. The 8 MSBs (DB9-DB2, pins 10 through 3) are enabled by HBE (pin 20) and the 2 LSBs (DB1, DB0 - pins 2 and 1) are enabled by LBE (pin 19). This allows simple interface to 8-bit system buses by overlapping the 2 MSBs and the 2 LSBs. The organization of the data is shown in Figure 14.

When the least significant bits are read, LBE brought low, the six remaining bits of the byte will contain meaningless data. These unwanted bits can be masked by logically ANDing the byte with 11000000 (C0 hex), which forces the 6 lower bits to logic 0 while preserving the two most significant bits of the byte.

Note that it is not possible to reconfigure the AD573 for right justified data.

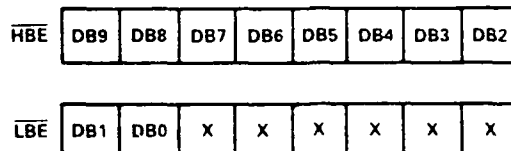


Figure 14. AD573 Output Data Format

In systems where all 10 bits are desired at the same time, HBE and LBE may be tied together. This is useful in interfacing to 16-bit bus systems. The resulting 10-bit word can then be placed at the high end of the 16-bit bus for left justification or at the low end for right justification.

It is also possible to use the AD573 in a "stand-alone" mode, where the output data buffers are automatically enabled at the end of a conversion cycle. In this mode, the DR output is wired to the HBE and LBE inputs. The outputs thus are forced into the high-impedance state during the conversion period, and valid data becomes available approximately 500ns after the DR signal goes low at the end of the conversion. The 500ns delay allows propagation of the least significant bit through the internal logic.

This mode is particularly useful for bench-testing of the AD573, and in applications where dedicated I/O ports of peripheral interface adapter chips are available.

AD573 A/D Converter Data Sheet (continued)

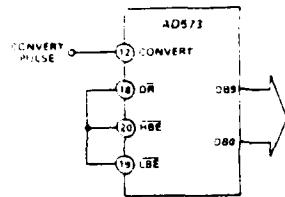


Figure 15. AD573 in "Stand-Alone" Mode
(Output Data Valid 500ns After DR Goes Low)

Apple II Microcomputer Interface

The AD573 can provide a flexible, low-cost analog interface for the popular Apple II microcomputer. The Apple II, based on a 1MHz 6502 microprocessor, meets all timing requirements for the AD573. Only a few TTL gates are required to decode the signals available on the Apple II's peripheral connector. The recommended connections are shown in Figure 16.

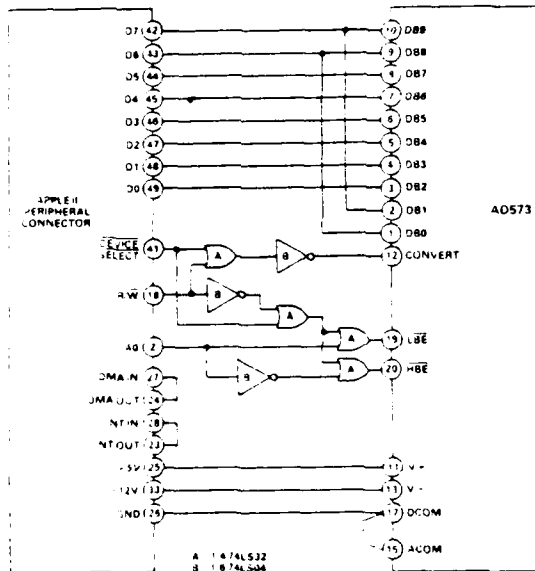


Figure 16. AD573 Interface to Apple II

The BASIC routine listed here will operate the AD573 circuit shown in Figure 16. The conversion is started by POKEing to the location which contains the AD573. The relatively slow execution speed of BASIC eliminates the need for a delay routine between starting and reading the converter. This routine assumes that the AD573 is connected for a ± 5 volt input range. Variable I represents the integer value from 0 to 1023 read from the AD573. Variable V represents the actual value of the input signal in volts.

```
100 PRINT "WHICH SLOT IS THE A/D IN? INPUT S
110 A = 49280 + 16*S
120 POKE A,0
130 L = PEEK A : H = PEEK A + 1
140 I = 4*H + INT L/64
150 V = (I/1024)*10-5
160 PRINT "THE INPUT SIGNAL IS "V" VOLTS."
```

It is also possible to write a faster-executing assembly-language routine to control the AD573. Such a routine will require a

delay between starting and reading the converter. This can be easily implemented by calling the Apple's WAIT subroutine which resides at location 3FCA8, after loading the accumulator with a number greater than or equal to two.

8085-Series Microprocessor Interface

The AD573 can also be used with 8085-series microprocessors. These processors use separate control signals for RD and WR, as opposed to the single R/\bar{W} control signal used in the 6800/6500 series processors.

There are two constraints related to operation of the AD573 with 8085-series processors. The first problem is the width of the CONVERT pulse. The circuit shown in Figure 17, essentially the same as that shown in Figure 13, will produce a wide enough CONVERT pulse when the 8085 is running at 5MHz. For 8085 systems running at slower clock rates (3MHz), the flip-flop-based circuit can be eliminated since the \bar{WR} pulse will be approximately 500ns wide.

The other consideration is the access time of the AD573's three-state output data buffers, which is 250ns maximum. It may be necessary to insert wait states during RD operations from the AD573. This will not be a problem in systems using memories with comparable access times, since wait states will have already been provided in the basic system design.

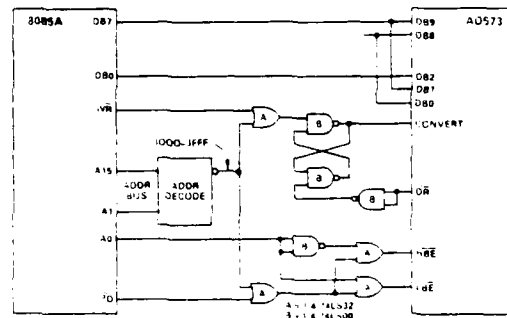


Figure 17. AD573-8085A Interface Connections

The following assembly-language subroutine can be used to control an AD573 residing at memory locations 3000_H and 3001_H. The 10 bits of data are returned left-justified in the DE register pair.

```
ADC: LXI H,3000 ;LOAD HL WITH AD573 ADDRESS
      MOV M,A ;START CONVERSION
      MVI B,06 ;LOAD DELAY PERIOD
      LOOP: DCR B ;DELAY LOOP
            JNZ LOOP
            MOV A,M ;READ LOW BYTE
            ANI 00 ;MASK LOWER 6 BITS
            MOV E,A ;STORE CLEAN LOW BYTE IN E
            INR L ;LOAD HIGH BYTE ADDRESS
            MOV D,M ;MOVE HIGH BYTE TO D
            RET ;EXIT
```


AD544 Operational Amplifier Data Sheet



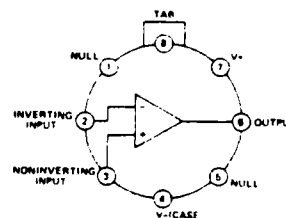
High Speed Implanted FET-Input Op Amp

AD544

FEATURES

- Low Bias Current: 25pA max, warmed-up
- Low Offset Voltage: 500 μ V max
- Low Offset Voltage Drift: 5 μ V/ $^{\circ}$ C max
- Low Input Voltage Noise: 2 μ V p-p
- Low Quiescent Current: 2.5mA max
- High Slew Rate: 13V/ μ s
- Fast Settling to $\pm 0.01\%$: 3 μ s
- Low Total Harmonic Distortion: 0.0015% at 1kHz

AD544 FUNCTIONAL BLOCK DIAGRAM



TO-99
TOP VIEW

PRODUCT DESCRIPTION

The AD544 is a high speed monolithic FET-input operational amplifier fabricated with the most advanced bipolar, JFET and laser trimming technologies. The AD544 offers bias currents significantly lower than currently available monolithic FET-input devices: 25pA max, warmed-up for the AD544K and L, 50pA max for the AD544J. In addition, the offset voltage is laser trimmed to less than 0.5mV on the AD544L and 1.0mV on the AD544K utilizing Analog's laser-wafer-trimming (LWT) process. When combined with the AD544's low offset voltage drift (5 μ V/ $^{\circ}$ C max for "L", 10 μ V/ $^{\circ}$ C max for "K"), these features offer the user IC performance truly superior to existing FET-input op amps—and at low, monolithic pricing.

The key technology required for monolithic JFET-input op amps is the ion-implanted JFET. Ion-implantation (as opposed to diffusion) permits the fabrication of precision, matched JFET's on a monolithic bipolar chip. Analog Devices optimizes the process to produce bias currents lower than other popular FET-input op amps and specifies each device for the maximum value at either input in the fully warmed-up condition. Additional benefits of this optimization include low voltage noise (2 μ V p-p, 0.1–10Hz), and low quiescent current.

The AD544 is recommended for any operational amplifier application requiring excellent ac and dc performance at low cost. The 2MHz bandwidth and low offset of the AD544 make it an excellent choice as an output amplifier for current output D/A Converters such as the AD7541, 12-Bit CMOS DAC. High common mode rejection (80dB, min on the "K" and "L" versions) and open-loop gain ensures better than "12-bit" linearity in high impedance buffer applications.

The AD544 is available in four versions, the "J", "K" and "L" are specified over the 0 to +70 $^{\circ}$ C temperature range and the "S" over the -55 $^{\circ}$ C to +125 $^{\circ}$ C operating temperature range. All devices are packaged in the hermetically-sealed, TO-99 metal can.

PRODUCT HIGHLIGHTS

1. Improved bipolar and JFET processing on the AD544 results in the lowest bias current available in a high speed monolithic FET op amp.
2. Analog Devices, unlike some manufacturers, specifies each device for the maximum bias current at either input in the warmed-up condition, thus assuring the user that the AD544 will meet its published specifications in actual use.
3. Laser-wafer-trimming reduces offset voltage to as low as 0.5mV max (AD544L), thus eliminating the need for external nulling in many situations.
4. If offset nulling is required, the additional offset voltage drift induced will be minimal. (In some devices, offset voltage drift can increase an additional 3 μ V/ $^{\circ}$ C per mV of offset nulled.)
5. Low voltage noise (2 μ V, p-p), and low offset voltage drift (5 μ V/ $^{\circ}$ C) enhance the AD544's performance as a precision op amp.
6. The high slew rate (13.0V/ μ s) and fast settling time to 0.01% (3.0 μ s) make the AD544 ideal for D/A, A/D, sample-and-hold circuits and high speed integrators.
7. Low harmonic distortion (0.0015%) makes the AD544 an ideal choice for audio applications.

AD544 Operational Amplifier Data Sheet (continued)

SPECIFICATIONS ($\alpha + 25^{\circ}\text{C}$ and $V_s = \pm 15\text{V}$ dc)

Model	AD544J			AD544K			AD544L			AD544S			Units
	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
OPEN LOOP GAIN ¹ $V_O = \pm 10\text{V}$, $R_L \geq 2\text{k}\Omega$ T_{min} to T_{max} , $R_L = 2\text{k}\Omega$	30,000 20,000			50,000 40,000			50,000 40,000			50,000 20,000			V/V V/V
OUTPUT CHARACTERISTICS													
Voltage ($R_L = 2\text{k}\Omega$, T_{min} to T_{max})	± 10	± 12		± 10	± 12		± 10	± 12		± 10	± 12		V
Voltage ($R_L = 10\text{k}\Omega$, T_{min} to T_{max})	± 12	± 13		± 12	± 13		± 12	± 13		± 12	± 13		V
Short Circuit Current		25			25			25			25		mA
FREQUENCY RESPONSE													
Unity Gain Small Signal		2.0			2.0			2.0			2.0		MHz
Full-Power Response		200			200			200			200		kHz
Slew Rate, Unity Gain	8.0	13.0		8.0	13.0		8.0	13.0		8.0	13.0		V/ μs
Settling Time to 0.01%		3.0			3.0			3.0			3.0		μs
Total Harmonic Distortion		0.0025			0.0025			0.0025			0.0025		%
INPUT OFFSET VOLTAGE ²													
Initial Offset			2.0			1.0			0.5			1.0	mV
Input Offset Voltage vs. Temp. or T_{min} to T_{max}			20			10			5			15	$\mu\text{V}/^{\circ}\text{C}$
Input Offset Voltage vs. Supply, T_{min} to T_{max}			200			100			100			100	$\mu\text{V}/\text{V}$
INPUT BIAS CURRENT ³													
Either Input		10	50		10	25		10	25		10	25	pA
Offset Current		5			2			2			2		pA
INPUT IMPEDANCE													
Differential ⁴		$10^{12}/6$			$10^{12}/6$			$10^{12}/6$			$10^{12}/6$		M Ω pF
Common Mode		$10^{12}/3$			$10^{12}/3$			$10^{12}/3$			$10^{12}/3$		M Ω pF
INPUT VOLTAGE RANGE													
Differential		± 20			± 20			± 20			± 20		V
Common Mode	± 10	± 12		± 10	± 12		± 10	± 12		± 10	± 12		V
Common Mode Rejection	74			80			80			80			dB
INPUT NOISE													
Voltage 0.1 Hz to 10 Hz		2			2			2			2		$\mu\text{V p-p}$
$f = 10\text{Hz}$		35			35			35			35		nV/ $\sqrt{\text{Hz}}$
$f = 100\text{Hz}$		22			22			22			22		nV/ $\sqrt{\text{Hz}}$
$f = 1\text{kHz}$		18			18			18			18		nV/ $\sqrt{\text{Hz}}$
$f = 10\text{kHz}$		16			16			16			16		nV/ $\sqrt{\text{Hz}}$
POWER SUPPLY													
Rated Performance		± 15			± 15			± 15			± 15		V
Operating	± 5		± 18	± 5		± 18	± 5		± 18	± 5		± 18	V
Quiescent Current		1.8	2.5		1.8	2.5		1.8	2.5		1.8	2.5	mA
TEMPERATURE RANGE													
Operating, Rated Performance	0		+70	0		+70	0		+70	-55		+125	$^{\circ}\text{C}$
Storage	-65		+150	-65		+150	-65		+150	-65		+150	$^{\circ}\text{C}$
PACKAGE ⁵													
TO-99 Style (H08B)	AD544JH			AD544KH			AD544LH			AD544SH			

NOTES

¹Open Loop Gain is specified with V_O both nulled and unnullified.
²Input Offset Voltage specifications are guaranteed after 5 minutes of operation at $T_A = +25^{\circ}\text{C}$.

³Bias Current specifications are guaranteed at maximum at either input after 5 minutes of operation at $T_A = +25^{\circ}\text{C}$. For higher temperatures, the current doubles every 10°C .

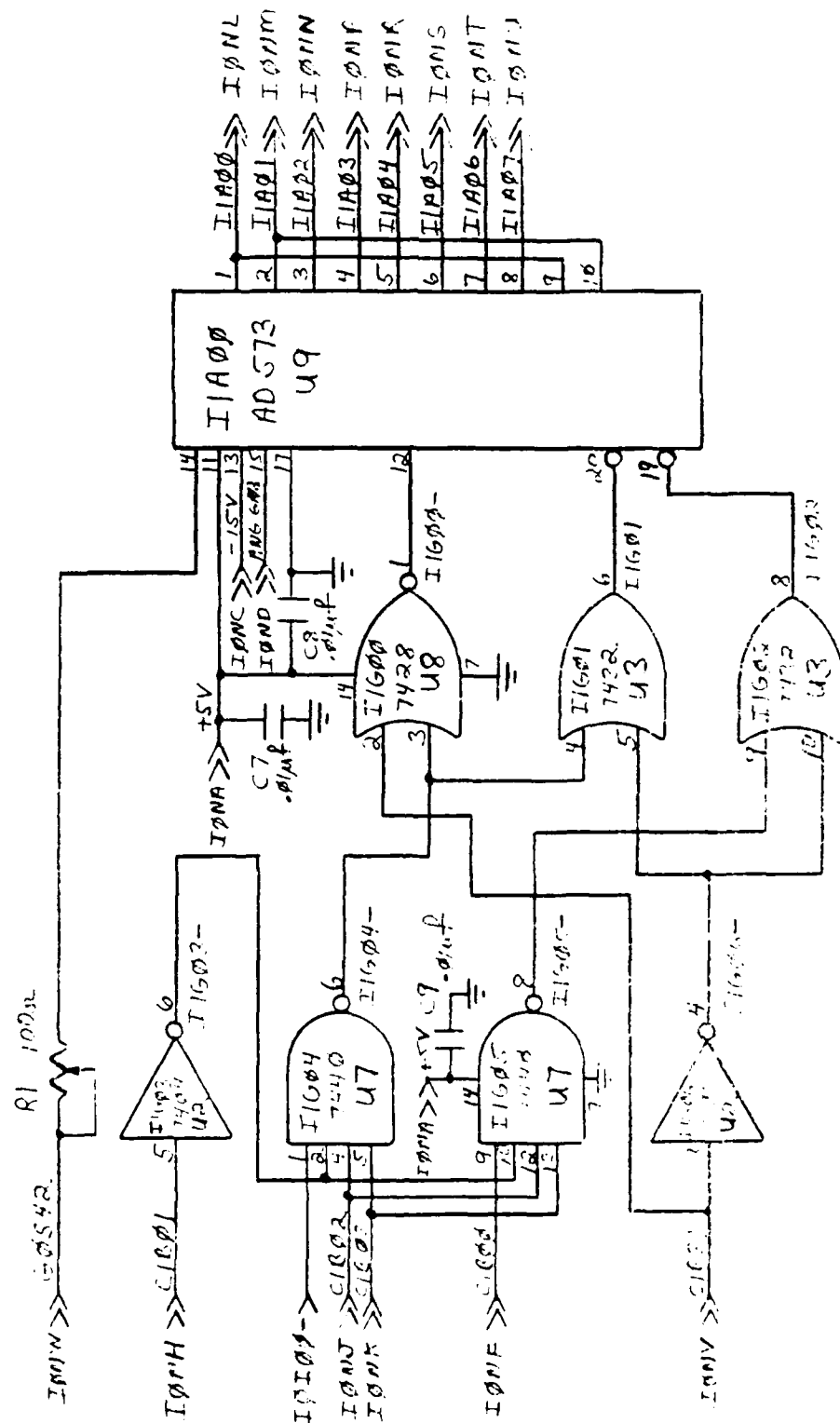
⁴Defined as voltage between inputs, such that neither exceeds $\pm 10\text{V}$ from ground.

⁵See Section 19 for package outline information.
Specifications subject to change without notice.

Specifications shown in boldface are tested on all production units at final electrical test. Results from those tests are used to calculate outgoing quality levels. All min and max specifications are guaranteed, although only those shown in boldface are tested on all production units.

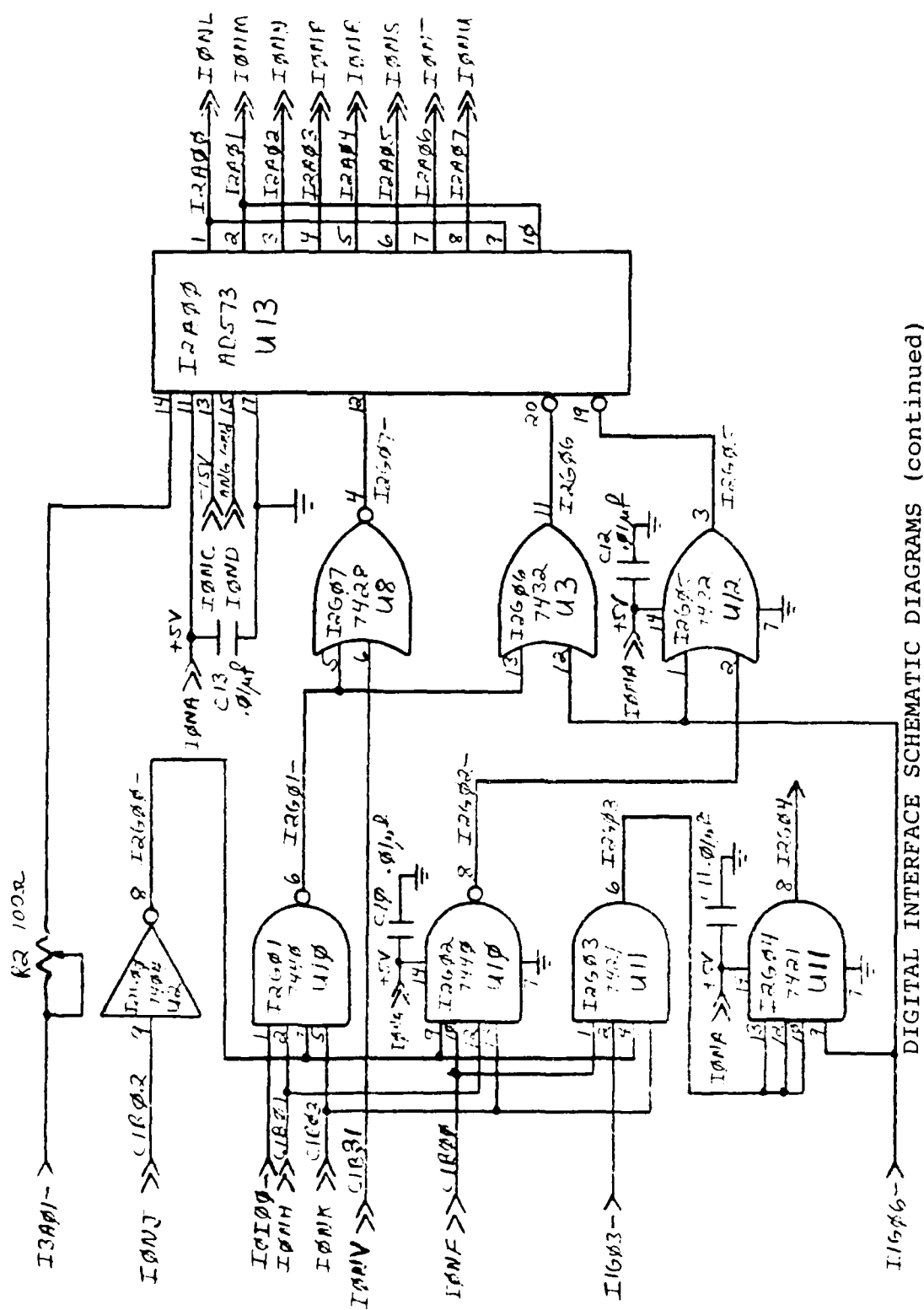
APPENDIX B

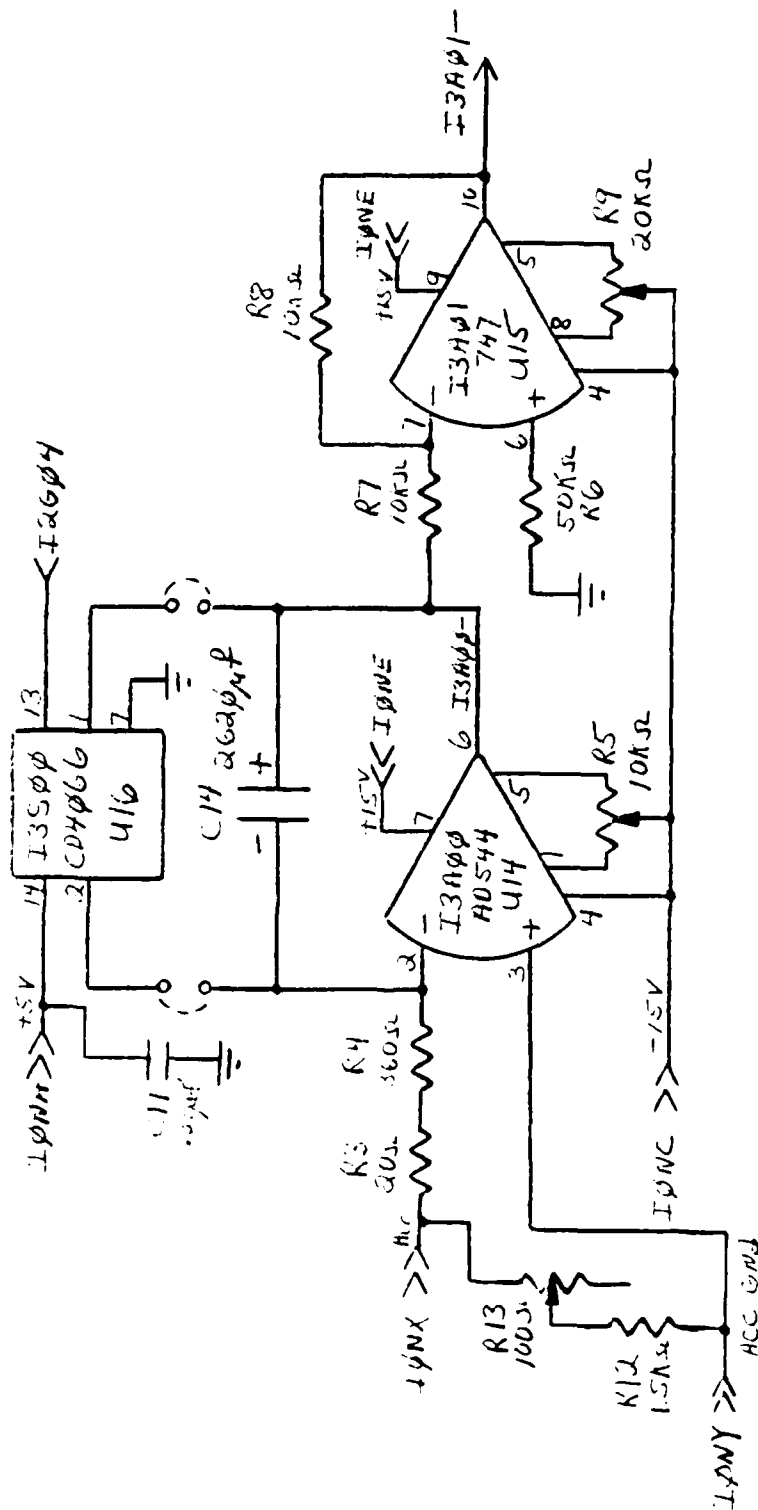
Digital Interface Schematic Diagrams	B-2
Circuit Card I: Digital Interface Parts List	B-7
Digital Interface Device Layout	B-10



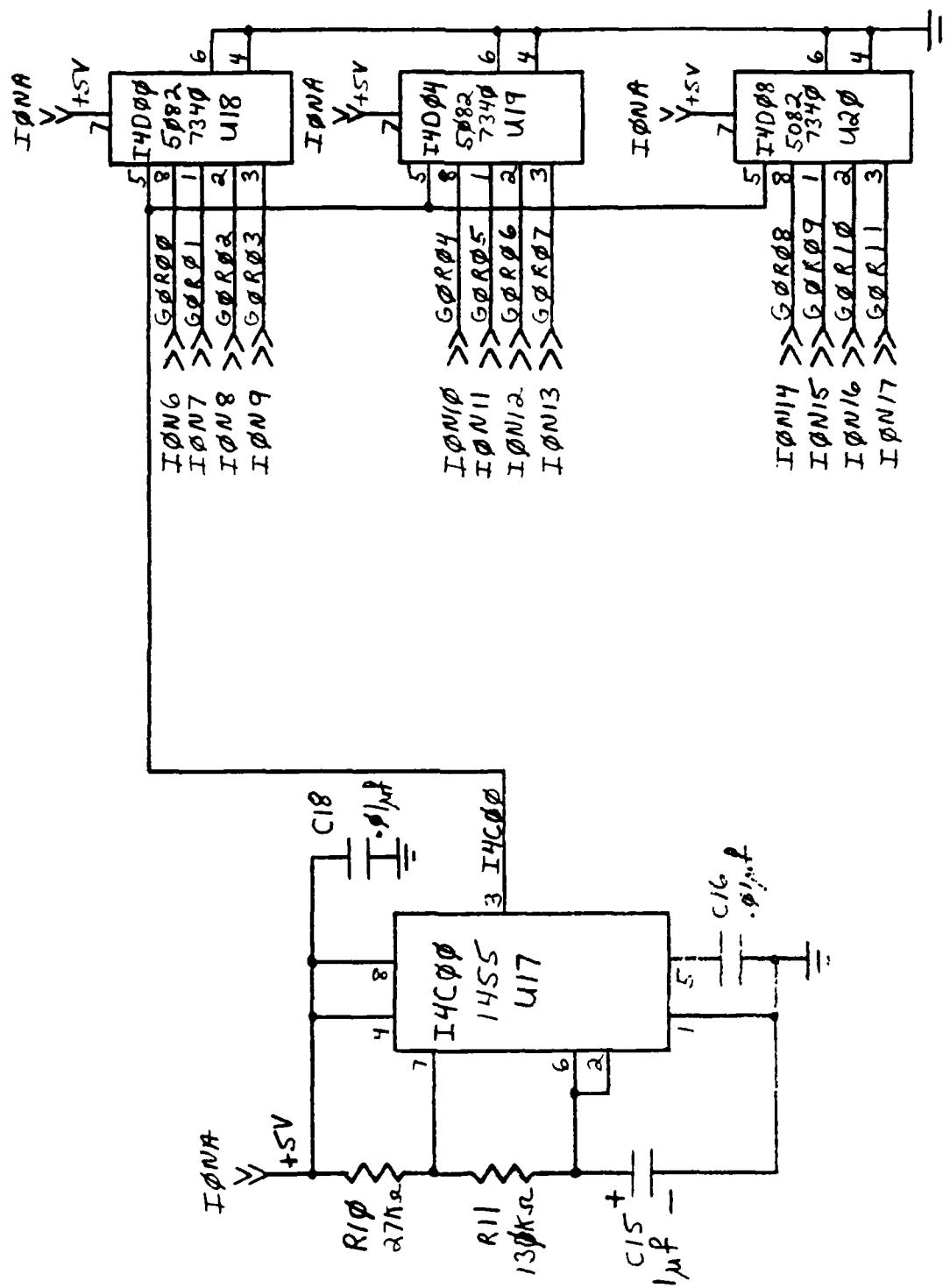
B-3

DIGITAL INTERFACE SCHEMATIC DIAGRAMS (continued)





DIGITAL INTERFACE SCHEMATIC DIAGRAMS (continued)



DIGITAL INTERFACE SCHEMATIC DIAGRAMS (continued)

Circuit Card I: Digital Interface
Parts List

Part Number	Type	Schematic Reference #
7440	Dual 4-Input Positive NAND Buffers	U1 U7 U10
7404	Hex Inverters	U2
7432	Quadruple 2-Input Positive-OR Gates	U3 U12
74173	4-Bit D-Type Register With 3-State Outputs	U4 U5 U6
7428	Quadruple 2-Input Positive-NOR Gates	U8
AD 573	10-Bit Analog to Digital Converter	U9 U13
7421	Dual 4-Input Positive AND Gates	U11
AD 544	Precision Operational Amplifier	U14
LM 747	Dual Operational Amplifiers	U15
CD 4066	Quad Bilateral Switch	U16
1445	Dual Monostable Multivibrator	U17

Circuit Card I: Digital Interface
Parts List (Continued)

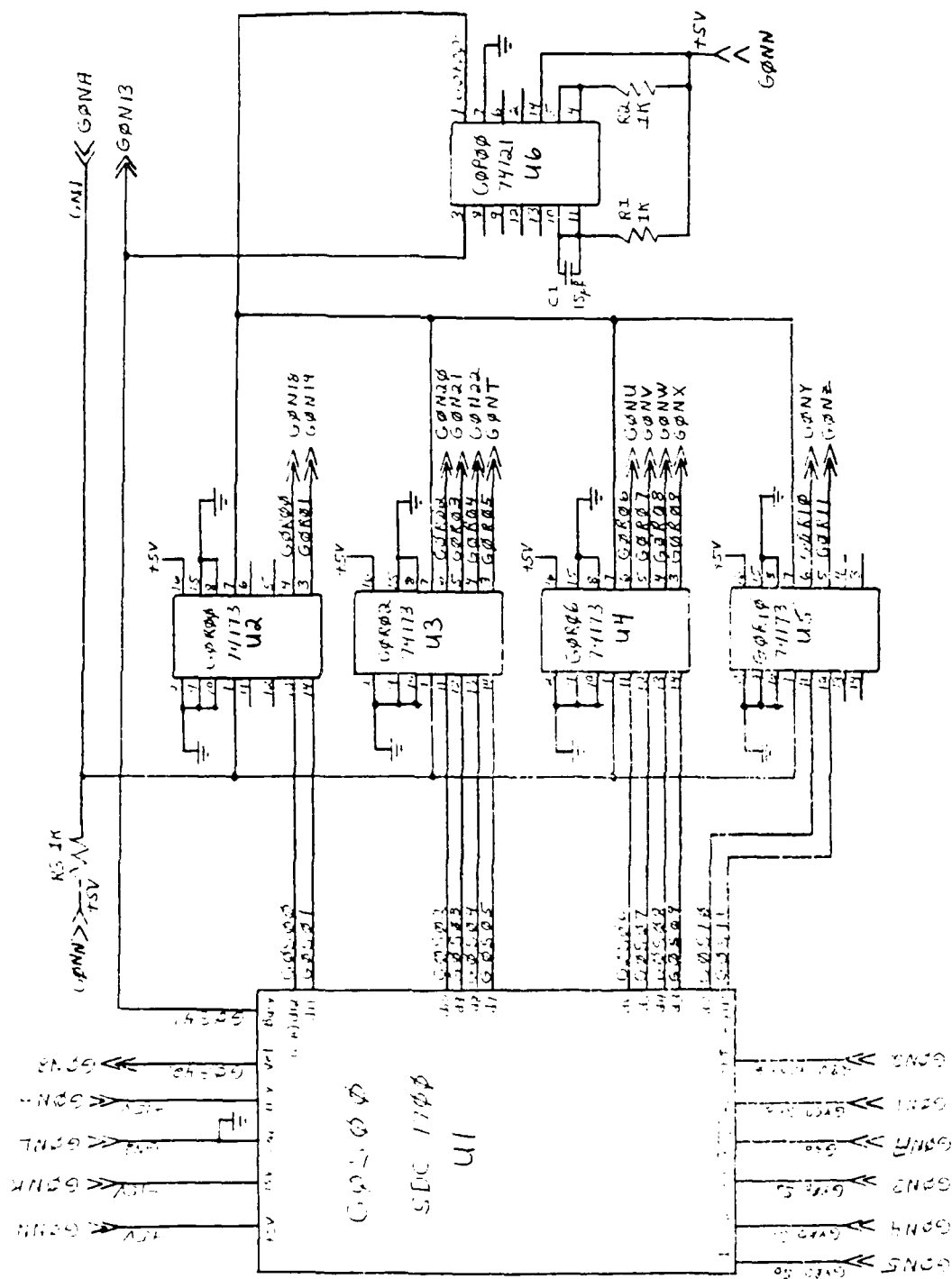
Part Number	Type	Schematic Reference #
5082/7340	Single Digit HEX LED Display with Latches and Driver	U18 U19 U20
100	Variable Resistor (Ohms)	R1 R2 R13
20	Resistor (Ohms)	R3
360	Resistor (Ohms)	R4
10K	Variable Resistor (Ohms)	R5
50K	Resistor (Ohms)	R6
10K	Resistor (Ohms)	R7 R8
20K	Variable Resistor (Ohms)	R9
27K	Resistor (Ohms)	R10
130K	Resistor (Ohms)	R11
1.5K	Resistor (Ohms)	R12

Circuit Card I: Digital Interface
Parts List (Continued)

Part Number	Type	Schematic Reference #
0.01 Micro	Capacitor (Farads)	C1
		C2
		C3
		C4
		C5
		C6
		C7
		C8
		C9
		C10
		C11
		C12
		C13
		C16
		C17
		C18
2220 Micro	Capacitor (Farads)	C14
1.0	Capacitor (Farads)	C15

APPENDIX C

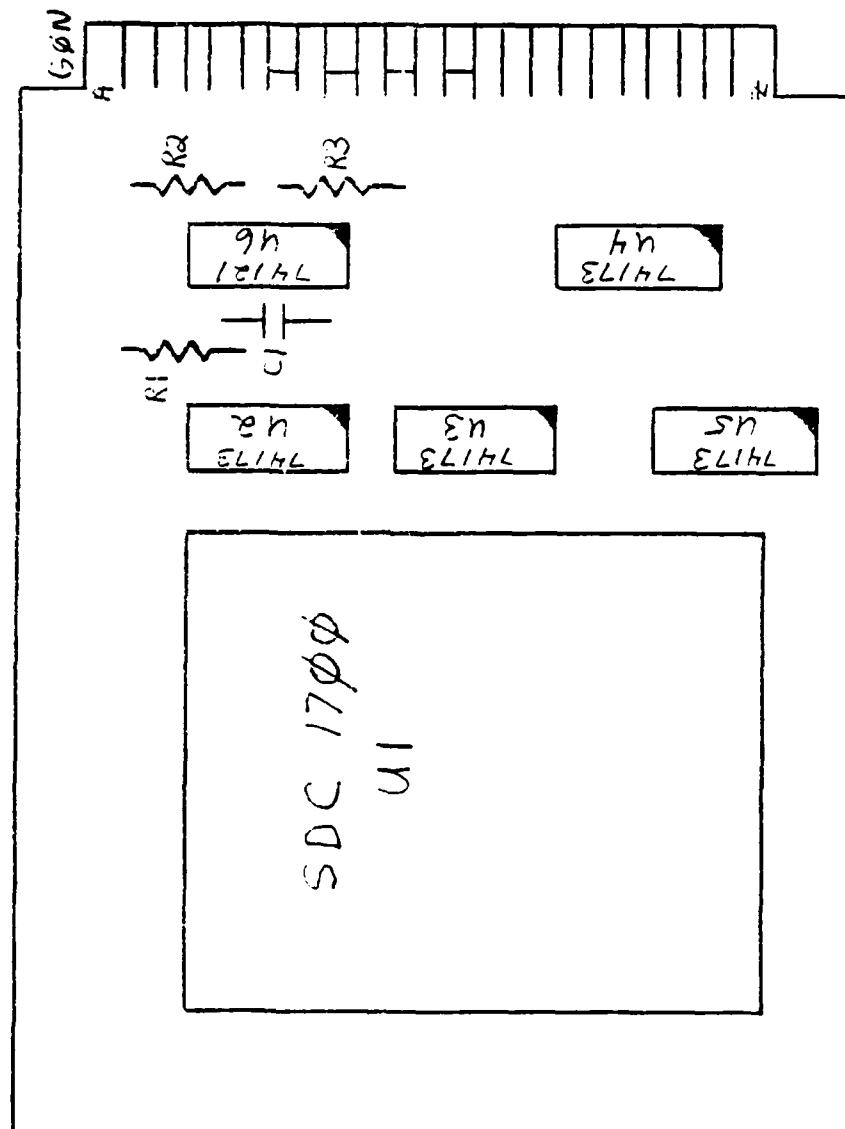
Syncro to Digital Schematic Diagram	C-2
Circuit Card G: Syncro to Digital Parts List	C-3
Syncro to Digital Device Layout	C-4



SYNCRO TO DIGITAL SCHEMATIC DIAGRAM

Circuit Card G: Syncro to Digital
Parts List

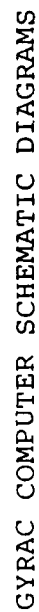
Part Number	Type	Schematic Reference #
SDC 1700	Syncro to Digital Converter	U1
74173	4-Bit D-Type Register With 3-State Outputs	U2 U3 U4 U5
74121	Monostable Multivibrator	U6
1K	Resistor (Ohms)	R1 R2 R3
15 pico	Capacitor (Farads)	C1

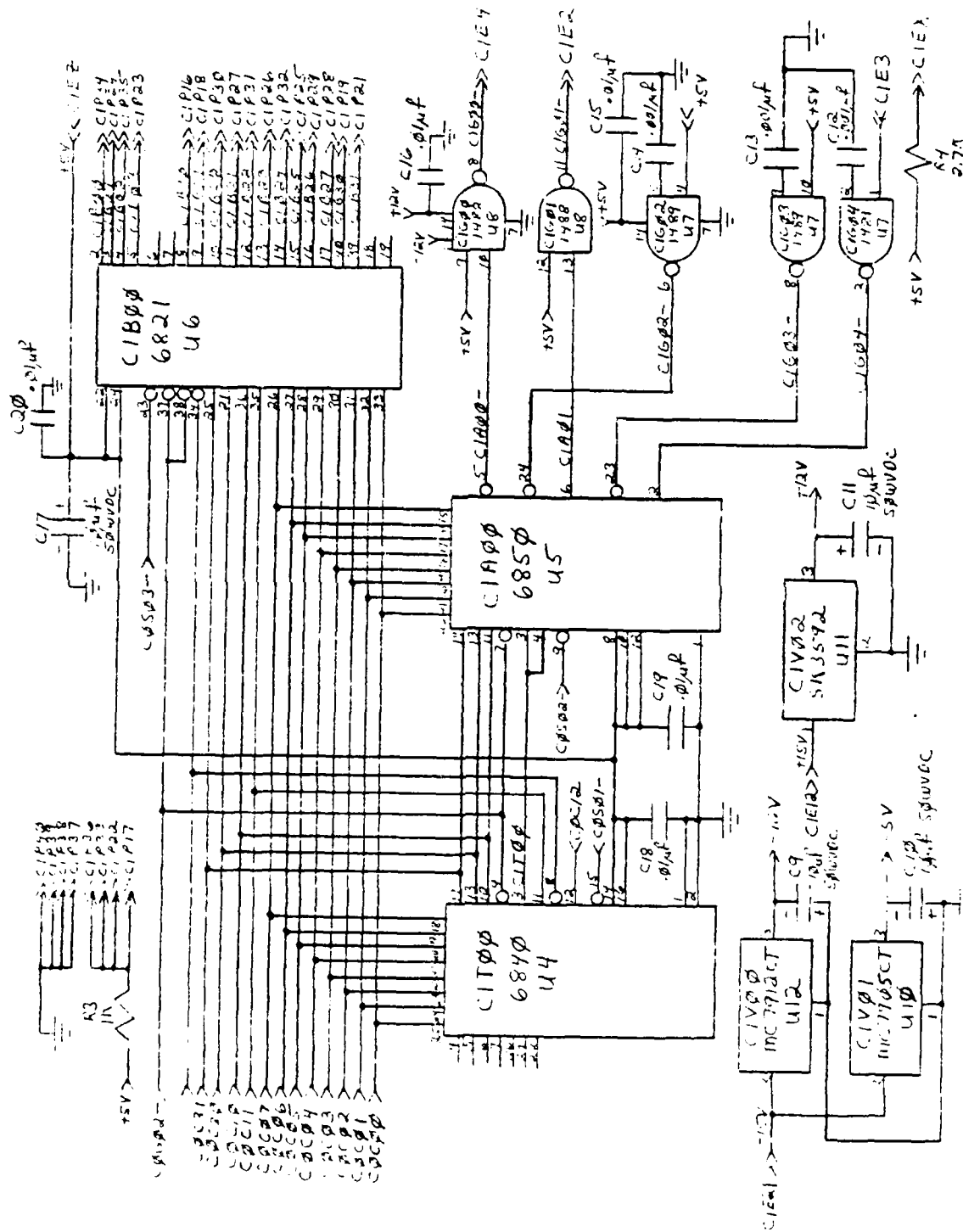


SYNCHRO TO DIGITAL DEVICE LAYOUT

APPENDIX D

GYRAC Computer Schematic Diagrams	D-2
Circuit Card C: Computer Controller Parts List	D-4
GYRAC Computer Memory Map	D-7
GYRAC Computer Device Layout	D-9





GYRAC COMPUTER SCHEMATIC DIAGRAMS (continued)

Circuit Card C: Computer Controller Parts List

Part Number	Type	Schematic Reference #
MC6802	Microprocessor with Clock and 128 Bytes RAM (CPU)	U1
74138	3-8 Line Decoders- Multiplexers	U2
2708	1024 x 8 bit U.V. Erasable PROM	U3
MC6840	Programmable Timer Module (PTM)	U4
MC6850	Asynchronous Communications Interface Adapter (ACIA)	U5
MC6821	Peripheral Interface Adapter (PIA)	U6
1489	Quad Line Receiver	U7
1488	Quad Line Driver	U8
7414	Hex Schmitt-Trigger Inverters	U9
MC7905CT	Negative 5 Volt Voltage Regulator	U10
SK3592	Positive 12 Volt Voltage Regulator	U11

Circuit Card C: Computer Controller
Parts List (Continued)

Part Number	Type	Schematic Reference #
MC7912CT	Negative 12 Volt Voltage Regulator	U12
MB8416A	CMOS 2048 x 8 Byte Static RAM	U13
1N914	Signal Diode	D1
3579.545 KC	Crystal Oscillator	X1
3K	Resistor (Ohms)	R1
10K	Resistor (Ohms)	R2
1K	Resistor (Ohms)	R3
2.7K	Resistor (Ohms)	R4
22 Pico	Capacitor (Farads)	C1 C2
0.01 Micro	Capacitor (Farads)	C3 C4 C5 C6 C7 C15 C16 C18 C19 C20
100 Micro	Capacitor (Farads) 50 WVDC	C8

Circuit Card C: Computer Controller
Parts List (Continued)

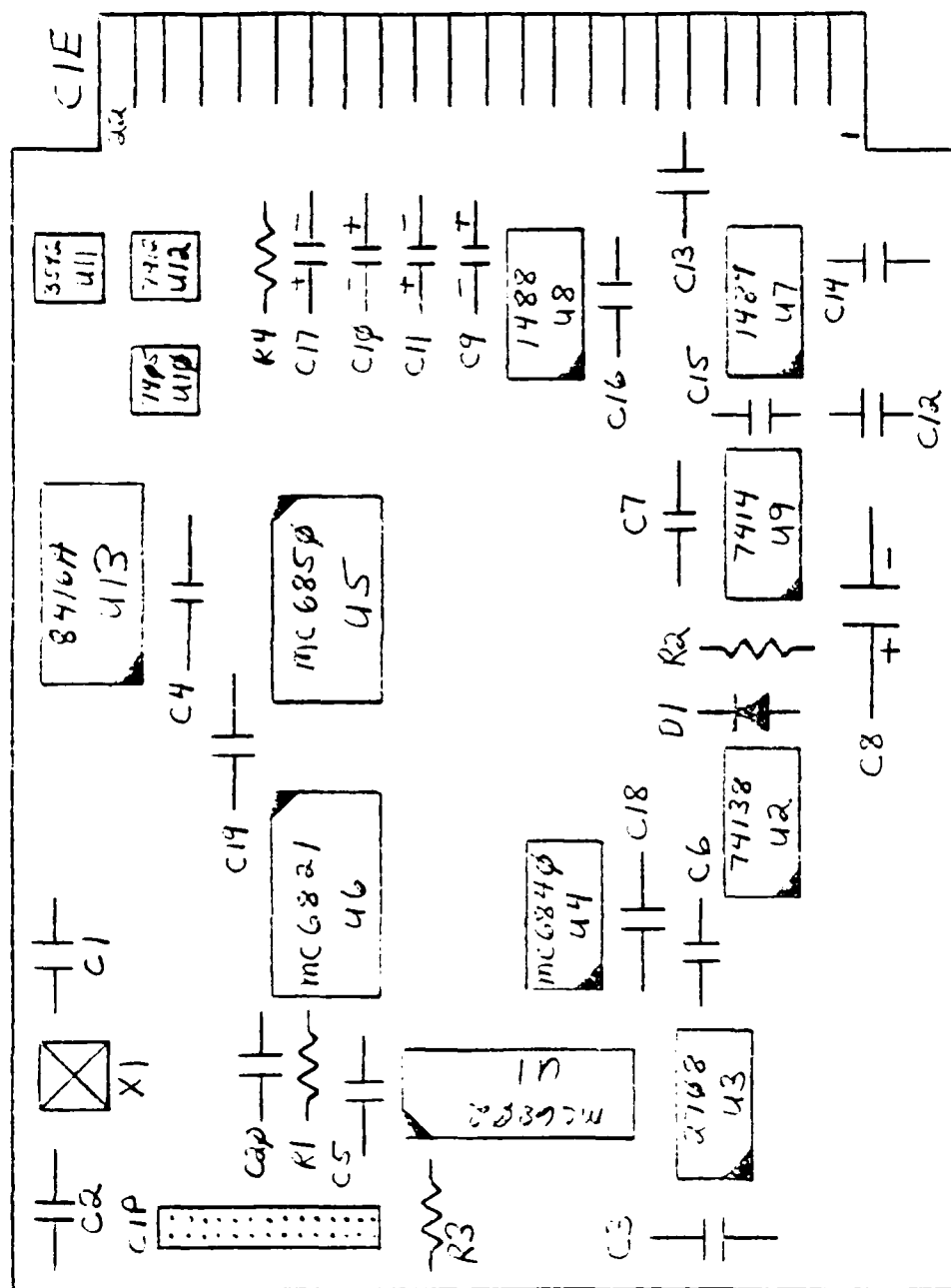
Part Number	Type	Schematic Reference #
10 Micro	Capacitor (Farads)	C9
	50 WVDC	C10
		C11
		C17
0.001 Micro	Capacitor (Farads)	C12
		C13
		C14

GYRAC COMPUTER MEMORY MAP

ADDRESS (HEX)	DEVICE
0000 to 007F	0128 Bytes of on Processor Scratchpad RAM
6000 to 67FF	2048 Bytes of RAM (Alternate Addresses 7000-77FF)
8000	PIA Data Direction/Peripheral Register A (Alternate Address 9000)
8001	PIA Control Register A (Alternate Address 9001)
8002	PIA Data Direction/Peripheral Register B (Alternate Address 9002)
8003	PIA Control Register B (Alternate Address 9003)
A000	ACIA Control/Status Register (Alternate Address B000)
A001	ACIA TX Data/RX Data Register (Alternate Address B001)
C000	PTM Write Control Register #3/#1 (Alternate Address D000)
C001	PTM Write Control #2/Status Registers (Alternate Address D001)
C002	PTM MSB Buffer #1 Register/Timer #1 Counter (Alternate Address D002)
C003	PTM Timer #1 Latches/LSB Buffer #1 Register (Alternate Address D003)
C004	PTM MSB Buffer #2 Register/Timer #2 Counter (Alternate Address D004)
C005	PTM timer #2 Latches/LSB Buffer #2 Register (Alternate Address D005)

GYRAC COMPUTER MEMORY MAP (continued)

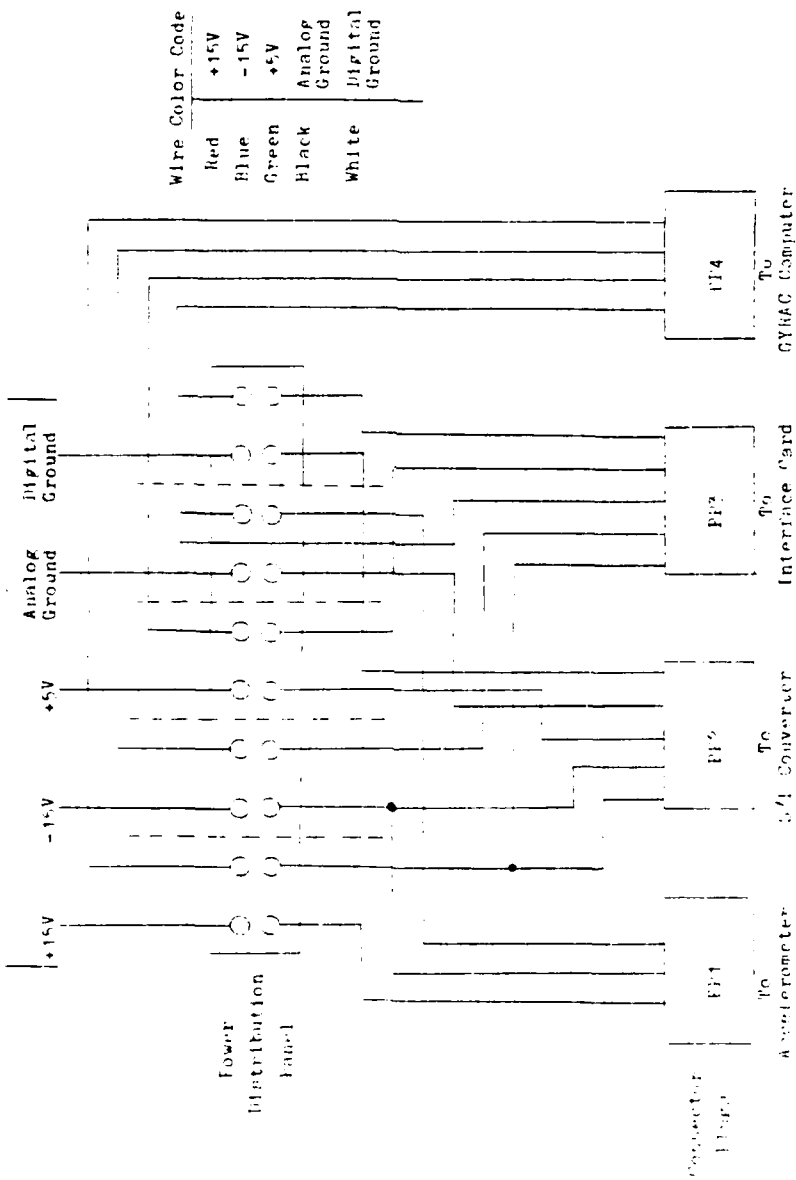
ADDRESS (HEX)	DEVICE
C0060	PTM MSB Buffer #3 Register/Timer #3 Counter (Alternate Address D006)
C007	PTM Timer #3 Latches/LSB Buffer #3 Register (Alternate Address D007)
E000 to E3FF	1024 Bytes of GYRAC Control Program in EPROM (Alternate Addresses F000-F3FF)
E3FE to E3FF	Address of Reset Vector
E3FC to E3FD	Address of Non-maskable Interrupt Vector
E3FA to E3FB	Address of Software Interrupt Vector
E3F8 to E3F9	Address of Interrupt Vector



APPENDIX E

GYRAC Power Panel	E-2
GON: Syncro to Digital Edge Connector	E-3
C1P: GYRAC Computer Sensor Bus Connector	E-4
C1E: GYRAC Computer Edge Connector	E-5
ION: Digital Interface Edge Connector	E-6
H89 to GYRAC RS-232 Cable	E-7
H89 to NAV T or Drive Computer RS-232 Cable	E-8
GYRAC to NAV L Computer RS-232 Cable	E-9
Drive Computer to NAV X RS-232 Cable	E-10
GYRAC Wiring Harness	E-11

From Gyro Base Assembly (except +5V is from External Supply)



GYRAC POWER PANEL

GON: Syncro to Digital Edge Connector

Pin Number	Signal Name	Description
GON1	None	GYRO Base Ground
GON2	None	GYRO Base 28 Volts 400 Hertz
GON3	None	GYRO Indicator S2
GON4	None	GYRO Indicator S1
GON5	None	GYRO Indicator S0
GON6		Unused
GON8	GOS42	Angular Velocity Analog Signal
GON13	GOS41	S to D Busy Signal (unused)
GON15		Unused
GON16		Unused
GON17		Unused
GON18	GOR00	D0 Data Bit
GON19	GOR01	D1 Data Bit
GON20	GOR02	D2 Data Bit
GON21	GOR03	D3 Data Bit
GON22	GOR04	D4 Data Bit
GONA	None	Digital Ground
GONF	None	+15 Volts DC
GONH	None	+15 Volts DC
GONJ	None	-15 Volts DC
GONK	None	-15 Volts DC
GONL	None	Analog Ground
GONM	None	Analog Ground
GONN	None	+5 Volts DC
GONP	None	+5 Volts DC
GONT	GOR05	D5 Data Bit
GONU	GOR06	D6 Data Bit
GONV	GOR07	D7 Data Bit
GONW	GOR08	D8 Data Bit
GONX	GOR09	D9 Data Bit
GONY	GOR10	D10 Data Bit
GONZ	GOR11	D11 Data Bit

C1P: GYRAC Computer Sensor Bus Connector

Pin Number	Signal Name	Description
C1P1		Unused
C1P2		Unused
C1P3		Unused
C1P4		Unused
C1P5		Unused
C1P6		Unused
C1P7		Unused
C1P8		Unused
C1P9		Unused
C1P10		Unused
C1P11		Unused
C1P12		Unused
C1P13		Unused
C1P14		Unused
C1P15		Unused
C1P16	C1B10	Reserved (unused)
C1P17	None	+5 Volts (thru pullup resistor)
C1P18	C1B11	Reserved (unused)
C1P19	C1B30	Interrupt In (unused)
C1P20		Unused
C1P21	C1B31	Read/Write Out
C1P22	None	+5 Volts (thru pullup resistor)
C1P23	C1B03	A3 Sensor Address Bit
C1P24	C1B01	A1 Sensor Address Bit
C1P25	C1B25	D5 Sensor Data Bus
C1P26	C1B23	D3 Sensor Data Bus
C1P27	C1B21	D1 Sensor Data Bus
C1P28	C1B27	D7 Sensor Data Bus
C1P29	C1B26	D6 Sensor Data Bus
C1P30	C1B20	D0 Sensor Data Bus
C1P31	C1B22	D2 Sensor Data Bus
C1P32	C1B24	D4 Sensor Data Bus
C1P33	None	+5 Volts (thru pullup resistor)
C1P34	C1B00	A0 Sensor Address Bit
C1P35	C1B02	A2 Sensor Address Bit
C1P36	None	+5 Volts (thru pullup resistor)
C1P37	None	Digital Ground
C1P38	None	Digital Ground
C1P39	None	Digital Ground
C1P40	None	Digital Ground

C1E: GYRAC Computer Edge Connector

Pin Number	Signal Name	Description
C1E1	None	Digital Ground
C1E2	C1G01-	Tx Data Out (active low)
C1E3	None	Rx Data In (active low)
C1E4	C1G00-	Ready to Send
C1E5	None	+5 Volts DC Out
C1E6		Unused
C1E7		Unused
C1E8	None	+5 Volts DC Out
C1E9		Unused
C1E10	None	Power On Reset (active low)
C1E11		Unused
C1E12	None	+15 Volts DC In
C1E13		Unused
C1E14	None	-12 Volts DC Out
C1E15		Unused
C1E16	None	+12 Volts DC Out
C1E17		Unused
C1E18	None	-5 Volts DC Out
C1E19		Unused
C1E20		Unused
C1E21	None	-15 Volts DC In
C1E22		Unused
C1EA	None	Digital Ground
C1EB		Unused
C1EC		Unused
C1ED		Unused
C1EE		Unused
C1EF		Unused
C1EH	None	Unused
C1EJ		Unused
C1EK		Unused
C1EL		Unused
C1EM		Unused
C1EN		Unused
C1EP		Unused
C1ER		Unused
C1ES		Unused
C1ET		Unused
C1EU		Unused
C1EV		Unused
C1EW		Unused
C1EX	None	+5 Volts (thru pullup resistor)
C1EY		Unused
C1EZ	None	+5 Volts DC In

ION: Digital Interface Edge Connector

Pin Number	Signal Name	Description
ION1		Unused
ION2		Unused
ION3		Unused
ION4		Unused
ION5		Unused
ION6	GOR00	S/D D0 Data Bit
ION7	GOR01	S/D D1 Data Bit
ION8	GOR02	S/D D2 Data Bit
ION9	GOR03	S/D D3 Data Bit
ION10	GOR04	S/D D4 Data Bit
ION11	GOR05	S/D D5 Data Bit
ION12	GOR06	S/D D6 Data Bit
ION13	GOR07	S/D D7 Data Bit
ION14	GOR08	S/D D8 Data Bit
ION15	GOR09	S/D D9 Data Bit
ION16	GOR10	S/D D10 Data Bit
ION17	GOR11	S/D D11 Data Bit
ION18		Reserved
ION19		Reserved
ION20		Reserved
ION21		Reserved
ION22		Reserved
IONA	None	+5 Volts DC
IONB	None	Digital Ground
IONC	None	-15 Volts DC
IOND	None	Analog Ground
IONE	None	+15 Volts DC
IONF	C1B00	A0 Address Bit
IONH	C1B01	A1 Address Bit
IONJ	C1B02	A2 Address Bit
IONK	C1B03	A3 Address Bit
IONL	Bus	D0 Data Bit
IONM	Bus	D1 Data Bit
IONN	Bus	D2 Data Bit
IONP	Bus	D3 Data Bit
IONR	Bus	D4 Data Bit
IONS	Bus	D5 Data Bit
IONT	Bus	D6 Data Bit
IONU	Bus	D7 Data Bit
IONV	C1B31	Read/Write Control Line
IONW	GOS42	Angular Velocity Analog Signal
IONX	None	Acceleration Analog Signal
IONY	None	Acceleration Ground
IONZ		Unused

H89 to GYRAC RS-232 Cable

89 Connector Pin Number	Signal Description	GYRAC Connector Pin Number
2	Tx Data	2
3	Rx Data	3
7	Ground	7

H89 to Nav T or Drive Computer RS-232 Cable

Nav T/Drive Computer		
H89 Connector Pin Number	Signal Description	Connector Pin Number
2	Tx Data	3
3	Rx Data	2
7	Ground	7

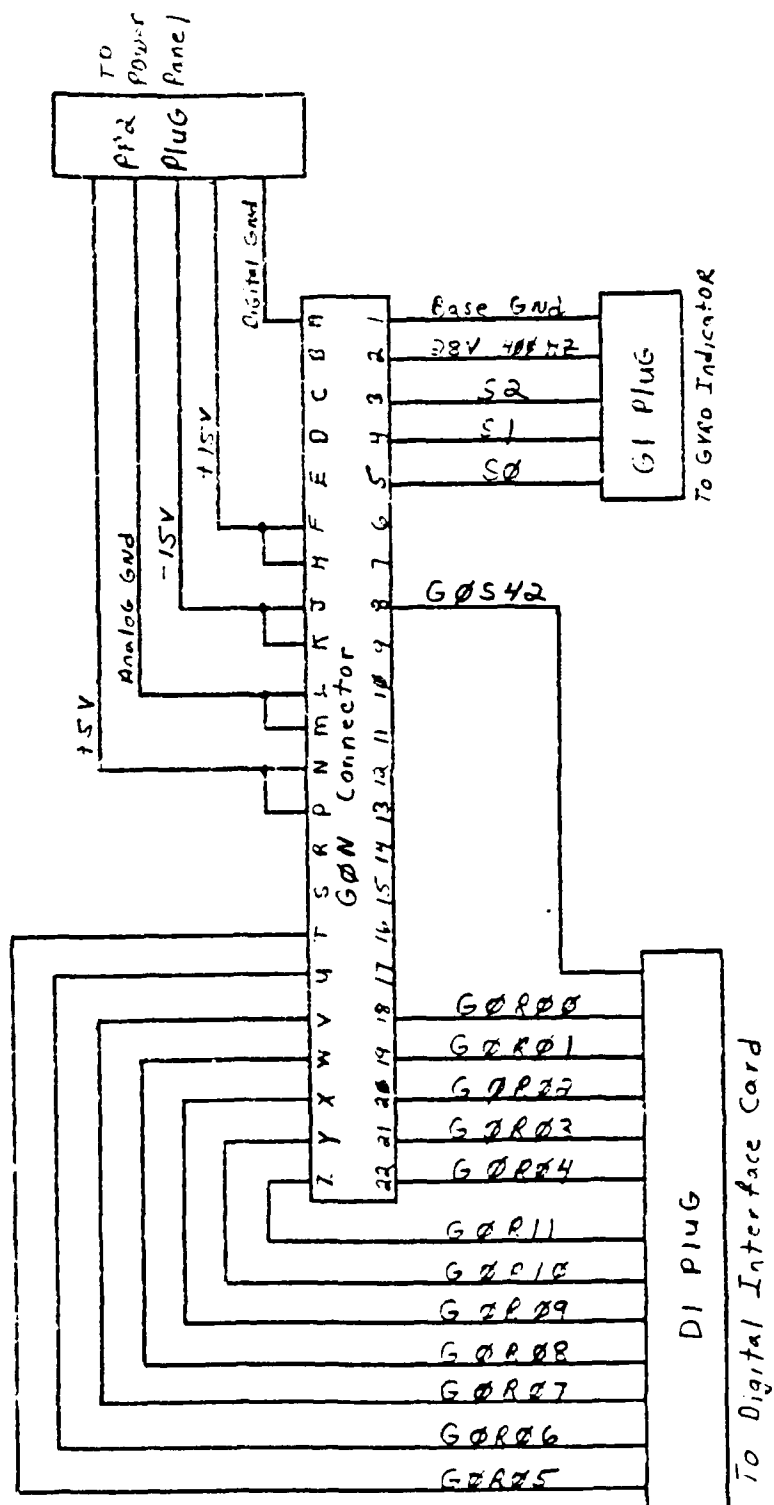
GYRAC to Nav L Computer RS-232 Cable

GYRAC Connector Pin Number	Signal Description	Nav L Connector Pin Number
2	Tx Data	3
3	Rx Data	2
7	Ground	7

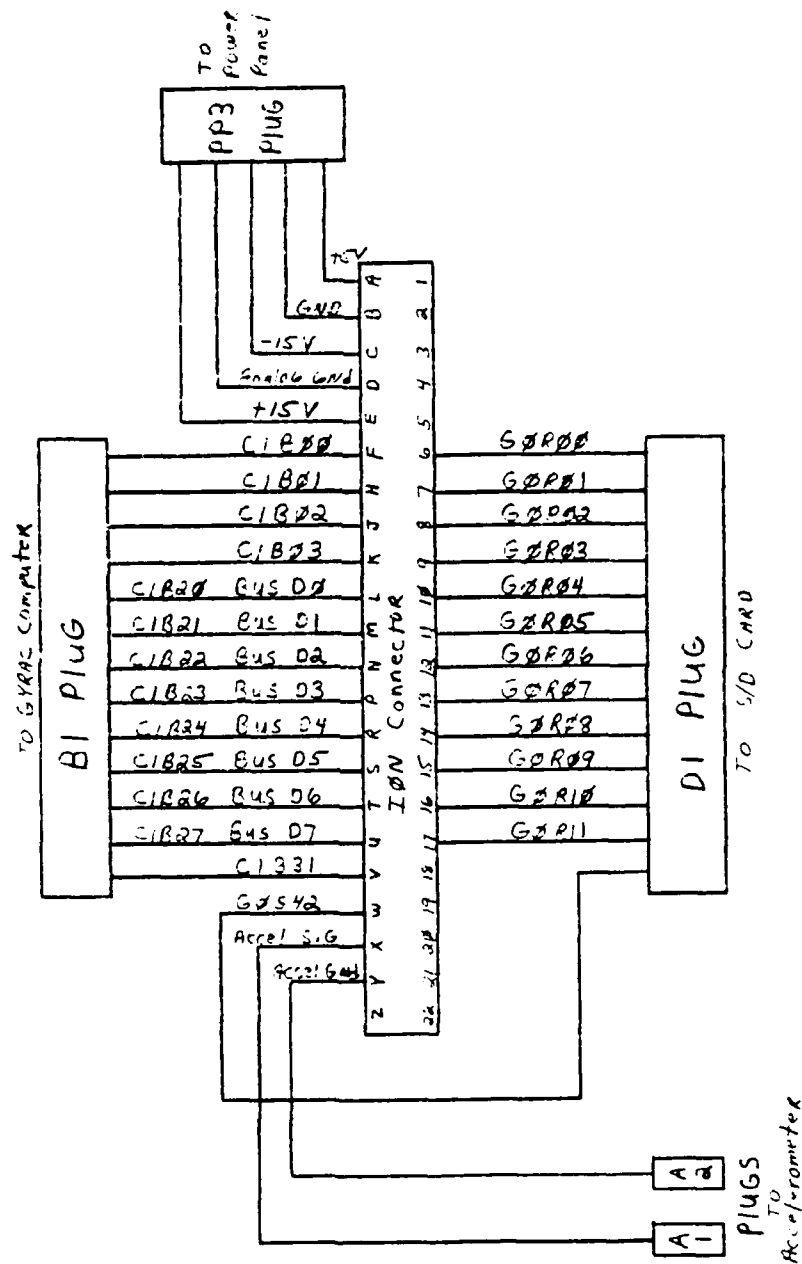
Drive Computer to Nav X RS-232 Cable

Drive Computer Connector Pin Number	Signal Description	Nav X Connector Pin Number
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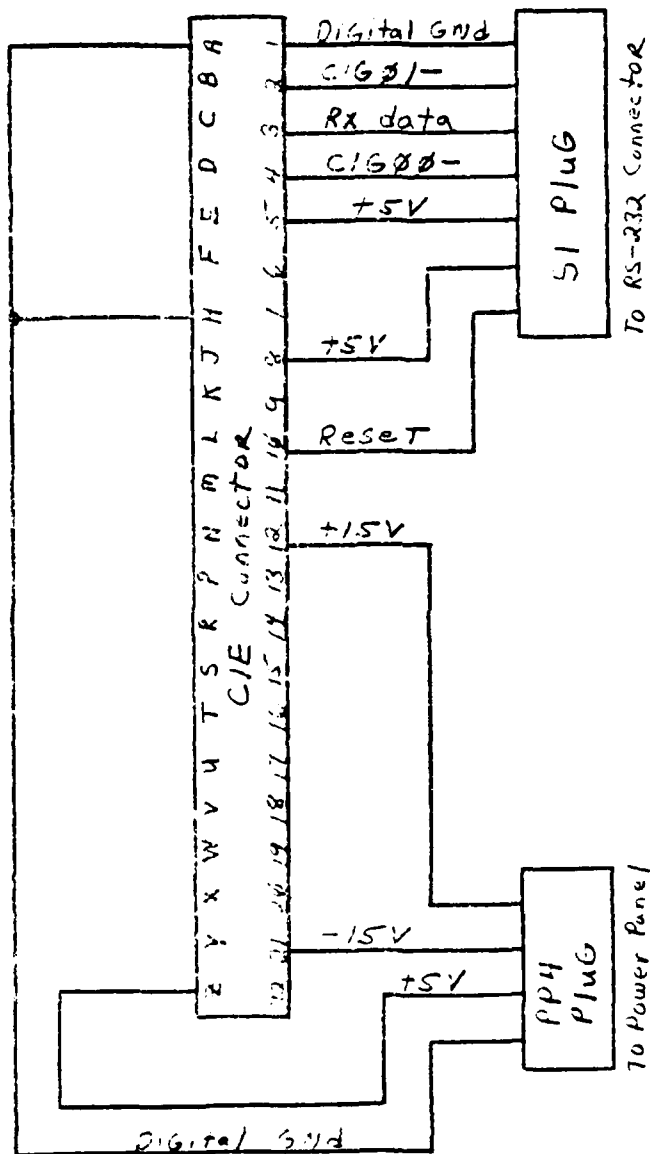
2	Tx Data	3
3	Rx Data	2
7	Ground	7



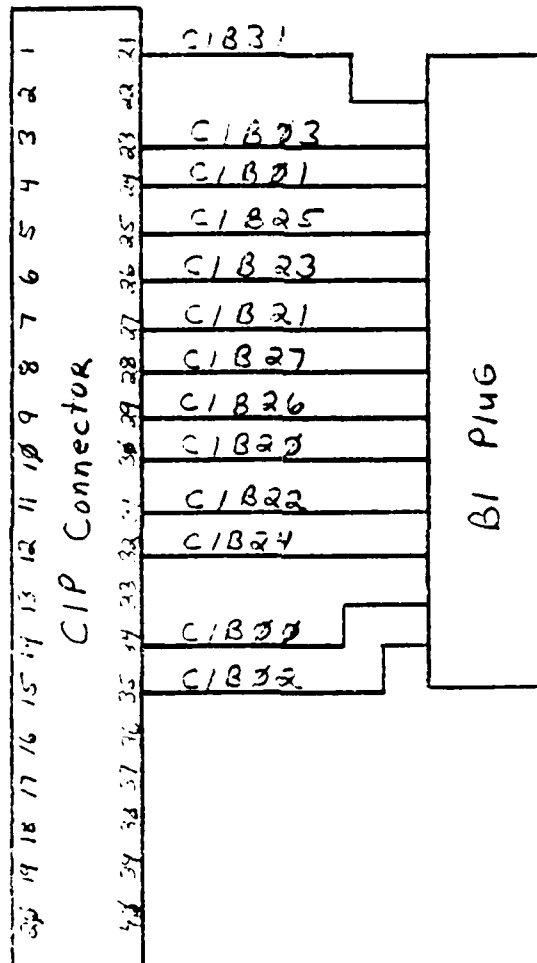
GYRAC WIRING HARNESS



GYRAC WIRING HARNESS (continued)



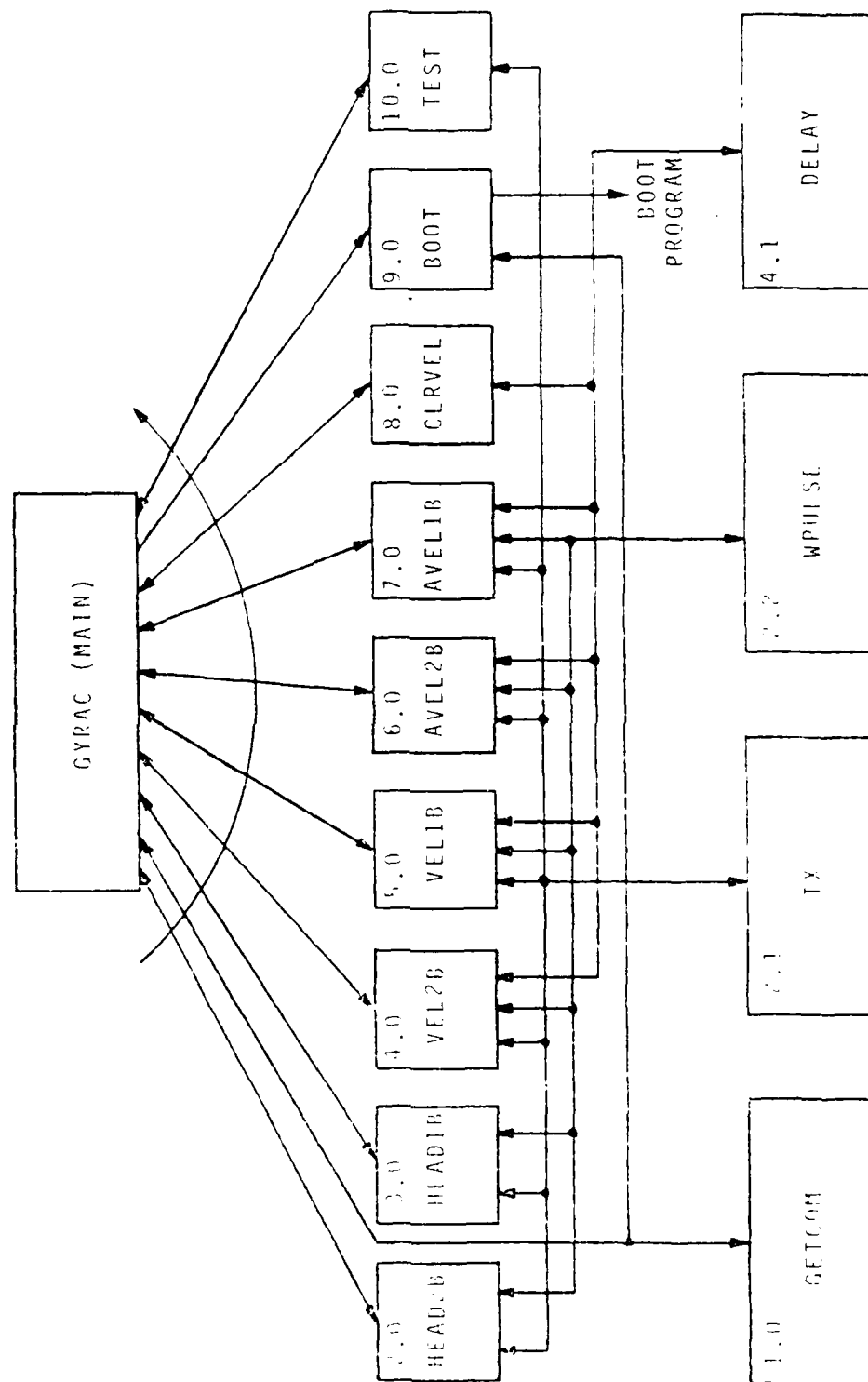
GYRAC WIRING HARNESS (continued)



GYRAC WIRING HARNESS (continued)

APPENDIX F

GYRAC.A Structure Chart	F-2
GYRAC.A Program Listing	F-3
GYRAC.A Operating Instructions	F-21



GYRAC.A STRUCTURE CHART

GYRAC.A PROGRAM LISTING

[illegible][illegible]

GYBAC.A PROGRAM LISTING (continued)

0	pinio	equ	10000H	Pinio module I/O address for pinio
0001	status	equ	10004H	Pinio status register
0002	read	equ	10008H	Pinio read port address
0003	write	equ	1000CH	Pinio write port address
0004	otwired	equ	10010H	OTW1 way to be written, use enable pinio
0005				General OT mode, continuous operating mode, output enabled
0006	pscount	equ	40000H	Pinio divide count to 5 = (2000000-1)/5
0007	addrreset	equ	100H	Pinio address reset
0008	addradd	equ	104H	Pinio address add
0009				Divide by 10 mode, 8 bits data plus 1 stop bit, all interrupts disabled
0010	txreg	equ	010H	Pinio transmit data register empty
0011	rxreg	equ	014H	Pinio receive data register full
0012	headaddl	equ	018H	Address of GRAC heading low byte
0013	headaddh	equ	01CH	Address of GRAC heading high byte
0014	veladdl	equ	020H	Address of GRAC velocity low byte
0015	veladdh	equ	024H	Address of GRAC velocity high byte
0016	avevel1	equ	028H	Address of GRAC avg vel low byte
0017	avevelh	equ	02CH	Address of GRAC avg vel high byte
0018	velclear	equ	030H	Address of GRAC vel clear device
0019	temp	equ	034H	Address of RAM scratch pad
0020	rambot	equ	038H	Address of bottom of 2 ram block
0021	blockb	equ	03CH	Address of number of blocks of data
0022	bytes	equ	040H	Address of number of bytes in a block
0023	char	equ	044H	Address of current ASCII character
0024	charcount	equ	048H	Address of current character count
0025	delay	equ	050H	Timing delay loop variable

Address	Hex	Asm	Comment
0000		begin	Beginning of main program
0001	0F	init	Disable interrupts
0002	0E 07 FF	lis	#stack
			Initialize stack pointer to top of RAM
0004	4F	liara	Put zero in A reg
0005	17 00 01	staa	Clear AIA control register A
0006	07 00 01	staa	Clear AIA control register B
0007	17 00 01	staa	Make AIA port B an input port
0008	07	liara	
0009	07 00 00	staa	Make AIA port A an output port
0010	07 00	liara	#read
0011	07 00 01	staa	Prepare port A for data out & read
0012	07 00 00	staa	Prepare port B for data input
0013	07 00 01	staa	
0014	07 00 01	staa	
0015	07 00 00	staa	
0016	07 00 00	staa	
0017	07 00 00	staa	
0018	07 00 00	staa	
0019	07 00 00	staa	
0020	07 00 00	staa	
0021	07 00 00	staa	
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0040	07 00 00	staa	
0041	07 00 00	staa	
0042	07 00 00	staa	
0043	07 00 00	staa	
0044	07 00 00	staa	
0045	07 00 00	staa	
0046	07 00 00	staa	
0047	07 00 00	staa	
0048	07 00 00	staa	
0049	07 00 00	staa	
0050	07 00 00	staa	
0051	07 00 00	staa	
0052	07 00 00	staa	
0053	07 00 00	staa	
0054	07 00 00	staa	
0055	07 00 00	staa	
0056	07 00 00	staa	
0057	07 00 00	staa	
0058	07 00 00	staa	
0059	07 00 00	staa	
0060	07 00 00	staa	
0061	07 00 00	staa	
0062	07 00 00	staa	
0063	07 00 00	staa	
0064	07 00 00	staa	
0065	07 00 00	staa	
0066	07 00 00	staa	
0067	07 00 00	staa	
0068	07 00 00	staa	
0069	07 00 00	staa	
0070	07 00 00	staa	
0071	07 00 00	staa	
0072	07 00 00	staa	
0073	07 00 00	staa	
0074	07 00 00	staa	
0075	07 00 00	staa	
0076	07 00 00	staa	
0077	07 00 00	staa	
0078	07 00 00	staa	
0079	07 00 00	staa	
0080	07 00 00	staa	
0081	07 00 00	staa	
0082	07 00 00	staa	
0083	07 00 00	staa	
0084	07 00 00	staa	
0085	07 00 00	staa	
0086	07 00 00	staa	
0087	07 00 00	staa	
0088	07 00 00	staa	
0089	07 00 00	staa	
0090	07 00 00	staa	
0091	0		

GYRAC.A PROGRAM LISTING (continued)

00000000	00	40	0000	#0000	:Put command 0 in B register
00000001	00	41	0001	#0001	:Is command 0 in A register?
00000002	00	42	0002	#0002	:If yes go execute command 2
00000003	00	43	0003	#0003	:Put command 3 in B register
00000004	00	44	0004	#0004	:Is command 3 in A register?
00000005	00	45	0005	#0005	:If yes go execute command 5
00000006	00	46	0006	#0006	:Put command 6 in B register
00000007	00	47	0007	#0007	:Is command 6 in A register?
00000008	00	48	0008	#0008	:If yes go execute command 8
00000009	00	49	0009	#0009	:Put command 9 in B register
0000000A	00	4A	000A	#000A	:Is command 9 in A register?
0000000B	00	4B	000B	#000B	:If yes go execute command B
0000000C	00	4C	000C	#000C	:Put command C in B register
0000000D	00	4D	000D	#000D	:Is command C in A register?
0000000E	00	4E	000E	#000E	:If yes go execute command E
0000000F	00	4F	000F	#000F	:Put command F in B register
00000010	00	50	0010	#0010	:Is command F in A register?
00000011	00	51	0011	#0011	:If yes go execute command F
00000012	00	52	0012	#0012	:Put command 0 in B register
00000013	00	53	0013	#0013	:Is command 0 in A register?
00000014	00	54	0014	#0014	:If yes go execute command 0
00000015	00	55	0015	#0015	:Put command 1 in B register
00000016	00	56	0016	#0016	:Is command 1 in A register?
00000017	00	57	0017	#0017	:If yes go execute command 1
00000018	00	58	0018	#0018	:Put command 2 in B register
00000019	00	59	0019	#0019	:Is command 2 in A register?
0000001A	00	5A	001A	#001A	:If yes go execute command 2
0000001B	00	5B	001B	#001B	:Put command 3 in B register
0000001C	00	5C	001C	#001C	:Is command 3 in A register?
0000001D	00	5D	001D	#001D	:If yes go execute command 3
0000001E	00	5E	001E	#001E	:Put command 4 in B register
0000001F	00	5F	001F	#001F	:Is command 4 in A register?
00000020	00	60	0020	#0020	:If yes go execute command 4
00000021	00	61	0021	#0021	:Put command 5 in B register
00000022	00	62	0022	#0022	:Is command 5 in A register?
00000023	00	63	0023	#0023	:If yes go execute command 5
00000024	00	64	0024	#0024	:Put command 6 in B register
00000025	00	65	0025	#0025	:Is command 6 in A register?
00000026	00	66	0026	#0026	:If yes go execute command 6
00000027	00	67	0027	#0027	:Put command 7 in B register
00000028	00	68	0028	#0028	:Is command 7 in A register?
00000029	00	69	0029	#0029	:If yes go execute command 7
0000002A	00	6A	002A	#002A	:Put command 8 in B register
0000002B	00	6B	002B	#002B	:Is command 8 in A register?
0000002C	00	6C	002C	#002C	:If yes go execute command 8
0000002D	00	6D	002D	#002D	:Put command 9 in B register
0000002E	00	6E	002E	#002E	:Is command 9 in A register?
0000002F	00	6F	002F	#002F	:If yes go execute command 9
00000030	00	70	0030	#0030	:Put command 0 in B register
00000031	00	71	0031	#0031	:Is command 0 in A register?
00000032	00	72	0032	#0032	:If yes go execute command 0
00000033	00	73	0033	#0033	:Put command 1 in B register
00000034	00	74	0034	#0034	:Is command 1 in A register?
00000035	00	75	0035	#0035	:If yes go execute command 1
00000036	00	76	0036	#0036	:Put command 2 in B register
00000037	00	77	0037	#0037	:Is command 2 in A register?
00000038	00	78	0038	#0038	:If yes go execute command 2
00000039	00	79	0039	#0039	:Put command 3 in B register
0000003A	00	7A			

GYRAC.A PROGRAM LISTING (continued)

```

B017 10 B1 17      jsr    avel1b    ;Read and T: 1 byte of ang vel
B018 7E B1 17      jmp     input     ;Get next command

B01E 00 B0 FA cmdD jsr    head1b    ;Read and T: 2 bytes of heading
B01F 7E B0 FA      jmp     input     ;Get next command

B044 10 B1 11 cmdD jsr    head1b    ;Read and T: 1 byte of heading
B045 7E B0 10      jmp     input     ;Get next command

B04A 10 B1 41 cmdE jsr    vel1b     ;Read and T: 2 bytes of velocity
B04B 7E B0 36      jmp     input     ;Get next command

B010 00 B1 74 cmdF jsr    vel1b     ;Read and T: 1 byte of velocity
B011 7E B0 36      jmp     input     ;Get next command

B056 00 B1 10 cmdG jsr    dirvel    ;Clear velocity constant
B057 7E B0 36      jmp     input     ;Get next command

B011 00 B1 36 cmdH jsr    avel2b    ;Read and T: 2 bytes of ang vel
B012 7E B0 36      jmp     input     ;Get next command

B002 00 B1 09 cmdI jsr    avel1b    ;Read and T: 1 byte of ang vel
B003 7E B0 16      jmp     input     ;Get next command

B013 00 B0 FA cmdJ jsr    head2b    ;Read and T: 2 bytes of heading
B014 10 B1 41      jsr    vel1b     ;Read and T: 2 bytes of velocity
B015 7E B0 36      jmp     input     ;Get next command

B001 00 B1 19 cmdK jsr    head1b    ;Read and T: 1 byte of heading
B004 10 B1 74      jsr    vel1b     ;Read and T: 1 byte of velocity
B007 7E B0 10      jmp     input     ;Get next command

B004 7E B0 06 cmdL jmp     init      ;Reset and re-initialize GYRAC

B000 7E B1 E2 cmdM jmp     boot      ;Load Raw heading from RS-232 & execute

B1E0 00 B1 60 cmdN jsr    test      ;If printable char to RS-232
B1E1 7E B0 31      jmp     input     ;Get next command

B0E1 00 B1 10 cmdO jsr    head1b    ;Read and T: 1 byte of heading
B0E2 10 B1 41      jsr    vel1b     ;Read and T: 2 bytes of velocity
B0E3 7E B0 10      jmp     input     ;Get next command

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 5/10/85
VERSION: 1.0
NAME: getcom
MODULE NUMBER: 1.0
DESCRIPTION: This module reads an Asci: byte from the ACIA
             and places it in the A register. It will loop
             indefinitely until a byte is read.
RAISED VARIABLES: None
RETURN: Asci: command byte in the A register
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAME JR.
HISTORY: NONE

```

```

BDEF 04 A0 00 getcom ldaa aciastat ;Get ACIA (PS-232) status
BDEF 04 01 andra ;Is receive data register full
BDEF 07 00 beq getcom ;If no loop to getcom
BDEF 05 A0 01 ldaa aciadata ;Else read command byte from ACIA
BDEF 07 01 rts

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 05/15/82
VERSION: 1.1
NAME: head2b
MODULE NUMBER: 204
DESCRIPTION: This module sends a write pulse to the gyro heading data register to latch the data, reads two bytes of heading data, and sends the data, high byte first, to the ACIA (53-132) output. Only the least significant 12 bits of the data is significant.
PASSED VARIABLES: None
RETURNS: Two bytes of heading data written to ACIA (53-132)
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: none
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: Version 1.0 original
        Version 1.1 change heading data format

```

```

E0FA 36 0F      head2b  ldaa  #headaddl ;Get address of LSB of heading
E0FC 87 30 00   staa  piadataA ;Write LSB heading address to GYRAC
E0FE 80 ED 23   jsr   wpulse  ;Send GYRAC write pulse to latch data
E100 85 0E      ldaa  #headaddh ;Get address of MSB of heading
E102 87 30 00   staa  piadataA ;Write MSB heading address to GYRAC
E104 85 18 02   ldsb  piadataB ;Read MSB of heading from GYRAC
E106 80 ED 1F   jsr   tx      ;Send MSB of heading to ACIA
E108 36 0F      ldaa  #headaddl ;Get address of LSB of heading
E10A 87 30 00   staa  piadataA ;Write LSB heading address to GYRAC
E10C 85 18 02   ldsb  piadataB ;Read LSB of heading from GYRAC
E10E 80 ED 1F   jsr   tx      ;Send LSB of heading to ACIA
E110 39         rts

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 05/14/85
VERSION: 1.0
NAME: 0A
MODULE NUMBER: 0.1
DESCRIPTION: This module tests the status of the ACIA for an
              empty transmit data register in an indefinite
              loop until found and then the data in the B
              register is sent to the ACIA RS-232 output.
PARM VARIABLES: Input data byte in the B register
RETURN: Data byte output to ACIA
LOCAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: readit
                  readb
                  v110
                  v115
                  ave10
                  ave11
AUTHOR: CAPT WILLIAM J. SAMEY JR.
HISTORY: NONE

```

```

E010 06 00 00 00 11aa acrststat ;Read ACIA (RS-232) status
E012 04 02      anda *done      ;Is ACIA transmit data register empty?
E014 07 00      beq  *n         ;If no, loop to 04 and recheck status
E016 07 00 01    stab acldata    ;Else transmit data in B reg to ACIA
E022 00

```


GYRAC.A PROGRAM LISTING (continued)

```

DATE: 10/10/85
VERSION: 1.1
NAME: wpulse
MODULE NUMBER: 0.1
DESCRIPTION: This module sends a write pulse to the GYRAC.
PULSED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: 0
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: head0b
                  head1b
                  head2b
                  head3b
                  head4b
                  head5b
AUTHOR: CAPT WILLIAM H. RAMEY JR.
HISTORY: NONE

```

```

8223 06 14 wpulse 1dab #write :Get GYRAC write command
8225 F7 30 01 stab piactr1A :Send GYRAC write level to latch data
8228 06 10 1dab #read :Get GYRAC read command
8229 F7 30 01 stab piactr1A :Send GYRAC read level to form a pulse
8230 39

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 05/16/85
VERSION: 1.1
NAME: headb
MODULE NUMBER: 0.0
DESCRIPTION: This module sends a write pulse to the gyro
             heading data register to latch the data, reads
             the bytes of heading data, adjusts the data to
             form a byte of the 8 most significant bits, and
             sends this byte of heading data to the ACIA
             (RS-232) output.
RAISED VARIABLES: None
RETURN: One byte of heading data written to ACIA (RS-232)
GLOBAL VARIABLES USED: temp
GLOBAL VARIABLES CHANGED: temp
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: tpr
             upulse
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: Version 1.0 original
         Version 1.1 change heading data format

```

```

E119 56 0F headb ldaa #headadd1 ;Get address of LSB of heading
E11B 87 80 00 staa piadataA ;Write address of LSB of head to GYRAC
E11E 8D E2 03 jsr upulse ;Send GYRAC write pulse to latch data
E121 56 0E ldaa #headaddh ;Get address of MSB of heading
E123 87 80 00 staa piadataA ;Write address of MSB of head to GYRAC
E126 85 80 02 ldab piadataB ;Read MSB of heading data
E129 58 aslb ;Move lower 4 bits to upper 4 bits
E12A 58 aslb ;and shift 4 bytes into the lower
E12B 58 aslb ;4 bits--i.e. 00000000. This will
E12C 58 aslb ;form the MS nibble of the heading byte
E12E 56 0F ldaa #headaddl ;Get address of LSB of heading
E12F 87 80 00 staa piadataA ;Write address of LSB of head to GYRAC
E132 85 80 02 ldab piadataB ;Read LSB of heading data
E135 44 lraa ;Move upper 4 bits to lower 4 bits
E136 44 lraa ;and shift 4 bytes into the upper
E137 44 lraa ;4 bits--i.e. 00000000. This will
E138 44 lraa ;form the LS nibble of the heading byte
E139 87 00 staa temp ;Save the LS nibble of the heading
E13B 8A 00 orab temp ;Form 8 bits of heading (1 byte)
E13D 8D E2 13 jsr tpr ;Output 1 byte of heading to ACIA
E140 39 rts

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 04/01/97
VERSION: 1.1
NAME: velli
MODULE NUMBER: 1.0
DESCRIPTION: This module sends a write pulse to the gyro
            vel[0], data register. After the data, reads
            the type and velocity data, generates and sends
            the data, MID first, to the ACIA RS-232C output.
            Only the least significant 10 bits of the data is
            significant.
PAUSED VARIABLES: None
RETURNS: Two bytes of velocity data written to ACIA (RS-232C)
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTER USED: A, C
FILE READ: None
FILE WRITTEN: None
MODULES CALLED: 0
                pause
                delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. SAMEY JR.
HISTORY: Version 1.0 Original
         Version 1.1 change velocity data format

```

E141	38	0A	vel12b	ldaa	#vel1addr	:Get address of MSB of Velocity
E142	B7	80	00	staa	piadataA	:Write MSB Velocity address to GYFAC
E143	00	02	00	ldsr	pulse	:Send GYFAC write pulse to latch data
E149	00	01	02	ldsr	delay	:Allow for A to D conversion time
E140	F7	30	02	ldac	piadataB	:Read MSB of heading from GYFAC
E14F	07	00		stab	temp	:Save raw MSB of data
E151	94			lsrb		:Form correct MSB
E151	94			lsrb		
E152	94			lsrb		
E154	94			lsrb		
E155	94			lsrb		
E156	94			lsrb		
E157	00	02	10	ldsr	h	:Send MSB of heading to ADIC
E15A	80	00		ldaa	#vel1addr	:Get address of LSB of velocity
E15B	B7	80	00	staa	piadataA	:Write LSB Velocity address to GYFAC
E15C	00	00	02	ldac	piadataB	:Read LSB of velocity from GYFAC
E15E	94			lsrb		:Adjust raw LSB to correct format
E15F	94			lsrb		
E160	94			lsrb		
E161	94			lsrb		
E162	94			lsrb		
E163	94			lsrb		
E164	94			lsrb		
E165	94			lsrb		
E166	94			lsrb		
E167	94			lsrb		
E168	94			lsrb		
E169	94			lsrb		
E16A	94			lsrb		
E16B	94			lsrb		
E16C	94			lsrb		
E16D	94			lsrb		
E16E	94			lsrb		
E16F	94			lsrb		
E170	94			lsrb		
E171	94			lsrb		
E172	94			lsrb		
E173	94			lsrb		
E174	94			lsrb		
E175	94			lsrb		
E176	94			lsrb		
E177	94			lsrb		
E178	94			lsrb		
E179	94			lsrb		
E17A	94			lsrb		
E17B	94			lsrb		
E17C	94			lsrb		
E17D	94			lsrb		
E17E	94			lsrb		
E17F	94			lsrb		
E180	94			lsrb		
E181	94			lsrb		
E182	94			lsrb		
E183	94			lsrb		
E184	94			lsrb		
E185	94			lsrb		
E186	94			lsrb		
E187	94			lsrb		
E188	94			lsrb		
E189	94			lsrb		
E18A	94			lsrb		
E18B	94			lsrb		
E18C	94			lsrb		
E18D	94			lsrb		
E18E	94			lsrb		
E18F	94			lsrb		
E190	94			lsrb		
E191	94			lsrb		
E192	94			lsrb		
E193	94			lsrb		
E194	94			lsrb		
E195	94			lsrb		
E196	94			lsrb		
E197	94			lsrb		
E198	94			lsrb		

GYRAC.A PROGRAM LISTING (continued)

```
DATE: 06-08-97  
VERSION: 1.1  
NAME: delay  
MODULE NUMBER: 4.1  
COMPILED: /usr/src/kernels/2.0.10es_4.0/mk_delay + 1.1173.1/7 +  
             file: 0.0 + 10.0 seconds.  
PAUSED VARIABLES: None  
RETURN: None  
LOCAL VARIABLES USED: None  
GLOBAL VARIABLES CHANGED: None  
REGISTERED: A  
FILES READ: None  
FILES WRITTEN: None  
MODULES CALLED: None  
CALLING MODULES: void  
                void  
                void  
                void  
                void  
AUTHOR: CAPT WILLIAM J. RAMEY JR.  
HISTORY: NONE
```

```

E21E 36 0F delay ldaa #del      ;Get time delay variable
E21F 01 next      dup
E220 44      decb
E221 08 00 bne next      ;if not done, do another loop delay
E222 03      rts

```

1
 E2F3 org vector
 E2F4 80 00 00 00 fdb begin, begin, begin, begin ;1st 16 vector table
 E2F5 80 00 00 00

E400 END

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 01-12-87
VERSION: 1.1
NAME: velib
MODULE NUMBER: 5.0
DESCRIPTION: This module sends a write pulse to the Gyrac
             velocity data register to latch the data, reads
             the MSB of heading data, and sends this byte of
             velocity data to the ACIA (RS-232) output.
PASSED VARIABLES: None
RETURNS: One byte of velocity data (latched to ACIA (RS-232))
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: tx
                pulse
                delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: Version 1.0 original
         Version 1.1 change velocity data format

```

```

B174  00 00  velib  ldaa  #veladdh ;Get address of MSB of velocity
B175  17 00 00  staa  piadataA ;Write address of MSB of vel to GYRAC
B176  00 00 01  jsr   mpulse   ;Send GYRAC write pulse to latch data
B177  00 00 02  jsr   delay    ;Allow for A to D conversion time
B178  00 00 01  loadb piadataB ;Read MSB of velocity data
B179  00 00 03  jsr   tx       ;Output 1 byte of velocity to ACIA
B180  00

```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 11/21/85
VERSION: 1.1
NAME: ave101
MODULE NUMBER: 0.0
DESCRIPTION: This module sends a write pulse to the gyro
angular velocity data register to latch the data,
reads two bytes of ang vel data, reformat and
sends the data, MSB first, to the AQIA (A3-100)
output. Only the least significant 10 bits of
the data is significant.
PASSED VARIABLES: None
RETURNS: Two bytes of angular velocity data written to AQIA
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED:
    module
    delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: Version 1.0 original
        Version 1.1 change angular velocity data format
    
```

```

E101 06 00 ave101: ldaa #ave1addh ;Get address of MSB of ang vel
E102 07 10 00 staa #ave1addh ;Write MSB ang vel address to GYRAC
E103 08 02 00 jrn #pulse ;Send GYRAC write pulse to latch data
E104 09 00 00 jrn #delay ;Allow for A to D conversion time
E105 0A 10 01 ldaa #addata0 ;Read MSB of ang vel from GYRAC
E106 0B 00 staa temp ;Save raw MSB of data
E107 0C 04 lsr ;Right correct MSB of data
E108 0D 04 lsr
E109 0E 04 lsr
E110 0F 04 lsr
E111 10 04 lsr
E112 11 04 lsr
E113 12 ED 03 lsr ;Send MSB of ang vel to AQIA
E114 13 00 ldaa #ave1addl ;Get address of LSB of ang vel
E115 14 10 00 staa #ave1addl ;Write LSB ang vel address to GYRAC
E116 15 10 01 ldaa #addata1 ;Read LSB of ang vel from GYRAC
E117 16 04 lsr ;Adjust raw LSB to correct format
E118 17 04 lsr
E119 18 04 lsr
E120 19 04 lsr
E121 1A 04 lsr
E122 1B 04 lsr
E123 1C 00 ldaa temp ;Get raw MSB of data
E124 1D 04 asr ;Adjust raw MSB to correct L/D format
E125 1E 04 asr
E126 1F 04 staa temp ;Store corrected L/D in temp
E127 20 00 ldaa temp ;Get L/D of ang vel to AQIA
E128 21 00 staa temp
    
```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 11/17/88
VERSION: 1.1
AUTHOR: ave11b
MODULE NUMBER: 7.3
DESCRIPTION: This module sends a write pulse to the gyro
angular velocity data register to latch the data,
reads the MSB of ang vel data, and sends this
byte of ang vel data to the ACIA (PS-012) output.

PAIRED VARIABLES: None
RETURNS: One byte of ang vel data written to ACIA (PS-012)
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED:
    wpulse
    delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM D. RAMEY JR.
HISTORY: Version 1.0 original
        Version 1.1 change angular velocity data format
    
```

```

B100 36 30 ave11b 10aa waveladdr :Get address of MSB of ang vel
B101 37 30 00 staa piadataA :Write address of MSB of ang vel to PS-01
B102 38 31 20 .sr wpulse :Send GYRAC write pulse to latch data
B103 39 31 20 .sr delay :Allow for A to D conversion time
B104 40 31 20 .lsh readataB :Read MSB of angular velocity data
B105 41 31 20 .jsr tx :Output 1 byte of ang vel to ACIA
B106 42 31 20 .rts
    
```

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 04/06/85
VERSION: 1.1
NAME: clearvel
MODULE NUMBER: 1.0
DESCRIPTION: This module clears, sets F=0, the integration
             constant in the acceleration integrator. This is
             equivalent to zero velocity, initial condition.
             PLACING THE ADDRESS in the GYRAC address bus
             and sending a write pulse, clears the velocity
             constant. The address and write level must be
             REMOVED to stop clearing the constant.
PASSED VARIABLES: None
RETURNS: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. FAMEY JR.
HISTORY: Version 1.0 original
         Version 1.1 add delay subroutine

```

```

0101 26 09 clearvel 10aa #vconsadd ;Get address of vel const clear device
0102 27 09 00      staa piadataA ;Start clearing velocity constant
0103 05 14        ltab #write    ;Get write command
0104 27 09 01      stab piactrlA ;Send GYRAC write level to start clear
0105 0D E2 1E      jsr delay     ;Allow time for capacitor to discharge
0106 11 10        ltab #read     ;Get read command
0107 27 09 01      stab piactrlA ;Send GYRAC read level to stop clear
0108 1F          cpra            ;Pulse
0109 17 09 00      staa piadataA ;Stop clearing velocity constant
0110 17          rts

```


AD-A164 036

GYRO AND ACCELEROMETER BASED NAVIGATION SYSTEM FOR A
MOBILE AUTONOMOUS RO (U) AIR FORCE INST OF TECH
WRIGHT-PATTERSON AFB OH SCHOOL OF ENGI

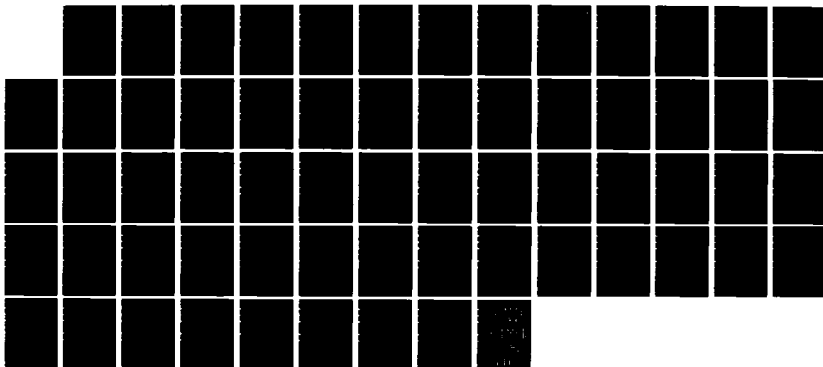
3/3

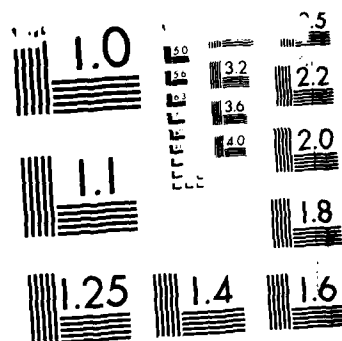
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MICROCOPY RESOLUTION TEST CHART
NATIONAL BUREAU OF STANDARDS-1963-A

GYRAC.A PROGRAM LISTING (continued)

```

DATE: 05/10/85
VERSION: 1.0
NAME: boot
MODULE NUMBER: 0.0
DESCRIPTION: This module will read a program from the ACIA
              (RS-232) serial port, load it into RAM memory,
              and then execute the program just loaded. This
              program does an unconditional jump to the loaded
              program.
PASSED VARIABLES: None
RETURNS: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, X
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: getcom
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: NONE

```

```

E1E2 50 E0 EF boot    jsr    getcom    ;Get the number of blocks to load
E1E3 17 01          staa   blocks    ;Save # of blocks in memory
E1E7 50 E0 EF        jsr    getcom    ;Get the number of bytes in a block
E1EA 07 02          staa   bytes     ;Save # of bytes in a block in memory
E1EC FE 60 00        ldw    rambot    ;Load pointer to RAM memory
E1EF 05 02          ldab   bytes     ;Get number of bytes
E1F1 50 E0 EF bytloop jsr    getcom    ;Get byte of data
E1F4 A7 00          staa   @,X       ;Load data byte into RAM
E1F6 0E            incw           ;Increment to next RAM pointer
E1F7 5A            decb           ;Decrement block byte count
E1F8 01 F7         bne    bytloop    ;If byte count not 0, get next byte
E1FA 7A 00 01       dec    blocks    ;Else block done, decrement blk count
E1FD 26 F2         bne    blkloop    ;If all block are not done, do another
E1FF 7E 60 00       jno    rambot    ;Else go execute loaded program

```

GYRAC.A PROGRAM LISTING (continued)

```

*****
DATE: 05/10/85
VERSION: 1.0
NAME: test
MODULE NUMBER: 1019
DESCRIPTION: This module will in command send to the ACIA
              (PS-220) output 10 printable ASCII characters.
              It will then return control to the calling
              program.
PASSED VARIABLES: None
RETURNS: 25 characters to ACIA
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: tx
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: NONE
*****

E002 16 21      test      ldab    #21H      ;Get first ASCII character
E004 07 03      stab     char      ;Save character in memory
E006 16 0D      ldaa     #09H      ;Get number of characters to be sent
E008 07 04      staa     charcnt    ;Save character count
E00A 04 03      charloop ldab     char      ;Get next ASCII character
E00C 0D 02 13    jsr      tx        ;Send ASCII character
E00E 70 06 03    inc     char      ;Increment to next ASCII character
E010 7A 09 04    dec     charcnt    ;Decrement character count
E012 16 F3      bne     charloop    ;If all characters have not been sent
E014 00 00      ;go send the next one
E016 00 00      ;Else return to calling program
E017 09

```

GYRAC.A PROGRAM LISTING (continued)

ac1a	E015	ac1adw1	E015	ac1adata	E011	ac1arss	E000
ac1astat	A000	ac1lib	E109	ac1b1	E185	ac1adddh	E000
ac1ad11	E000	begin	E000	bitloop	E1EF	ac1ar1s	E001
ac1b	E1E1	bytes	E000	bitloop	E1F1	char	E000
ac1b1	E004	charloop	E004	bitvel	E000	ch1A	E000
ac1b1	E000	ch1b	E000	ch1b	E004	ch1E	E004
ac1b1	E000	ch1b	E000	ch1b	E000	ch1l	E000
ac1b1	E000	ch1b	E000	ch1b	E004	ch1M	E000
ac1b1	E000	ch1b	E000	ch1A	E041	ch1B	E042
ac1b1	E042	ch1b	E044	ch1E	E045	ch1F	E046
ac1b1	E047	ch1b	E048	ch1l	E049	ch1l	E04A
ac1b1	E04E	ch1L	E040	ch1M	E040	ch1N	E04E
ac1b1	E04F	delay	E00E	getcom	E0EF	head1b	E119
head1b	E0FA	headadddh	E00E	headad11	E00F	init	E000
in1b1	E000	next	E000	pl1	E004	pl1ad1A	E001
pl1ad1B	E002	pl1adata	E000	pl1adataB	E002	pl1	E01A
pl1ad1	E004	pl1ad1	E000	pl1ad1	E002	pl1ad11	E000
pl1ad1	E001	pl1ad1	E000	pl1ad1	E000	pl1ad1	E001
read	E000	stack	E0FF	td1	E000	td1	E000
test	E000	test	E000	tx	E016	tx1adddh	E000
vector	E1F8	vel1b	E174	vel1b	E141	vel1adddh	E00A
vel1ad1	E00E	mpulse	E023	write	E004		

GYRAC.A OPERATING INSTRUCTIONS

STEP 1: Power up the H89 computer system. Place the System disk in Drive A and the gyro program disk in Drive B. Boot the system (type "B 29") and change the mode of Drive B to single sided double density (type "mode B:ss,dd"). Change the working drive to Drive B (type "B:").

STEP 2: Connect the H89 to GYRAC RS-232 cable between the H89 DCE connector and the MARRS-1 GYRAC Computer connector. Connect the external power cable to the GYRAC and turn on the power supplies. Flip the GYRAC power switch to the on position and press the GYRAC computer reset button. Flip the gyro control switch to the slaved mode for automatic tracking or the free mode for manual heading changes.

STEP 3: Load and run the M72 modem program on the H89 by typing "M72". When this program is running type "T" to enter the M72 Terminal mode. Set the H89 keyboard caps lock on.

STEP 4: Commands may now be issued to the GYRAC. They consist of single character capital letters from A thru O inclusive. The resulting hexadecimal data, as detailed below, is displayed (if it is printable) on the CRT. NOTE: Not all data returned is printable, so some results may not be displayed. Repeat STEP 4 as often as desired.

COMMAND	RESULT
A	Two bytes of heading data (12 valid bits) Two bytes of velocity data (10 valid bits) Two bytes of angular velocity data (10 valid bits)
B	One byte of heading data (8 valid bits) One byte of velocity data (8 valid bits) One byte of angular velocity data (8 valid bits)
C	Two bytes of heading data (12 valid bits)
D	One byte of heading data (8 valid bits)
E	Two bytes of velocity data (10 valid bits)
F	One byte of velocity data (8 valid bits)
G	Clear velocity constant (no data returned)

GYRAC.A OPERATING INSTRUCTIONS (continued)

COMMAND	RESULT
H	Two bytes of angular velocity data (10 valid bits)
I	One byte of angular velocity data (8 valid bits)
J	Two bytes of heading data (12 valid bits) Two bytes of velocity data (10 valid bits)
K	One byte of heading data (8 valid bits) One byte of velocity data (8 valid bits)
L	Soft reset of GYRAC computer (no data returned)
M	Boot load a program from the H89 into RAM and then execute to loaded program. No data is returned unless the user program sends it. Control is not automatically returned to the GYRAC control program, but is at the mercy of the boot program.
N	Returns 93 printable ASCII characters This tests the communications link
O	One byte of heading data (8 valid bits) Two bytes of velocity data (10 valid bits)

STEP 6: Shutdown all systems. Type "control shift " followed by "control E" to exit Terminal mode. To exit M72, type "CPM". Remove both disks from the drives and turn off the power to the H89 system. Turn off power to the GYRAC and external power supplies.

NOTE: All references to "control" and "shift" in the H89 command lines refer to the control and shift keys and not the words control and shift. It is assumed that the robot has been "pointed" to the desired initial heading before testing commences or that the gyro free mode will be used to vary the heading.

APPENDIX G

GTEST.A Structure Chart	G-2
GTEST.A Test Program Overlay	G-3
GTEST.A Operating Instructions	G-11

PUTD
OVERLAY

PUTC
OVERLAY

PUTB
OVERLAY

GTEST.A TEST PROGRAM OVERLAY

THE FOLLOWING
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PAGES

```

DATE: 7-10-75
VERSION: 1.1
TITLE: G-RAC DATA COLLECTION PROGRAM
FILENAME: GTEST.A
COORDINATOR: CAPT WILLIAM J. RAME, JR.
PROJECT: GYRO AND GYRO-METER BASED NAVIGATION SYSTEM FOR A
        MOBILE AUTONOMOUS ROBOT (THEMIS)
OPERATING SYSTEM: DEANCO, MSAM V1.2AL MAGNOLIA MICROSYSTEM 1000
LANGUAGE: ROBO-A ASSEMBLER V1.1 VIRTUAL DEVICES 1984
USE: This program is assembled and then transferred to the
     robot's navigation computer via an RS-232 link. It must
     be loaded into the nav computer using MOD in terminal (T)
     mode with the following command: L,CEFA,03E9 followed by
     transmitting the file gtest.hall. All of the original
     commands still work: G to begin transmitting data or H9,
     T to pause data Tx, C to resume data Tx, and D to
     download the nav computer. Each time the nav computer is
     rebooted (warm or cold) the gtest.hall file must be
     reloaded in order to send gyro data.

CONTENTS: putb
          putc
          puts

FUNCTION: This program overlays three existing subroutines in
         the nav computer's RAM memory to allow the nav
         computer to request data from the G-RAC (over an
         RS-232 link) and then transmit the G-RAC's data
         to an external computer over an RS-232 link. It
         replaces the subroutines that send sonar S, C, and
         D data with subroutines that send velocity, heading,
         and angular velocity respectively. The data is
         represented as four hexadecimal digits.
    
```

0000	begin	0000	0000	start address of the main func
0001	beginh	0001	0001	start address of the head in
0002	beginv	0002	0002	start address of the head in
0003	begin	0003	0003	start address of the head in
0004	begin	0004	0004	start address of the head in
0005	begin	0005	0005	start address of the head in
0006	begin	0006	0006	start address of the head in
0007	begin	0007	0007	start address of the head in
0008	begin	0008	0008	start address of the head in
0009	begin	0009	0009	start address of the head in
000A	begin	000A	000A	start address of the head in
000B	begin	000B	000B	start address of the head in
000C	begin	000C	000C	start address of the head in
000D	begin	000D	000D	start address of the head in
000E	begin	000E	000E	start address of the head in
000F	begin	000F	000F	start address of the head in
0010	begin	0010	0010	start address of the head in
0011	begin	0011	0011	start address of the head in
0012	begin	0012	0012	start address of the head in
0013	begin	0013	0013	start address of the head in
0014	begin	0014	0014	start address of the head in
0015	begin	0015	0015	start address of the head in
0016	begin	0016	0016	start address of the head in
0017	begin	0017	0017	start address of the head in
0018	begin	0018	0018	start address of the head in
0019	begin	0019	0019	start address of the head in
001A	begin	001A	001A	start address of the head in
001B	begin	001B	001B	start address of the head in
001C	begin	001C	001C	start address of the head in
001D	begin	001D	001D	start address of the head in
001E	begin	001E	001E	start address of the head in
001F	begin	001F	001F	start address of the head in
0020	begin	0020	0020	start address of the head in
0021	begin	0021	0021	start address of the head in
0022	begin	0022	0022	start address of the head in
0023	begin	0023	0023	start address of the head in
0024	begin	0024	0024	start address of the head in
0025	begin	0025	0025	start address of the head in
0026	begin	0026	0026	start address of the head in
0027	begin	0027	0027	start address of the head in
0028	begin	0028	0028	start address of the head in
0029	begin	0029	0029	start address of the head in
002A	begin	002A	002A	start address of the head in
002B	begin	002B	002B	start address of the head in
002C	begin	002C	002C	start address of the head in
002D	begin	002D	002D	start address of the head in
002E	begin	002E	002E	start address of the head in
002F	begin	002F	002F	start address of the head in
0030	begin	0030	0030	start address of the head in
0031	begin	0031	0031	start address of the head in
0032	begin	0032	0032	start address of the head in
0033	begin	0033	0033	start address of the head in
0034	begin	0034	0034	start address of the head in
0035	begin	0035	0035	start address of the head in
0036	begin	0036	0036	start address of the head in
0037	begin	0037	0037	start address of the head in
0038	begin	0038	0038	start address of the head in
0039	begin	0039	0039	start address of the head in
003A	begin	003A	003A	start address of the head in
003B	begin	003B	003B	start address of the head in
003C	begin	003C	003C	start address of the head in
003D	begin	003D	003D	start address of the head in
003E	begin	003E	003E	start address of the head in
003F	begin	003F	003F	start address of the head in
0040	begin	0040	0040	start address of the head in
0041	begin	0041	0041	start address of the head in
0042	begin	0042	0042	start address of the head in
0043	begin	0043	0043	start address of the head in
0044	begin	0044	0044	start address of the head in
0045	begin	0045	0045	start address of the head in
0046	begin	0046	0046	start address of the head in
0047	begin	0047	0047	start address of the head in
0048	begin	0048	0048	start address of the head in
0049	begin	0049	0049	start address of the head in
004A	begin	004A	004A	start address of the head in
004B	begin	004B	004B	start address of the head in
004C	begin	004C	004C	start address of the head in
004D	begin	004D	004D	start address of the head in
004E	begin	004E	004E	start address of the head in
004F	begin	004F	004F	start address of the head in
0050	begin	0050	0050	start address of the head in
0051	begin	0051	0051	start address of the head in
0052	begin	0052	0052	start address of the head in
0053	begin	0053	0053	start address of the head in
0054	begin	0054	0054	start address of the head in
0055	begin	0055	0055	start address of the head in
0056	begin	0056	0056	start address of the head in
0057	begin	0057	0057	start address of the head in
0058	begin	0058	0058	start address of the head in
0059	begin	0059	0059	start address of the head in
005A	begin	005A	005A	start address of the head in
005B	begin	005B	005B	start address of the head in
005C	begin	005C	005C	start address of the head in
005D	begin	005D	005D	start address of the head in
005E	begin	005E	005E	start address of the head in
005F	begin	005F	005F	start address of the head in
0060	begin	0060	0060	start address of the head in
0061	begin	0061	0061	start address of the head in
0062	begin	0062	0062	start address of the head in
0063	begin	0063	0063	start address of the head in
0064	begin	0064	0064	start address of the head in
0065	begin	0065	0065	start address of the head in
0066	begin	0066	0066	start address of the head in
0067	begin	0067	0067	start address of the head in
0068	begin	0068	0068	start address of the head in
0069	begin	0069	0069	start address of the head in
006A	begin	006A	006A	start address of the head in
006B	begin	006B	006B	start address of the head in
006C	begin	006C	006C	start address of the head in
006D	begin	006D	006D	start address of the head in
006E	begin	006E	006E	start address of the head in
006F	begin	006F	006F	start address of the head in
0070	begin	0070	0070	start address of the head in
0071	begin	0071	0071	start address of the head in
0072	begin	0072	0072	start address of the head in
0073	begin	0073	0073	start address of the head in
0074	begin	0074	0074	start address of the head in
0075	begin	0075	0075	start address of the head in
0076	begin	0076	0076	start address of the head in
0077	begin	0077	0077	start address of the head in
0078	begin	0078	0078	start address of the head in
0079	begin	0079	0079	start address of the head in
007A	begin	007A	007A	start address of the head in
007B	begin	007B	007B	start address of the head in
007C	begin	007C	007C	start address of the head in
007D	begin	007D	007D	start address of the head in
007E	begin	007E	007E	start address of the head in
007F	begin	007F	007F	start address of the head in
0080	begin	0080	0080	start address of the head in
0081	begin	0081	0081	start address of the head in
0082	begin	0082	0082	start address of the head in
0083	begin	0083	0083	start address of the head in
0084	begin	0084	0084	start address of the head in
0085	begin	0085	0085	start address of the head in
0086	begin	0086	0086	start address of the head in
0087	begin	0087	0087	start address of the head in
0088	begin	0088	0088	start address of the head in
0089	begin	0089	0089	start address of the head in
008A	begin	008A	008A	start address of the head in
008B	begin	008B	008B	start address of the head in
008C	begin	008C	008C	start address of the head in
008D	begin	008D	008D	start address of the head in
008E	begin	008E	008E	start address of the head in
008F	begin	008F	008F	start address of the head in
0090	begin	0090	0090	start address of the head in
0091	begin	0091	0091	start address of the head in
0092	begin	0092	0092	start address of the head in
0093	begin	0093	0093	start address of the head in
0094	begin	0094	0094	start address of the head in
0095	begin	0095	0095	start address of the head in
0096	begin	0096	0096	start address of the head in
0097	begin	0097	0097	start address of the head in
0098	begin	0098	0098	start address of the head in
0099	begin	0099	0099	start address of the head in
009A	begin	009A	009A	start address of the head in
009B	begin	009B	009B	start address of the head in
009C	begin	009C	009C	start address of the head in
009D	begin	009D	009D	start address of the head in
009E	begin	009E	009E	start address of the head in
009F	begin	009F	009F	start address of the head in
00A0	begin	00A0	00A0	start address of the head in
00A1	begin	00A1	00A1	start address of the head in
00A2	begin	00A2	00A2	start address of the head in
00A3	begin	00A3	00A3	start address of the head in
00A4	begin	00A4	00A4	start address of the head in
00A5	begin	00A5	00A5	start address of the head in
00A6	begin	00A6	00A6	start address of the head in
00A7	begin	00A7	00A7	start address of the head in
00A8	begin	00A8	00A8	start address of the head in
00A9	begin	00A9	00A9	start address of the head in
00AA	begin	00AA	00AA	start address of the head in
00AB	begin	00AB	00AB	start address of the head in
00AC	begin	00AC	00AC	start address of the head in
00AD	begin	00AD	00AD	start address of the head in
00AE	begin	00AE	00AE	start address of the head in
00AF	begin	00AF	00AF	start address of the head in
00B0	begin	00B0	00B0	start address of the head in
00B1	begin	00B1	00B1	start address of the head in
00B2	begin	00B2	00B2	start address of the head in
00B3	begin	00B3	00B3	start address of the head in
00B4	begin	00B4	00B4	start address of the head in
00B5	begin	00B5	00B5	start address of the head in
00B6	begin	00B6	00B6	start address of the head in
00B7	begin	00B7	00B7	start address of the head in
00B8	begin	00B8	00B8	start address of the head in
00B9	begin	00B9	00B9	start address of the head in
00BA	begin	00BA	00BA	start address of the head in
00BB	begin	00BB	00BB	start address of the head in
00BC	begin	00BC	00BC	start address of the head in
00BD	begin	00BD	00BD	start address of the head in
00BE	begin	00BE	00BE	start address of the head in
00BF	begin	00BF	00BF	start address of the head in
00C0	begin	00C0	00C0	start address of the head in
00C1	begin	00C1	00C1	start address of the head in
00C2	begin	00C2	00C2	start address of the head in
00C3	begin	00C3	00C3	start address of the head in
00C4	begin	00C4	00C4	start address of the head in
00C5	begin	00C5	00C5	start address of the head in
00C6	begin	00C6	00C6	start address of the head in
00C7	begin	00C7	00C7	start address of the head in
00C8	begin	00C8	00C8	start address of the head in
00C9	begin	00C9	00C9	start address of the head in
00CA	begin	00CA	00CA	start address of the head in
00CB	begin	00CB	00CB	start address of the head in
00CC	begin	00CC	00CC	start address of the head in
00CD	begin	00CD	00CD	start address of the head in
00CE	begin	00CE	00CE	start address of the head in
00CF	begin	00CF	00CF	start address of the head in
00D0	begin	00D0	00D0	start address of the head in
00D1	begin	00D1	00D1	start address of the head in
00D2	begin	00D2	00D2	start address of the head in
00D3	begin	00D3		

GTEST.A TEST PROGRAM OVERLAY (continued)

```

DATE: 10/19/83
VERSION: 1.0
NAME: putt
MODULE NUMBER: 1.0
DESCRIPTION: This module reads 2 bytes of velocity data from
the GPRAC, converts it to four hexadecimal
characters and places the characters in the output
buffer.
PARSED VARIABLES: None
RETURN: Four hex velocity characters in the data buffer.
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, D, X
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: putchar
CALLING MODULES: extend
AUTHOR: CAPT WILLIAM L. RAMEY JR.
HISTORY: NONE

```

```

00FA 0E 05 40 00b0 00 beginv ;Start of putt subroutine
00FB 0E 05 40 00b0 00 ;Point to velocity data buffer
00FC 0E 05 40 00b0 00 ;Get GPRAC 2 byte vel command
00FD 0E 05 40 00b0 00 ;Get GPRAC ACIA status
00FE 0E 05 40 00b0 00 ;Is transmit data reg empty?
00FF 0E 05 40 00b0 00 ;If no loop to 00b0 & recheck status
0100 0E 05 40 00b0 00 ;Else send GPRAC command
0101 0E 05 40 00b0 00 ;Get GPRAC ACIA status
0102 0E 05 40 00b0 00 ;Is receive data register full?
0103 0E 05 40 00b0 00 ;If no loop to 00b0 & recheck status
0104 0E 05 40 00b0 00 ;Else get the MSD of velocity data
0105 0E 05 40 00b0 00 ;Convert data to hex & save in buffer
0106 0E 05 40 00b0 00 ;Get GPRAC ACIA status
0107 0E 05 40 00b0 00 ;Is receive data register full?
0108 0E 05 40 00b0 00 ;If no loop to 00b0 & recheck status
0109 0E 05 40 00b0 00 ;Else get the LSD of velocity data
010A 0E 05 40 00b0 00 ;Convert data to hex & save in buffer
010B 0E 05 40 00b0 00 ;Drawn out remainder of old program
010C 0E 05 40 00b0 00
010D 0E 05 40 00b0 00
010E 0E 05 40 00b0 00
010F 0E 05 40 00b0 00
0110 0E 05 40 00b0 00
0111 0E 05 40 00b0 00
0112 0E 05 40 00b0 00
0113 0E 05 40 00b0 00
0114 0E 05 40 00b0 00
0115 0E 05 40 00b0 00
0116 0E 05 40 00b0 00
0117 0E 05 40 00b0 00
0118 0E 05 40 00b0 00
0119 0E 05 40 00b0 00
011A 0E 05 40 00b0 00
011B 0E 05 40 00b0 00
011C 0E 05 40 00b0 00
011D 0E 05 40 00b0 00
011E 0E 05 40 00b0 00
011F 0E 05 40 00b0 00
0120 0E 05 40 00b0 00
0121 0E 05 40 00b0 00
0122 0E 05 40 00b0 00
0123 0E 05 40 00b0 00
0124 0E 05 40 00b0 00
0125 0E 05 40 00b0 00
0126 0E 05 40 00b0 00
0127 0E 05 40 00b0 00
0128 0E 05 40 00b0 00
0129 0E 05 40 00b0 00
012A 0E 05 40 00b0 00
012B 0E 05 40 00b0 00
012C 0E 05 40 00b0 00
012D 0E 05 40 00b0 00
012E 0E 05 40 00b0 00
012F 0E 05 40 00b0 00
0130 0E 05 40 00b0 00
0131 0E 05 40 00b0 00
0132 0E 05 40 00b0 00
0133 0E 05 40 00b0 00
0134 0E 05 40 00b0 00
0135 0E 05 40 00b0 00
0136 0E 05 40 00b0 00
0137 0E 05 40 00b0 00
0138 0E 05 40 00b0 00
0139 0E 05 40 00b0 00
013A 0E 05 40 00b0 00
013B 0E 05 40 00b0 00
013C 0E 05 40 00b0 00
013D 0E 05 40 00b0 00
013E 0E 05 40 00b0 00
013F 0E 05 40 00b0 00
0140 0E 05 40 00b0 00
0141 0E 05 40 00b0 00
0142 0E 05 40 00b0 00
0143 0E 05 40 00b0 00
0144 0E 05 40 00b0 00
0145 0E 05 40 00b0 00
0146 0E 05 40 00b0 00
0147 0E 05 40 00b0 00
0148 0E 05 40 00b0 00
0149 0E 05 40 00b0 00
014A 0E 05 40 00b0 00
014B 0E 05 40 00b0 00
014C 0E 05 40 00b0 00
014D 0E 05 40 00b0 00
014E 0E 05 40 00b0 00
014F 0E 05 40 00b0 00
0150 0E 05 40 00b0 00
0151 0E 05 40 00b0 00
0152 0E 05 40 00b0 00
0153 0E 05 40 00b0 00
0154 0E 05 40 00b0 00
0155 0E 05 40 00b0 00
0156 0E 05 40 00b0 00
0157 0E 05 40 00b0 00
0158 0E 05 40 00b0 00
0159 0E 05 40 00b0 00
015A 0E 05 40 00b0 00
015B 0E 05 40 00b0 00
015C 0E 05 40 00b0 00
015D 0E 05 40 00b0 00
015E 0E 05 40 00b0 00
015F 0E 05 40 00b0 00
0160 0E 05 40 00b0 00
0161 0E 05 40 00b0 00
0162 0E 05 40 00b0 00
0163 0E 05 40 00b0 00
0164 0E 05 40 00b0 00
0165 0E 05 40 00b0 00
0166 0E 05 40 00b0 00
0167 0E 05 40 00b0 00
0168 0E 05 40 00b0 00
0169 0E 05 40 00b0 00
016A 0E 05 40 00b0 00
016B 0E 05 40 00b0 00
016C 0E 05 40 00b0 00
016D 0E 05 40 00b0 00
016E 0E 05 40 00b0 00
016F 0E 05 40 00b0 00
0170 0E 05 40 00b0 00
0171 0E 05 40 00b0 00
0172 0E 05 40 00b0 00
0173 0E 05 40 00b0 00
0174 0E 05 40 00b0 00
0175 0E 05 40 00b0 00
0176 0E 05 40 00b0 00
0177 0E 05 40 00b0 00
0178 0E 05 40 00b0 00
0179 0E 05 40 00b0 00
017A 0E 05 40 00b0 00
017B 0E 05 40 00b0 00
017C 0E 05 40 00b0 00
017D 0E 05 40 00b0 00
017E 0E 05 40 00b0 00
017F 0E 05 40 00b0 00
0180 0E 05 40 00b0 00
0181 0E 05 40 00b0 00
0182 0E 05 40 00b0 00
0183 0E 05 40 00b0 00
0184 0E 05 40 00b0 00
0185 0E 05 40 00b0 00
0186 0E 05 40 00b0 00
0187 0E 05 40 00b0 00
0188 0E 05 40 00b0 00
0189 0E 05 40 00b0 00
018A 0E 05 40 00b0 00
018B 0E 05 40 00b0 00
018C 0E 05 40 00b0 00
018D 0E 05 40 00b0 00
018E 0E 05 40 00b0 00
018F 0E 05 40 00b0 00
0190 0E 05 40 00b0 00
0191 0E 05 40 00b0 00
0192 0E 05 40 00b0 00
0193 0E 05 40 00b0 00
0194 0E 05 40 00b0 00
0195 0E 05 40 00b0 00
0196 0E 05 40 00b0 00
0197 0E 05 40 00b0 00
0198 0E 05 40 00b0 00
0199 0E 05 40 00b0 00
019A 0E 05 40 00b0 00
019B 0E 05 40 00b0 00
019C 0E 05 40 00b0 00
019D 0E 05 40 00b0 00
019E 0E 05 40 00b0 00
019F 0E 05 40 00b0 00
01A0 0E 05 40 00b0 00
01A1 0E 05 40 00b0 00
01A2 0E 05 40 00b0 00
01A3 0E 05 40 00b0 00
01A4 0E 05 40 00b0 00
01A5 0E 05 40 00b0 00
01A6 0E 05 40 00b0 00
01A7 0E 05 40 00b0 00
01A8 0E 05 40 00b0 00
01A9 0E 05 40 00b0 00
01AA 0E 05 40 00b0 00
01AB 0E 05 40 00b0 00
01AC 0E 05 40 00b0 00
01AD 0E 05 40 00b0 00
01AE 0E 05 40 00b0 00
01AF 0E 05 40 00b0 00
01B0 0E 05 40 00b0 00
01B1 0E 05 40 00b0 00
01B2 0E 05 40 00b0 00
01B3 0E 05 40 00b0 00
01B4 0E 05 40 00b0 00
01B5 0E 05 40 00b0 00
01B6 0E 05 40 00b0 00
01B7 0E 05 40 00b0 00
01B8 0E 05 40 00b0 00
01B9 0E 05 40 00b0 00
01BA 0E 05 40 00b0 00
01BB 0E 05 40 00b0 00
01BC 0E 05 40 00b0 00
01BD 0E 05 40 00b0 00
01BE 0E 05 40 00b0 00
01BF 0E 05 40 00b0 00
01C0 0E 05 40 00b0 00
01C1 0E 05 40 00b0 00
01C2 0E 05 40 00b0 00
01C3 0E 05 40 00b0 00
01C4 0E 05 40 00b0 00
01C5 0E 05 40 00b0 00
01C6 0E 05 40 00b0 00
01C7 0E 05 40 00b0 00
01C8 0E 05 40 00b0 00
01C9 0E 05 40 00b0 00
01CA 0E 05 40 00b0 00
01CB 0E 05 40 00b0 00
01CC 0E 05 40 00b0 00
01CD 0E 05 40 00b0 00
01CE 0E 05 40 00b0 00
01CF 0E 05 40 00b0 00
01D0 0E 05 40 00b0 00
01D1 0E 05 40 00b0 00
01D2 0E 05 40 00b0 00
01D3 0E 05 40 00b0 00
01D4 0E 05 40 00b0 00
01D5 0E 05 40 00b0 00
01D6 0E 05 40 00b0 00
01D7 0E 05 40 00b0 00
01D8 0E 05 40 00b0 00
01D9 0E 05 40 00b0 00
01DA 0E 05 40 00b0 00
01DB 0E 05 40 00b0 00
01DC 0E 05 40 00b0 00
01DD 0E 05 40 00b0 00
01DE 0E 05 40 00b0 00
01DF 0E 05 40 00b0 00
01E0 0E 05 40 00b0 00
01E1 0E 05 40 00b0 00
01E2 0E 05 40 00b0 00
01E3 0E 05 40 00b0 00
01E4 0E 05 40 00b0 00
01E5 0E 05 40 00b0 00
01E6 0E 05 40 00b0 00
01E7 0E 05 40 00b0 00
01E8 0E 05 40 00b0 00
01E9 0E 05 40 00b0 00
01EA 0E 05 40 00b0 00
01EB 0E 05 40 00b0 00
01EC 0E 05 40 00b0 00
01ED 0E 05 40 00b0 00
01EE 0E 05 40 00b0 00
01EF 0E 05 40 00b0 00
01F0 0E 05 40 00b0 00
01F1 0E 05 40 00b0 00
01F2 0E 05 40 00b0 00
01F3 0E 05 40 00b0 00
01F4 0E 05 40 00b0 00
01F5 0E 05 40 00b0 00
01F6 0E 05 40 00b0 00
01F7 0E 05 40 00b0 00
01F8 0E 05 40 00b0 00
01F9 0E 05 40 00b0 00
01FA 0E 05 40 00b0 00
01FB 0E 05 40 00b0 00
01FC 0E 05 40 00b0 00
01FD 0E 05 40 00b0 00
01FE 0E 05 40 00b0 00
01FF 0E 05 40 00b0 00

```

GTEST.A TEST PROGRAM OVERLAY (continued)

0107	01	END
0108	01	END
0109	01	END
0110	01	END
0111	01	END
0112	01	END
0113	01	END
0114	01	END
0115	01	END
0116	01	END
0117	01	END
0118	01	END
0119	01	END
0120	01	END
0121	01	END
0122	01	END
0123	01	END
0124	01	END
0125	01	END
0126	01	END
0127	01	END
0128	01	END
0129	01	END
0130	01	END
0131	01	END
0132	01	END
0133	01	END
0134	01	END
0135	01	END
0136	01	END
0137	01	END
0138	01	END
0139	01	END
0140	01	END
0141	01	END
0142	01	END
0143	01	END
0144	01	END
0145	01	END
0146	01	END
0147	01	END
0148	01	END
0149	01	END
0150	01	END

GTEST.A TEST PROGRAM OVERLAY (continued)

```

DATE: 12/11/85
VERSION: 1.1
NAME: puts
MODULE NUMBER: 1
DESCRIPTION: This module reads two bytes of heading data from
the GVRAC, converts it to four hexidecimal
characters and places the characters in the output
buffer. The most significant nibble is masked
out.
RAISED VARIABLES: None
RETURNS: Four hex heading characters in the data buffer.
GLOBAL VARIABLES USED: None
LOCAL VARIABLES CHANGED: None
REGISTERS USED: A, D, X
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: putchar
CALLING MODULES: tend
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: Version 1.0 original program
Version 1.1 add mask to MS nibble of the heading

```

```

004A CE 25 50 puts      ldx #incheading ;Point to heading data buffer
004C 03 43             ldab #comd      ;Get GVRAC 2 byte heading command
004E 05 08 10 txc      ldab acialstat ;Get GVRAC ACIA status
0050 34 62             anda #tdre     ;Is transmit data reg empty?
0052 27 F2             beq txc        ;If no loop to txc and recheck status
0054 F7 02 11          stab acialdata ;Else send GVRAC command
0056

0059 34 08 10 rxcm      ldab acialstat ;Get GVRAC ACIA status
005B 34 31             anda #drxf     ;Is receive data register full?
005D 27 F2             beq rxcm       ;If no loop to rxcm & recheck status
005F 34 08 11          ldab acialdata ;Else get the MSB of heading data
0061 34 0F             anda #mask     ;Mask out most significant nibble
0063 20 02 50          lsr outche     ;Convert data to hex & save in buffer
0065

0068 34 08 10 rxcm      ldab acialstat ;Get GVRAC ACIA status
006A 34 31             anda #drxf     ;Is receive data register full?
006C 27 F2             beq rxcm       ;If no loop to rxcm & recheck status
006E 34 08 11          ldab acialdata ;Else get the LSB of heading data
0070 20 02 50          lsr outche     ;Convert data to hex & save in buffer
0072

0074 01               ror             ;Shift out remainder of hex digit
0076 01               ror
0078 01               ror
007A 01               ror
007C 01               ror
007E 01               ror
0080 01               ror
0082 01               ror
0084 01               ror
0086 01               ror
0088 01               ror
008A 01               ror
008C 01               ror
008E 01               ror
0090 01               ror
0092 01               ror
0094 01               ror
0096 01               ror
0098 01               ror
009A 01               ror
009C 01               ror
009E 01               ror
00A0 01               ror
00A2 01               ror
00A4 01               ror
00A6 01               ror
00A8 01               ror
00AA 01               ror
00AC 01               ror
00AE 01               ror
00B0 01               ror
00B2 01               ror
00B4 01               ror
00B6 01               ror
00B8 01               ror
00BA 01               ror
00BC 01               ror
00BE 01               ror
00C0 01               ror
00C2 01               ror
00C4 01               ror
00C6 01               ror
00C8 01               ror
00CA 01               ror
00CC 01               ror
00CE 01               ror
00D0 01               ror
00D2 01               ror
00D4 01               ror
00D6 01               ror
00D8 01               ror
00DA 01               ror
00DC 01               ror
00DE 01               ror
00E0 01               ror
00E2 01               ror
00E4 01               ror
00E6 01               ror
00E8 01               ror
00EA 01               ror
00EC 01               ror
00EE 01               ror
00F0 01               ror
00F2 01               ror
00F4 01               ror
00F6 01               ror
00F8 01               ror
00FA 01               ror
00FC 01               ror
00FE 01               ror
00FF 01               ror

```

GTEST.A TEST PROGRAM OVERLAY (continued)

0007	01	nop
0008	01	nop
0009	01	nop
000A	01	nop
000B	01	nop
000C	01	nop
000D	01	nop
000E	01	nop
000F	01	nop
0010	01	nop
0011	01	nop
0012	01	nop
0013	01	nop
0014	01	nop
0015	01	nop
0016	01	nop
0017	01	nop
0018	01	nop
0019	01	nop

GTEST.A TEST PROGRAM OVERLAY (continued)

```

*****
DATE: 07 17 85
VERSION: 1.0
NAME: gtest
MODULE NUMBER: 1.0
DESCRIPTION: This module reads two bytes of ang velocity data
             from the GVRAC, converts it to four hexadecimal
             characters and places the characters in the output
             buffer.
PASSED VARIABLES: None
RETURNS: Four hex ang velocity characters in the data buffer.
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, D, X
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: puthex
CALLING MODULES: send
AUTHOR: CAPT WILLIAM O. RAMEY JR.
HISTORY: NONE
*****

03FA 0E 05 5A puts      ldx #Indreading ;Point to ang velocity data buffer
03FD 03 48          ldab #comH ;Get GVRAC 2 byte ang vel command
03FF 05 00 10 bnd      ldaa acialstat ;Get GVRAC ACIA status
04A2 04 02          anda #tdre ;Is transmit data reg empty?
04A4 17 F0          beq bnd ;If no loop to bnd and recheck status
04A6 F7 00 11          stab acialdata ;Else send GVRAC command

04A9 05 00 10 rndr      ldaa acialstat ;Get GVRAC ACIA status
04AC 04 01          anda #ndrf ;Is receive data register full?
04AE 17 F0          beq rndm ;If no loop to rndm & recheck status
04B0 05 00 11          ldaa acialdata ;Else get the MSB of ang velocity data
04B3 10 02 53          jsr put2hex ;Convert data to hex & save in buffer

04B6 05 00 10 rndl      ldaa acialstat ;Get GVRAC ACIA status
04B9 04 01          anda #ndrf ;Is receive data register full?
04BB 17 F0          beq rndl ;If no loop to rndl & recheck status
04BD 05 00 11          ldaa acialdata ;Else get the LSB of ang velocity data
04BF 10 02 53          jsr put2hex ;Convert data to hex & save in buffer
04C2 17          rts ;Blank out remainder of old program
04C3 01          nop
04C4 01          nop
04C5 01          nop
04C6 01          nop
04C7 01          nop
04C8 01          nop
04C9 01          nop
04CA 01          nop
04CB 01          nop
04CC 01          nop
04CD 01          nop
04CE 01          nop
04CF 01          nop
04D0 01          nop
04D1 01          nop
04D2 01          nop
04D3 01          nop
04D4 01          nop
04D5 01          nop
04D6 01          nop
04D7 01          nop
04D8 01          nop
04D9 01          nop
04DA 01          nop
04DB 01          nop
04DC 01          nop
04DD 01          nop
04DE 01          nop
04DF 01          nop
04E0 01          nop
04E1 01          nop
04E2 01          nop
04E3 01          nop
04E4 01          nop
04E5 01          nop
04E6 01          nop
04E7 01          nop
04E8 01          nop
04E9 01          nop
04EA 01          nop
04EB 01          nop
04EC 01          nop
04ED 01          nop
04EE 01          nop
04EF 01          nop
04F0 01          nop
04F1 01          nop
04F2 01          nop
04F3 01          nop
04F4 01          nop
04F5 01          nop
04F6 01          nop
04F7 01          nop
04F8 01          nop
04F9 01          nop
04FA 01          nop
04FB 01          nop
04FC 01          nop
04FD 01          nop
04FE 01          nop
04FF 01          nop

```

GTEST.A TEST PROGRAM OVERLAY (continued)

03D8	01	nop
03D9	01	nop
03DA	01	nop
03DB	01	nop
03DC	01	nop
03DD	01	nop
03DE	01	nop
03DF	01	nop
03E0	01	nop
03E1	01	nop
03E2	01	nop
03E3	01	nop
03E4	01	nop
03E5	01	nop
03E6	01	nop
03E7	01	nop
03E8	01	nop
03E9	01	nop

03EA	END
------	-----

GTEST.A TEST PROGRAM OVERLAY (continued)

[illegible]

GTEST.A OPERATING INSTRUCTIONS

STEP 1: Power up the H89 computer system. Place the System disk in Drive A and the gyro program disk in Drive B. Boot the system (type "B 29") and change the mode of Drive B to single sided double density (type "mode B:ss,dd"). Change the working drive to Drive B (type "B:").

STEP 2: Power up the MARRS-1 Robot. Make sure the batteries are fully charged and the charger power line is connected and turned on. Press both the system reset key on the keypad and the Nav computer reset button.

STEP 3: Load and run the M72 modem program on the H89 by typing "M72". When this program is running type "SPD" to change the transmission time delays. When prompted for time delays reply with a "1" for both character and line delay times. Set the H89 keyboard caps lock on.

STEP 4: Connect the H89 to Robot RS-232 cable between the H89 DCE connector and the MARRS-1 Nav T connector. Connect the GYRAC to Nav Computer RS-232 cable between the GYRAC connector and the Nav L connector on MARRS-1. Connect the teaching pendant cable to MARRS-1. Connect the external power cable to the GYRAC and turn on the power supplies. Flip the GYRAC power switch to the on position and press the GYRAC computer reset button. Flip the gyro control switch to the slaved mode.

STEP 5: Load and transmit the GTEST.HEX file to the robot's navigation computer. This is done by typing "L,02FA,03E9" to load the file at Nav computer memory address 02FA (HEX). Next type "T filename". This will place the CRT in terminal mode and create an input buffer to store incoming data in disk file filename. Follow this by typing "control shift " then "control T" and "GTEST.HEX" to transmit the program file to the Nav computer. Reply with Yes when asked for time delays. The program data will be displayed as it is transmitted. If an error is made in STEP 5, the navigation reset button must be pressed and the entire step done over.

STEP 6: Begin program execution. First type "control shift " and then "control Y" to open the input data buffer. Now type "C" to begin program execution. Next steer the robot using the teaching pendant (MARRS-1 manual mode, 4) along the desired course. During program execution the robot will send optical shaft encoder data, heading data, angular velocity data, and velocity data at 0.1 second time

GTEST.A OPERATING INSTRUCTIONS (continued)

intervals to the H89 computer. This data will be displayed on the CRT and stored in the input buffer.

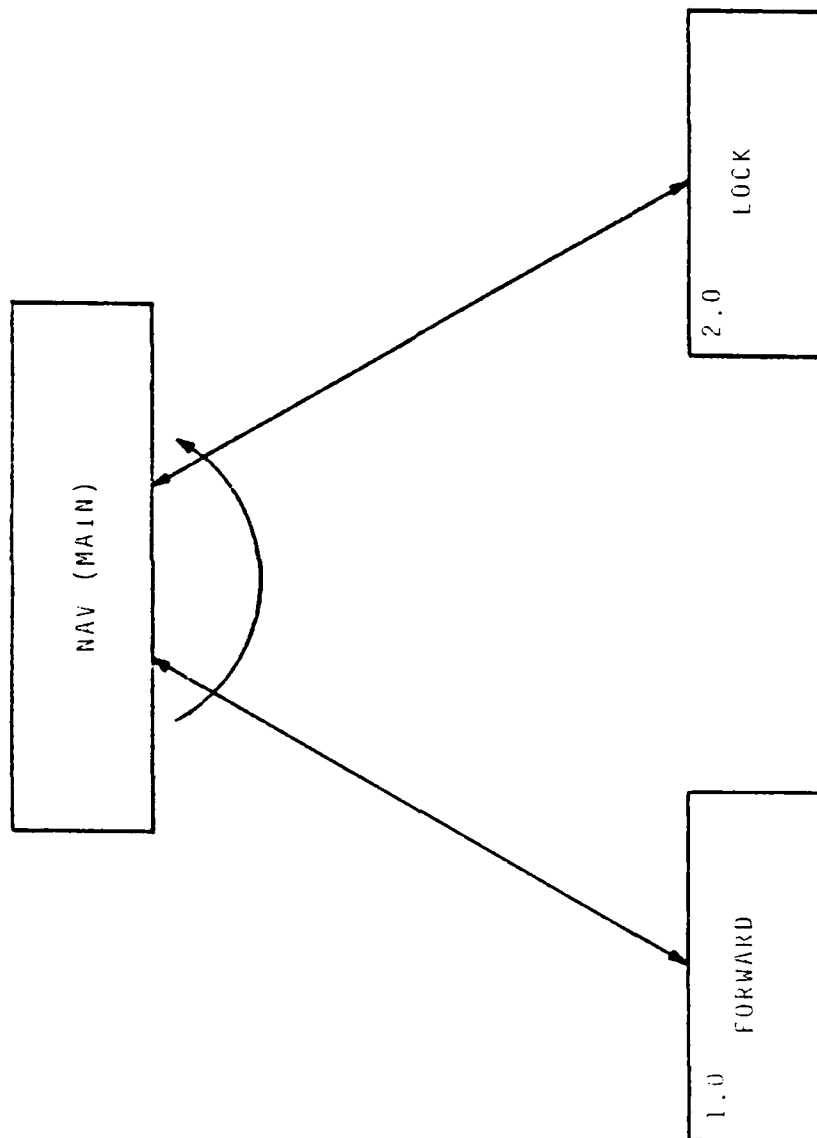
STEP 7: When the robot run is completed (i.e. you have manually stopped it with the MARRS-1 system reset button) the data stored in the input buffer may be written to disk. To do this press the Nav Computer reset button on MARRS-1 and/or type "control C" on the H89. Next type "control shift " followed by "control E". Now type "WRT" to save the data to disk ("del" may also be typed to dump buffered data). If additional runs are required continue with STEP 5 and press all three MARRS-1 reset buttons.

STEP 8: Shutdown all systems. To exit M72, type "CPM". Remove both disks from the drives and turn off the power to the H89 system. Turn off power to the robot, GYRAC, and external power supplies.

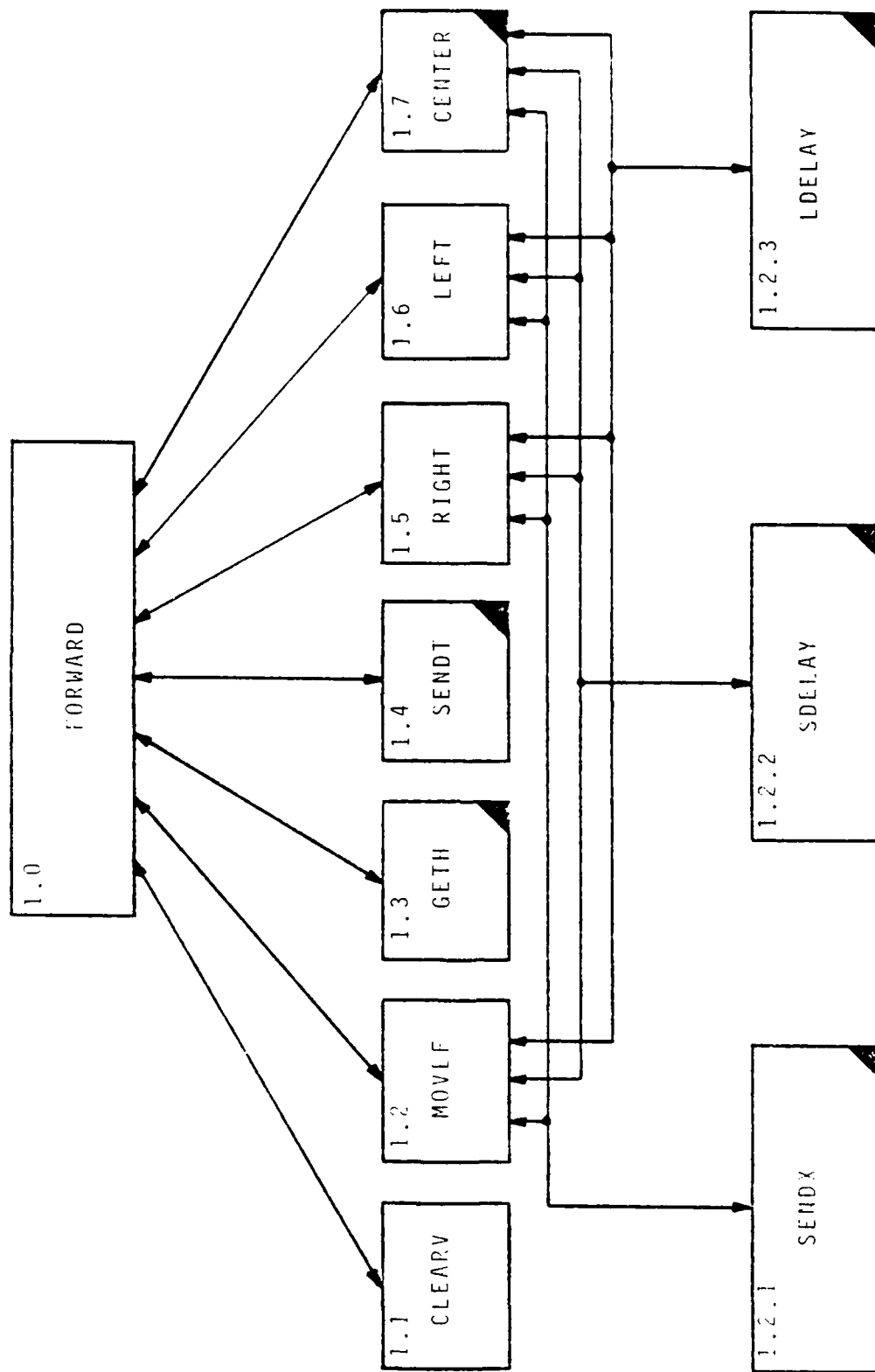
NOTE: All references to "control" and "shift" in the H89 command lines refer to the control and shift keys and not the words control and shift. Care must be taken to ensure the various cables to MARRS-1 do not become tangled during movement. In addition, it is assumed that the robot has been "pointed" to the desired initial heading before movement commences. The actual direction of travel is human controlled from the teaching pendant. Also, the GTEST.HEX program must be loaded each time a run is attempted, since the program is cleared on navigation computer reset or by a "control C" in M72 Terminal mode.

APPENDIX H

NAV.A Structure Charts	H-2
NAV.A Program Listing	H-5
MARRS.NAV Program Listing	H-26
NAV.A Operating Instructions	H-27

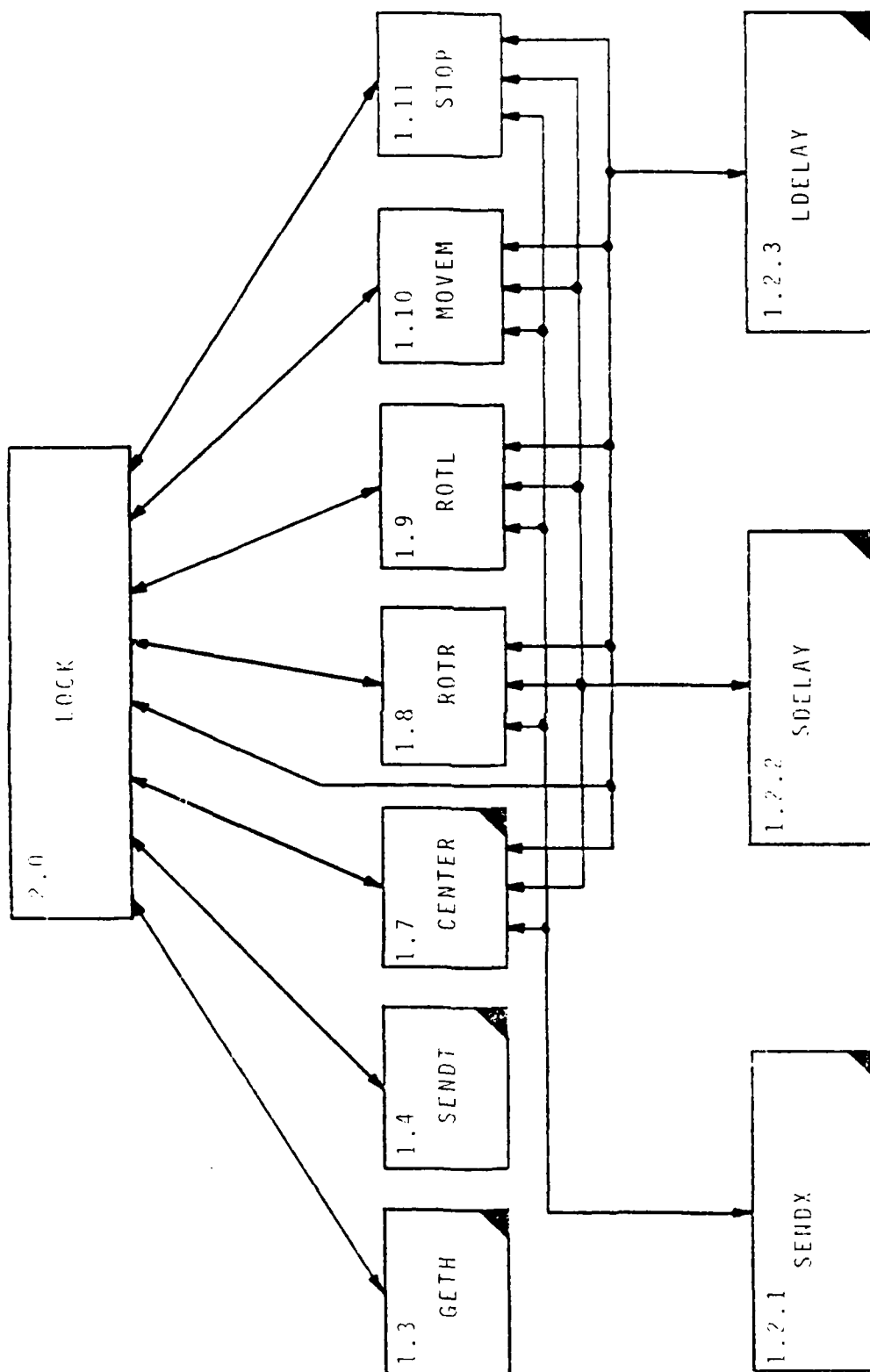


NAV.A STRUCTURE CHARTS



H-3

NAV.A STRUCTURE CHARTS (continued)



H-4

NAV.A STRUCTURE CHARTS (continued)

NAV.A PROGRAM LISTING

NAV.A PROGRAM LISTING
 NAV.A PROGRAM LISTING
 NAV.A PROGRAM LISTING
 NAV.A PROGRAM LISTING
 NAV.A PROGRAM LISTING

DATE: 19/12/85
 VERSION: 1.4
 TITLE: NAVIGATION AND HEADING DATA COLLECTION PROGRAM
 FILENAME: NAV.A
 COORDINATOR: CAPT WILLIAM J. FAMEL JR.
 PROJECT: GYRO AND ACCELEROMETER BASED NAVIGATION SYSTEM FOR A
 MOBILE AUTONOMOUS ROBOT (THESIS)
 OPERATING SYSTEM: IBM PC/XT/AT V2.142 MAGNOLIA MICROSYSTEM 1980
 LANGUAGE: RISC-A ASSEMBLER V1.1 VIRTUAL DEVICE 1984
 USE: This program is assembled and then transferred to the
 robot's navigation computer via an RS-232 link (This must
 be preceded by loading the file MAPRS.NAV to the drive
 computer (MENOS) using MTD in Term mode). NAV.HEX can
 be loaded into the nav computer using MTD in terminal (T)
 mode with the following commands: L,1000,1000 followed by
 transmitting the file NAV.HEX. The heading of travel
 is ghead00, ghead01, and ghead11 which are set at assembly
 time and therefore fixed in NAV.HEX.
 CONTENTS: forward
 lock
 clearv
 movev
 getch
 sendt
 sendv
 right
 left
 center
 rctr
 rot
 movev
 stop
 sdelay
 ldelay
 FUNCTION: This program allows the nav computer to request data
 from the GYRA0 (over an RS-232 link), transmit GYRA0
 heading data to an external computer over an RS-232
 link, and steer the robot along a specified heading.

1000	begin	equ	01000H	Starting address of nav program
0001	ghead00	equ	010H	MSB 00 nibble given heading in hex
0002	ghead01	equ	00FH	LSB 00 nibble given heading in hex
0003	ghead11	equ	00FH	MSB 10 nibble given heading in hex
0004	ghead11	equ	00FH	LSB 10 nibble given heading in hex
0005	ghead11	equ	00FH	MSB 11 nibble given heading in hex
0006	ghead11	equ	00FH	LSB 11 nibble given heading in hex
0007	ghead11	equ	00FH	MSB 12 nibble given heading in hex
0008	ghead11	equ	00FH	LSB 12 nibble given heading in hex
0009	ghead11	equ	00FH	MSB 13 nibble given heading in hex
000A	ghead11	equ	00FH	LSB 13 nibble given heading in hex
000B	ghead11	equ	00FH	MSB 14 nibble given heading in hex
000C	ghead11	equ	00FH	LSB 14 nibble given heading in hex
000D	ghead11	equ	00FH	MSB 15 nibble given heading in hex
000E	ghead11	equ	00FH	LSB 15 nibble given heading in hex
000F	ghead11	equ	00FH	MSB 16 nibble given heading in hex
0010	ghead11	equ	00FH	LSB 16 nibble given heading in hex
0011	ghead11	equ	00FH	MSB 17 nibble given heading in hex
0012	ghead11	equ	00FH	LSB 17 nibble given heading in hex
0013	ghead11	equ	00FH	MSB 18 nibble given heading in hex
0014	ghead11	equ	00FH	LSB 18 nibble given heading in hex
0015	ghead11	equ	00FH	MSB 19 nibble given heading in hex
0016	ghead11	equ	00FH	LSB 19 nibble given heading in hex
0017	ghead11	equ	00FH	MSB 20 nibble given heading in hex
0018	ghead11	equ	00FH	LSB 20 nibble given heading in hex
0019	ghead11	equ	00FH	MSB 21 nibble given heading in hex
001A	ghead11	equ	00FH	LSB 21 nibble given heading in hex
001B	ghead11	equ	00FH	MSB 22 nibble given heading in hex
001C	ghead11	equ	00FH	LSB 22 nibble given heading in hex
001D	ghead11	equ	00FH	MSB 23 nibble given heading in hex
001E	ghead11	equ	00FH	LSB 23 nibble given heading in hex
001F	ghead11	equ	00FH	MSB 24 nibble given heading in hex
0020	ghead11	equ	00FH	LSB 24 nibble given heading in hex
0021	ghead11	equ	00FH	MSB 25 nibble given heading in hex
0022	ghead11	equ	00FH	LSB 25 nibble given heading in hex
0023	ghead11	equ	00FH	MSB 26 nibble given heading in hex
0024	ghead11	equ	00FH	LSB 26 nibble given heading in hex
0025	ghead11	equ	00FH	MSB 27 nibble given heading in hex
0026	ghead11	equ	00FH	LSB 27 nibble given heading in hex
0027	ghead11	equ	00FH	MSB 28 nibble given heading in hex
0028	ghead11	equ	00FH	LSB 28 nibble given heading in hex
0029	ghead11	equ	00FH	MSB 29 nibble given heading in hex
002A	ghead11	equ	00FH	LSB 29 nibble given heading in hex
002B	ghead11	equ	00FH	MSB 30 nibble given heading in hex
002C	ghead11	equ	00FH	LSB 30 nibble given heading in hex
002D	ghead11	equ	00FH	MSB 31 nibble given heading in hex
002E	ghead11	equ	00FH	LSB 31 nibble given heading in hex
002F	ghead11	equ	00FH	MSB 32 nibble given heading in hex
0030	ghead11	equ	00FH	LSB 32 nibble given heading in hex
0031	ghead11	equ	00FH	MSB 33 nibble given heading in hex
0032	ghead11	equ	00FH	LSB 33 nibble given heading in hex
0033	ghead11	equ	00FH	MSB 34 nibble given heading in hex
0034	ghead11	equ	00FH	LSB 34 nibble given heading in hex
0035	ghead11	equ	00FH	MSB 35 nibble given heading in hex
0036	ghead11	equ	00FH	LSB 35 nibble given heading in hex
0037	ghead11	equ	00FH	MSB 36 nibble given heading in hex
0038	ghead11	equ	00FH	LSB 36 nibble given heading in hex
0039	ghead11	equ	00FH	MSB 37 nibble given heading in hex
003A	ghead11	equ	00FH	LSB 37 nibble given heading in hex
003B	ghead11	equ	00FH	MSB 38 nibble given heading in hex
003C	ghead11	equ	00FH	LSB 38 nibble given heading in hex
003D	ghead11	equ	00FH	MSB 39 nibble given heading in hex
003E	ghead11	equ	00FH	LSB 39 nibble given heading in hex
003F	ghead11	equ	00FH	MSB 40 nibble given heading in hex
0040	ghead11	equ	00FH	LSB 40 nibble given heading in hex
0041	ghead11	equ	00FH	MSB 41 nibble given heading in hex
0042	ghead11	equ	00FH	LSB 41 nibble given heading in hex
0043	ghead11	equ	00FH	MSB 42 nibble given heading in hex
0044	ghead11	equ	00FH	LSB 42 nibble given heading in hex
0045	ghead11	equ	00FH	MSB 43 nibble given heading in hex
0046	ghead11	equ	00FH	LSB 43 nibble given heading in hex
0047	ghead11	equ	00FH	MSB 44 nibble given heading in hex
0048	ghead11	equ	00FH	LSB 44 nibble given heading in hex
0049	ghead11	equ	00FH	MSB 45 nibble given heading in hex
004A	ghead11	equ	00FH	LSB 45 nibble given heading in hex
004B	ghead11	equ	00FH	MSB 46 nibble given heading in hex
004C	ghead11	equ	00FH	LSB 46 nibble given heading in hex
004D	ghead11	equ	00FH	MSB 47 nibble given heading in hex
004E	ghead11	equ	00FH	LSB 47 nibble given heading in hex
004F	ghead11	equ	00FH	MSB 48 nibble given heading in hex
0050	ghead11	equ	00FH	LSB 48 nibble given heading in hex
0051	ghead11	equ	00FH	MSB 49 nibble given heading in hex
0052	ghead11	equ	00FH	LSB 49 nibble given heading in hex
0053	ghead11	equ	00FH	MSB 50 nibble given heading in hex
0054	ghead11	equ	00FH	LSB 50 nibble given heading in hex
0055	ghead11	equ	00FH	MSB 51 nibble given heading in hex
0056	ghead11	equ	00FH	LSB 51 nibble given heading in hex
0057	ghead11	equ	00FH	MSB 52 nibble given heading in hex
0058	ghead11	equ	00FH	LSB 52 nibble given heading in hex
0059	ghead11	equ	00FH	MSB 53 nibble given heading in hex
005A	ghead11	equ	00FH	LSB 53 nibble given heading in hex
005B	ghead11	equ	00FH	MSB 54 nibble given heading in hex
005C	ghead11	equ	00FH	LSB 54 nibble given heading in hex
005D	ghead11	equ	00FH	MSB 55 nibble given heading in hex
005E	ghead11	equ	00FH	LSB 55 nibble given heading in hex
005F	ghead11	equ	00FH	MSB 56 nibble given heading in hex
0060	ghead11	equ	00FH	LSB 56 nibble given heading in hex
0061	ghead11	equ	00FH	MSB 57 nibble given heading in hex
0062	ghead11	equ	00FH	LSB 57 nibble given heading in hex
0063	ghead11	equ	00FH	MSB 58 nibble given heading in hex
0064	ghead11	equ	00FH	LSB 58 nibble given heading in hex
0065	ghead11	equ	00FH	MSB 59 nibble given heading in hex
0066	ghead11	equ	00FH	LSB 59 nibble given heading in hex
0067	ghead11	equ	00FH	MSB 60 nibble given heading in hex
0068	ghead11	equ	00FH	LSB 60 nibble given heading in hex
0069	ghead11	equ	00FH	MSB 61 nibble given heading in hex
006A	ghead11	equ	00FH	LSB 61 nibble given heading in hex
006B	ghead11	equ	00FH	MSB 62 nibble given heading in hex
006C	ghead11	equ	00FH	LSB 62 nibble given heading in hex
006D	ghead11	equ	00FH	MSB 63 nibble given heading in hex
006E	ghead11	equ	00FH	LSB 63 nibble given heading in hex
006F	ghead11	equ	00FH	MSB 64 nibble given heading in hex
0070	ghead11	equ	00FH	LSB 64 nibble given heading in hex
0071	ghead11	equ	00FH	MSB 65 nibble given heading in hex
0072	ghead11	equ	00FH	LSB 65 nibble given heading in hex
0073	ghead11	equ	00FH	MSB 66 nibble given heading in hex
0074	ghead11	equ	00FH	LSB 66 nibble given heading in hex
0075	ghead11	equ	00FH	MSB 67 nibble given heading in hex
0076	ghead11	equ	00FH	LSB 67 nibble given heading in hex
0077	ghead11	equ	00FH	MSB 68 nibble given heading in hex
0078	ghead11	equ	00FH	LSB 68 nibble given heading in hex
0079	ghead11	equ	00FH	MSB 69 nibble given heading in hex
007A	ghead11	equ	00FH	LSB 69 nibble given heading in hex
007B	ghead11	equ	00FH	MSB 70 nibble given heading in hex
007C	ghead11	equ	00FH	LSB 70 nibble given heading in hex
007D	ghead11	equ	00FH	MSB 71 nibble given heading in hex
007E	ghead11	equ	00FH	LSB 71 nibble given heading in hex
007F	ghead11	equ	00FH	MSB 72 nibble given heading in hex
0080	ghead11	equ	00FH	LSB 72 nibble given heading in hex
0081	ghead11	equ	00FH	MSB 73 nibble given heading in hex
0082	ghead11	equ	00FH	LSB 73 nibble given heading in hex
0083	ghead11	equ	00FH	MSB 74 nibble given heading in hex
0084	ghead11	equ	00FH	LSB 74 nibble given heading in hex
0085	ghead11	equ	00FH	MSB 75 nibble given heading in hex
0086	ghead11	equ	00FH	LSB 75 nibble given heading in hex
0087	ghead11	equ	00FH	MSB 76 nibble given heading in hex
0088	ghead11	equ	00FH	LSB 76 nibble given heading in hex
0089	ghead11	equ	00FH	MSB 77 nibble given heading in hex
008A	ghead11	equ	00FH	LSB 77 nibble given heading in hex
008B	ghead11	equ	00FH	MSB 78 nibble given heading in hex
008C	ghead11	equ	00FH	LSB 78 nibble given heading in hex
008D	ghead11	equ	00FH	MSB 79 nibble given heading in hex
008E	ghead11	equ	00FH	LSB 79 nibble given heading in hex
008F	ghead11	equ	00FH	MSB 80 nibble given heading in hex
0090	ghead11	equ	00FH	LSB 80 nibble given heading in hex
0091	ghead11	equ	00FH	MSB 81 nibble given heading in hex
0092	ghead11	equ	00FH	LSB 81 nibble given heading in hex
0093	ghead11	equ	00FH	MSB 82 nibble given heading in hex
0094	ghead11	equ	00FH	LSB 82 nibble given heading in hex
0095	ghead11	equ	00FH	MSB 83 nibble given heading in hex
0096	ghead11	equ	00FH	LSB 83 nibble given heading in hex
0097	ghead11	equ	00FH	MSB 84 nibble given heading in hex
0098	ghead11	equ	00FH	LSB 84 nibble given heading in hex
0099	ghead11	equ	00FH	MSB 85 nibble given heading in hex
009A	ghead11	equ	00FH	LSB 85 nibble given heading in hex
009B	ghead11	equ	00FH	MSB 86 nibble given heading in hex
009C	ghead11	equ	00FH	LSB 86 nibble given heading in hex
009D	ghead11	equ	00FH	MSB 87 nibble given heading in hex
009E	ghead11	equ	00FH	LSB 87 nibble given heading in hex
009F	ghead11	equ	00FH	MSB 88 nibble given heading in hex
00A0	ghead11	equ	00FH	LSB 88 nibble given heading in hex
00A1	ghead11	equ	00FH	MSB 89 nibble given heading in hex
00A2	ghead11	equ	00FH	LSB 89 nibble given heading in hex
00A3	ghead11	equ	00FH	MSB 90 nibble given heading in hex
00A4	ghead11	equ	00FH	LSB 90 nibble given heading in hex
00A5	ghead11	equ	00FH	MSB 91 nibble given heading in hex
00A6	ghead11	equ	00FH	LSB 91 nibble given heading in hex
00A7	ghead11	equ	00FH	MSB 92 nibble given heading in hex
00A8	ghead11	equ	00FH	LSB 92 nibble given heading in hex
00A9	ghead11	equ	00FH	MSB 93 nibble given heading in hex
00AA	ghead11	equ	00FH	LSB 93 nibble given heading in hex
00AB	ghead11	equ	00FH	MSB 94 nibble given heading in hex
00AC	ghead11	equ	00FH	LSB 94 nibble given heading in hex
00AD	ghead11	equ	00FH	MSB 95 nibble given heading in hex
00AE	ghead11	equ	00FH	LSB 95 nibble given heading in hex
00AF	ghead11	equ	00FH	MSB 96 nibble given heading in hex
00B0	ghead11	equ	00FH	LSB 96 nibble given heading in hex
00B1	ghead11	equ	00FH	MSB 97 nibble given heading in hex
00B2	ghead11	equ	00FH	LSB 97 nibble given heading in hex
00B3	ghead11	equ	00FH	MSB 98 nibble given heading in hex
00B4	ghead11	equ	00FH	LSB 98 nibble given heading in hex
00B5	ghead11	equ	00FH	MSB 99 nibble given heading in hex
00B6	ghead11	equ	00FH	LSB 99 nibble given heading in hex
00B7	ghead11	equ	00FH	MSB 100 nibble given heading in hex
00B8	ghead11	equ	00FH	LSB 100 nibble given heading in hex
00B9	ghead11	equ	00FH	MSB 101 nibble given heading in hex
00BA	ghead11	equ	00FH	LSB 101 nibble given heading in hex
00BB	ghead11	equ	00FH	MSB 102 nibble given heading in hex
00BC	ghead11	equ	00FH	LSB 102 nibble given heading in hex
00BD	ghead11	equ	00FH	MSB 103 nibble given heading in hex
00BE	ghead11	equ	00FH	LSB 103 nibble given heading in hex
00BF	ghead11	equ	00FH	MSB 104 nibble given heading in hex
00C0	ghead11	equ	00FH	LSB 104 nibble given heading in hex
00C1	ghead11	equ	00FH	MSB 105 nibble given heading in hex
00C2	ghead11	equ	00FH	LSB 105 nibble given heading in hex
00C3				

NAV.A PROGRAM LISTING (continued)

1000	aciatstat	equ	00000000	;ACIA serial port status & control reg
1001	aciatdata	equ	00000000	;ACIA serial port data register
1002	aciatstat	equ	00000000	;MENUB serial port status & control reg
1003	aciatdata	equ	00000000	;MENUB serial port data register
1004	aciatstat	equ	00000000	;TERMINAL serial port status & control reg
1005	aciatdata	equ	00000000	;TERMINAL serial port data register
1006	tdre	equ	00000000	;ACIA transmit data register (0000)
1007	rdre	equ	00000000	;ACIA receive data register (0000)
1008	rdre	equ	00000000	;MEO mask for heading data
1009				
1010	00 10 40 main	lfr	begin	;Starting Address of nav program
1011	00 10 40	lfr	look	;Rotate robot until facing given head
1012	00 10 40	lfr	forward	;Move forward along given heading
1013	00 10 40	lfr	main	

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NAV.A PROGRAM LISTING (continued)

```

DATE: 10/12/65
VERSION: 1.1
NAME: forward
MODULE NUMBER: 1.1
DESCRIPTION: This module will steer the robot along a given
            heading (ghead000, ghead01, & ghead1).
PASSED VARIABLES: None
RETURNING: None
GLOBAL VARIABLES USED: ghead0, ghead1
GLOBAL VARIABLES CHANGED: ghead0, ghead1
REGISTERS USED: A, D
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: clearv
                center
                movef
                beth
                sendt
                right
                left
CALLING MODULES: main
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: None

```

```

1003 00 10 A2 forward jsr clearv      ;Clear OVRAC velocity constant
1004 00 14 05         jsr center      ;Center the steering wheel
1005 00 10 AF         jsr movef       ;Start drive motor fast speed

1011 00 10 E5 forward jsr beth        ;Get OVRAC heading
1012 00 11 0F         jsr sendt      ;Send heading to external computer
1013 00 13 03         ldaa ghead0    ;Get MSB of current heading
1014 01 00           cmpa #ghead0m   ;If given & current heading are equal
1015 01 07           beq 1100m        ;Then test LSB MS nibble of heading
1016 01 10           bcl left        ;Else if current & given head turn left
1017 00 11 10         jsr right      ;Else turn right
1018 00 00           bra forward     ;Go get next heading update

1025 01 13 07 loutbm ldaa ghead1     ;Get LSB of current heading
1026 04           lsr               ;Use only the MS nibble
1027 04           lsr
1028 04           lsr
1029 04           lsr
1030 01 1E           cmpa #ghead1    ;If current & given test LSB MS of head
1031 01 07           beq 1100m        ;Else if current & given head turn left
1032 01 0A           bcl left        ;Else turn right
1033 00 11 10         jsr right      ;Else turn right
1034 00 00           bra forward     ;Go get next heading update

1041 00 10 A2 loutb jsr center        ;Center steering wheel
1042 01 05           bra forward

1049 01 11 17 left jsr left         ;Turn left
1050 01 01           bra forward     ;Go get next heading update
1051 01 00           bra forward

```

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NAV.A PROGRAM LISTING (continued)

```

NAME: 00000005
VERSION: 1.0
NAME: clearv
FILE NUMBER: 1.1
DESCRIPTION: This module clears the GPR velocity constants.
RAISED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, D
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: forward
AUTHOR: CAPT WILLIAM C. RAMEY JR.
HISTORY: None

```

```

1000 00 47      clearv      1000 #0000      :Get GPRAC clear vel command
1004 04 01 10 00v      1004 acia1stat      :Get GPRAC ACIA status
1007 04 01      acia #0000      :Is transmit data reg empty
1009 17 00      seq 00v      :If no rescheck status
1010 07 00 11      stop acia1data      :Else send GPRAC clear vel command
1015 00      rts

```

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 PAGES

NAV.A PROGRAM LISTING (continued)

```

ENTER: 00000000
LEADING: 100
NAME: None
NUMBER OF MODES: 100
DESCRIPTION: This file sends A drive motor fast forward sequence to the drive controller. MENDOS.
ORIGINAL_VARIABLES: None
ORIGINAL_VARIABLES USED: None
ORIGINAL_VARIABLES CHANGED: None
REGISTERED USER: J. B
FILE READ: None
FILES WRITTEN: None
MODES CALLED: fast, delay, delay
CALLING MODES: forward
AUTHOR: CAPT WILLIAM L. RANNEY JR.
LISTED : None

```

```

1007 10 41 movef      load # 1          ;Send to drive computer (MEMO)
1008 00 10 00      sr send          ;start drive motor fast cont command
1009 00 10 40      sr sdelay         ;sequence

1007 10 44         load # 0
1008 00 10 00      sr send
1009 00 10 40      sr sdelay

1007 00 00         load # 0
1008 00 10 00      sr send
1009 00 10 40      sr sdelay

1007 10 01         load # 1
1008 10 10 10      sr send
1009 00 10 40      sr sdelay

1007 10 10         load # 0
1008 10 10 10      sr send
1009 00 10 40      sr sdelay

1007 10 00         load # 0
1008 00 10 00      sr send
1009 00 10 40      sr sdelay

1007 10 00         load # 0
1008 00 10 00      sr send
1009 00 10 40      sr sdelay

1007 10 10 00      sr sdelay
1008 10 10 00      sr sdelay
1009 00 00         rcs

```

NAV.A PROGRAM LISTING (continued)

```

DATE: 05/04/86
VERSION: 1.0
NAME: send
MODULE NUMBER: 1.1.1
DESCRIPTION: This module sends a type of data to the drive
              and then MENOS over an RS-232 link.
PARAMETER VARIABLES: Output data (passed in C reg)
RETURN: None
LOCAL VARIABLES USED: None
LOCAL VARIABLES CHANGED: None
REGISTERS USED: A, C
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: None, right, left, center, rotate, stop
AUTHOR: CAPT WILLIAM L. RAMEY JR.
EDITOR: None
    
```

```

1100 01 00 00 send 10aa acia stat ;Get MENOS ACIA status
1101 04 00      acia #dore ;Is transmit data reg empty
1102 07 00      reg send ;If no recheck status
1103 09 00 00      stat acia data ;Else send data to MENOS
1104 0A 00      rts
    
```

NAV.A PROGRAM LISTING (continued)

```

DATE: 08/11/05
VERSION: 1.0
NAME: sdelay
MODULE NUMBER: 1.0.1
DESCRIPTION: This module provides for a short fixed time delay
PASTED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, D
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: Move, Right, Left, Center, Rotate, Stop
AUTHOR: CAPT WILLIAM C. FAME, JR.
HISTORY: None

```

```

1040 17      sdelay  tshb      ;Save registers
1041 18      psha
1042 19      ldsa #16H      ;Execute fixed time delay
1043 0A FF      sd1      ldsb #16FH
1044 01      sd2      bnp
1045 01      sd2      dsb
1046 0A      bne sd2
1047 0A      dsca
1048 0A      bne sd1
1049 0A      jsra      ;Restore registers
1050 0A      dsrb
1051 0A      rts

```

NAV.A PROGRAM LISTING (continued)

```

.....
DATE: 08/11/87
VERSION: 1.0
NAME: delay
MODULE NUMBER: 1.0.0
DESCRIPTION: This module provides a fixed long time delay
POISED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: None
LOCAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: move, right, left, center, rotate, stop, lock
AUTHOR: DART WILLIAM J. RAMEY JR.
HISTORY: None
.....

0040 07      delay      pshb          ;Save registers
0041 08          psha
0042 09 09      lsl      lsl      ;Execute time delay
0043 10 FF      ldi      lsl      ;Execute time delay
0044 11          pop
0045 12          lslb
0046 13 00      tne      lsl
0047 14 00      decb
0048 15 00      tne      lsl
0049 16          popa
0050 17          pushb
0051 18          rts

0052          break      rts 1
0053          break      rts 1
0054          END

```

NAV.A PROGRAM LISTING (continued)

```

*****
INITIAL: 1000
END: 1000
COUNTERS: 1000
DESCRIPTION: This routine reads 10 bytes of heading data from
the GPRAC and stores it in global variables
(heading and check).
ALIASED VARIABLES: None
SET ENV: None
LOCAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: heading, check
REGISTERS USED: R1, R2
FILES READ: None
FILES WRITTEN: None
ROUTINES CALLED: None
CALLING ROUTINES: Forward, Back
AUTHOR: CAPT WILLIAM D. ARNE, JR.
HISTORY: None
*****

```

```

1000 00 40 00 peth 10ab #0000 ;Get GPRAC 2 byte heading command
1001 00 00 10 01 10aa acia1stat ;Get GPRAC ACIA status
1002 04 00 00 01 00aa #0000 ;Is transmit data reg empty
1003 07 00 00 01 00aa #0000 ;If no recheck status
1004 07 00 11 00aa acia1data ;Else send GPRAC command

1005 00 00 10 02 10aa acia1stat ;Get GPRAC ACIA status
1006 04 01 00 01 00aa #0000 ;Is receive data register full?
1007 07 00 00 01 00aa #0000 ;If no recheck status
1008 04 00 11 00aa acia1data ;Else get the MIB of heading data
1009 04 00 00 00 00aa #0000 ;Mask out MIB 4 bits
1010 07 13 00 00aa check ;Save current MIB heading

1011 00 00 10 03 10aa acia1stat ;Get GPRAC ACIA status
1012 04 01 00 01 00aa #0000 ;Is receive data register full?
1013 07 00 00 01 00aa #0000 ;If no recheck status
1014 04 00 11 00aa acia1data ;Else get the LID of heading data
1015 07 11 00 00aa check ;Save current LID of heading
1016 00 00 00 00 0000 #0000 ;

```


NAV.A PROGRAM LISTING (continued)

```
DATE: 10/26/10
VERSION: 1.0
NAME: send
MODULE NUMBER: 1.4
DESCRIPTION: This module sends two bytes of reading data
             represented as four ASCII hex characters to an
             external terminal.
PASSED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: cheadn, cheadl
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: None
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: None
CALLING MODULES: forward, lock
AUTHOR: CAPT WILLIAM G. RANEY JR.
HISTORY: None
```

1110	03 10 00	sendt	ldab cheadn	:Get MSD of current heading
1111	04		lsrb	:Convert upper nibble of MSD to ASCII
1112	04		lsrb	
1113	04		lsrb	
1114	04 0F		andb #0FH	
1115	00 00		adcb #00H	
1116	00 00		adcb #00H	
1117	00 00		adcb #00H	
1118	00 00		adcb #00H	
1119	00 00		adcb #00H	
1120	00 00	0 s1	ldaa adatastat	:Get terminal ASCII status
1121	00 00		anda #00H	:Is transmit data reg empty
1122	00 00		bcc s1	:If no recheck status
1123	00 00	s1	staa adata	:Else send data
1124	03 10 00		ldab cheadn	:Get MSD of current heading
1125	04		andb #0FH	:Convert upper nibble of MSD to ASCII
1126	04		lsrb	
1127	04		lsrb	
1128	04		lsrb	
1129	04 0F		andb #0FH	
1130	00 00		adcb #00H	
1131	00 00		adcb #00H	
1132	00 00		adcb #00H	
1133	00 00		adcb #00H	
1134	00 00	0 s1	ldaa adatastat	:Get terminal ASCII status
1135	00 00		anda #00H	:Is transmit data reg empty
1136	00 00		bcc s1	:If no recheck status
1137	00 00	s1	staa adata	:Else send data
1138	03 10 00		ldab cheadn	:Get MSD of current heading
1139	04		lsrb	:Convert upper nibble of MSD to ASCII
1140	04		lsrb	
1141	04		lsrb	
1142	04		lsrb	
1143	04 0F		andb #0FH	
1144	00 00		adcb #00H	
1145	00 00		adcb #00H	
1146	00 00		adcb #00H	
1147	00 00		adcb #00H	
1148	00 00	0 s1	ldaa adatastat	:Get terminal ASCII status
1149	00 00		anda #00H	:Is transmit data reg empty
1150	00 00		bcc s1	:If no recheck status
1151	00 00	s1	staa adata	:Else send data

NAV.A PROGRAM LISTING (continued)

```

1150  P 10 45      ldat #0001      ;Get LID of current heading
1151  Q4 15        addr #01H      ;Convert lower nibble of LID to ASCII
1152  Q4 16        addr #00H
1153  Q4 17        addr #01H
1154  Q4 18        btr  s4
1155  Q4 19        addr #07H
1156  Q4 20 00 45  ldaa astatstat ;Get terminal ASCII status
1157  Q4 21 00 45  anda #tdre      ;Is transmit data reg empty
1158  Q4 22 00 45  btr  s4         ;If no receive status
1159  Q4 23 00 45  staa astatdata ;Else send data

1171  Q4 10        ldat #000H     ;Get carriage return
1172  Q4 11 00 45  ldaa astatstat ;Get terminal ASCII status
1173  Q4 12 00 45  anda #00H      ;Is transmit data reg empty
1174  Q4 13 00 45  btr  s4         ;If no receive status
1175  Q4 14 00 45  staa astatdata ;Else send data

1177  Q4 00        ldat #0AH      ;Get line feed
1178  Q4 01 00 45  ldaa astatstat ;Get terminal ASCII status
1179  Q4 02 00 45  anda #tdre      ;Is transmit data reg empty
1180  Q4 03 00 45  btr  s4         ;If no receive status
1181  Q4 04 00 45  staa astatdata ;Else send data

1182  Q4 20        rts

```

NAV.A PROGRAM LISTING (continued)

```

DATE: 08/14/55
VERSION: 1.0
NAME: right
MODULE NUMBER: 1.5
DESCRIPTION: This module sends a right turn relative mode
             command sequence to the drive computer (MENDS).
PASSED VARIABLES: None
RETURNS: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: delay, delay, send
CALLING MODULES: forward
AUTHOR: CAPT WILLIAM J. RAMEY JR.
HISTORY: None
    
```

```

1131 06 41      right      ldab # A      ;Send to drive computer (MENDS)
1132 00 13 00      jsr sendx      right turn continuous command
1133 00 13 48      jsr delay      ;sequence

1134 06 44      ldab # D
1135 00 13 00      jsr sendx
1136 00 13 48      jsr delay

1137 06 36      ldab # B
1138 00 13 00      jsr sendx
1139 00 13 48      jsr delay

1140 06 31      ldab # C
1141 00 13 00      jsr sendx
1142 00 13 48      jsr delay

1143 06 26      ldab # A
1144 00 13 00      jsr sendx
1145 00 13 48      jsr delay

1146 06 13      ldab # 0
1147 00 13 00      jsr sendx
1148 00 13 48      jsr delay

1149 06 11      ldab # 0
1150 00 13 00      jsr sendx
1151 00 13 48      jsr delay

1152 06 08      ldab # 0
1153 00 13 00      jsr sendx
1154 00 13 48      jsr delay

1155 06 05      ldab # 0
1156 00 13 00      jsr sendx
1157 00 13 48      jsr delay
    
```

NAV.A PROGRAM LISTING (continued)

```

DATE: 10/11/75
VERSION: 1.0
NAME: Left
MODULE NUMBER: 1.3
DESCRIPTION: This module sends a left turn relative mode
              command sequence to the driver computer (MENDS)
RAISED VARIABLES: None
RETURN: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: delay, delay, send
CALLING MODULES: forward
AUTHOR: CAPT WILLIAM D. RAMEY JR.
HISTORY: None

```

1109	16 41	Left	ldab #1A	!Send to drive computer (MENDS)
1111	20 13 30		jsr send	!left turn continuous command
110E	20 13 40		jsr delay	!sequence
1101	16 44		ldab #10	
1103	20 13 10		jsr send	
1101	20 13 40		jsr delay	
1107	06 50		ldab #6	
110D	20 13 30		jsr send	
110E	20 13 40		jsr delay	
1101	01 11		ldab #11	
1101	20 13 10		jsr send	
1101	20 13 40		jsr delay	
1107	01 10		ldab #10	
1101	20 13 10		jsr send	
110E	20 13 40		jsr delay	
1107	01 14		ldab #14	
1101	20 13 10		jsr send	
1101	20 13 40		jsr delay	
1107	01 12 50		jsr delay	
1107	01 13 10		jsr delay	
1107	01 13 50		jsr delay	
1107	01 13 50		jsr delay	

NAV.A PROGRAM LISTING (continued)

```

DATE: 10/1/65
VERSION: 1.1
NAME: center
MODULE NUMBER: 117
DESCRIPTION: This module sends a center steering motor
              drive sequence to the drive computer (MEM01).
PARAMETER VARIABLES: None
RETURNS: None
GLOBAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, D
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: send, sdelay, delay
CALLING MODULES: forward, back
AUTHOR: CAPT WILLIAM W. RAMELBERG
EDITOR: None
    
```

```

1205 00 41 center 1dat #A1 ;Send to drive computer (MEM01)
1206 00 13 00 ;sr send ;center steering motor command
1207 00 13 48 ;sr sdelay ;sequence

1215 00 44 1dab #D1
1216 00 13 00 ;sr send
1217 00 13 48 ;sr sdelay

1225 00 40 1dab #B1
1226 00 13 00 ;sr send
1227 00 13 48 ;sr sdelay

1235 00 41 1dab #F1
1236 00 13 00 ;sr send
1237 00 13 48 ;sr sdelay

1245 00 40 1dab #A1
1246 00 13 00 ;sr send
1247 00 13 48 ;sr sdelay

1255 00 40 1dab #E1
1256 00 13 00 ;sr send
1257 00 13 48 ;sr sdelay

1265 00 10 58 ;sr delay
1266 00 10 58 ;sr delay

1275 00 10 58 ;sr delay
1276 00 10 58 ;sr delay
1277 00 10 58 ;sr delay
    
```

NAV.A PROGRAM LISTING (continued)

```

DATE: 1981-05-15
TIME: 11:11
NAME: Bob
CIRCLE NUMBER: 100
DESCRIPTION: This circle will issue the command sequence to
cause the steering drive to move so the robot
will rotate right when the drive ratio is turned
on.
PAIRED VARIABLES: None
RETURN: None
LOCAL VARIABLES USED: None
LOCAL VARIABLES CHANGED: None
REGISTERS USED: A, B
PUSH READ: None
PUSH WRITE: None
CIRCLES CALLED: send, delay, delay,
CALLING CIRCLES: None
NOTES: CAPT WILLIAM J. RAMEY JR.
EDITOR: None

```

1241	00 10 41	rcvr	label # 20	send to drive computer (MENDS)
1242	00 10 40		len send	turn on tag right steering motor
1243	00 10 40		len delay	command sequence
1244	00 10 44		label # 21	
1245	00 10 40		len send	
1246	00 10 40		len delay	
1247	00 10 40		label # 22	
1248	00 10 40		len send	
1249	00 10 40		len delay	
1250	00 10 40		label # 23	
1251	00 10 40		len send	
1252	00 10 40		len delay	
1253	00 10 40		label # 24	
1254	00 10 40		len send	
1255	00 10 40		len delay	
1256	00 10 40		label # 25	
1257	00 10 40		len send	
1258	00 10 40		len delay	
1259	00 10 40		label # 26	
1260	00 10 40		len send	
1261	00 10 40		len delay	
1262	00 10 40		label # 27	
1263	00 10 40		len send	
1264	00 10 40		len delay	
1265	00 10 40		label # 28	
1266	00 10 40		len send	
1267	00 10 40		len delay	
1268	00 10 40		label # 29	
1269	00 10 40		len send	
1270	00 10 40		len delay	
1271	00 10 40		label # 30	
1272	00 10 40		len send	
1273	00 10 40		len delay	
1274	00 10 40		label # 31	
1275	00 10 40		len send	
1276	00 10 40		len delay	
1277	00 10 40		label # 32	
1278	00 10 40		len send	
1279	00 10 40		len delay	
1280	00 10 40		label # 33	
1281	00 10 40		len send	
1282	00 10 40		len delay	
1283	00 10 40		label # 34	
1284	00 10 40		len send	
1285	00 10 40		len delay	
1286	00 10 40		label # 35	
1287	00 10 40		len send	
1288	00 10 40		len delay	
1289	00 10 40		label # 36	
1290	00 10 40		len send	
1291	00 10 40		len delay	
1292	00 10 40		label # 37	
1293	00 10 40		len send	
1294	00 10 40		len delay	
1295	00 10 40		label # 38	
1296	00 10 40		len send	
1297	00 10 40		len delay	
1298	00 10 40		label # 39	
1299	00 10 40		len send	
1300	00 10 40		len delay	

NAV.A PROGRAM LISTING (continued)

```

: DATE: 04-02-85
: TIME: 11:04
: NAME: test
: MODULE NUMBER: 1.0
: DESCRIPTION: This module will issue the command sequence to
:             raise the steering wheel to give to the robot
:             will rotate left when the drive motor is turned
:             on.
:
: GLOBAL VARIABLES: None
: RETURN: None
: GLOBAL VARIABLES USED: None
: GLOBAL VARIABLES CHANGED: None
: REGISTERS USED: A, B
: FILES READ: None
: FILES WRITTEN: None
: MODULES CALLED: send, delay, delay
: CALLING MODULES: 1001
: AUTHOR: CAPT WILLIAM J. RAMEY JR.
: HISTORY: None

```

1239	00 41	net1	load #A1	;Send to drive computer (MEMO3)
1240	00 10 10		usr send	turn 90 deg left steering wheel
1241	00 10 40		usr delay	;command sequence
1242	00 44		load #C1	
1243	00 10 10		usr send	
1244	00 10 40		usr delay	
1245	00 10		load #10	
1246	00 10 10		usr send	
1247	00 10 40		usr delay	
1248	00 11		load #11	
1249	00 10 10		usr send	
1250	00 10 40		usr delay	
1251	00 11		load #1	
1252	00 10 10		usr send	
1253	00 10 40		usr delay	
1254	00 4		load #F	
1255	00 10 10		usr send	
1256	00 10 40		usr delay	
1257	00 11 00		usr delay	
1258	00 11 00		usr delay	
1259	00 11 00		usr delay	
1260	00 11 00		usr delay	
1261	00 11 00		usr delay	
1262	00 11 00		usr delay	
1263	00 11 00		usr delay	
1264	00 11 00		usr delay	
1265	00 11 00		usr delay	
1266	00 11 00		usr delay	
1267	00 11 00		usr delay	
1268	00 11 00		usr delay	
1269	00 11 00		usr delay	
1270	00 11 00		usr delay	
1271	00 11 00		usr delay	
1272	00 11 00		usr delay	
1273	00 11 00		usr delay	
1274	00 11 00		usr delay	
1275	00 11 00		usr delay	
1276	00 11 00		usr delay	
1277	00 11 00		usr delay	
1278	00 11 00		usr delay	
1279	00 11 00		usr delay	
1280	00 11 00		usr delay	
1281	00 11 00		usr delay	
1282	00 11 00		usr delay	
1283	00 11 00		usr delay	
1284	00 11 00		usr delay	
1285	00 11 00		usr delay	
1286	00 11 00		usr delay	
1287	00 11 00		usr delay	
1288	00 11 00		usr delay	
1289	00 11 00		usr delay	
1290	00 11 00		usr delay	
1291	00 11 00		usr delay	
1292	00 11 00		usr delay	
1293	00 11 00		usr delay	
1294	00 11 00		usr delay	
1295	00 11 00		usr delay	
1296	00 11 00		usr delay	
1297	00 11 00		usr delay	
1298	00 11 00		usr delay	
1299	00 11 00		usr delay	
1300	00 11 00		usr delay	
1301	00 11 00		usr delay	
1302	00 11 00		usr delay	
1303	00 11 00		usr delay	
1304	00 11 00		usr delay	
1305	00 11 00		usr delay	
1306	00 11 00		usr delay	
1307	00 11 00		usr delay	
1308	00 11 00		usr delay	
1309	00 11 00		usr delay	
1310	00 11 00		usr delay	
1311	00 11 00		usr delay	
1312	00 11 00		usr delay	
1313	00 11 00		usr delay	
1314	00 11 00		usr delay	
1315	00 11 00		usr delay	
1316	00 11 00		usr delay	
1317	00 11 00		usr delay	
1318	00 11 00		usr delay	
1319	00 11 00		usr delay	
1320	00 11 00		usr delay	
1321	00 11 00		usr delay	

NAV.A PROGRAM LISTING (continued)

```

DATE: 0010 85
VERSION: 1.0
NAME: 1.0
MODULE NUMBER: 1.0
DESCRIPTION: This module will issue the command sequence to
             cause the drive motor to turn on at 90 deg speed.
CALLED VARIABLES: None
RETURNS: None
LOCAL VARIABLES USED: None
GLOBAL VARIABLES CHANGED: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: send, delay, delay
CALLING MODULES: 1.0
AUTHOR: CAPT WILLIAM D. RAMEY JR.
HISTORY: None
    
```

```

1007 00 40 00 movem load # 01 ;Send to drive computer (MENGE)
1008 00 10 00 jsr send ;turn 90 deg steering motor command
1009 00 10 40 jsr delay ;sequence

1007 00 44 00 load # 01
1008 00 10 00 jsr send
1009 00 10 40 jsr delay

1007 00 20 00 load # 01
1008 00 10 00 jsr send
1009 00 10 40 jsr delay

1007 00 01 00 load # 1
1008 00 10 00 jsr send
1009 00 10 40 jsr delay

1007 00 01 00 load # 1
1008 00 10 00 jsr send
1009 00 10 40 jsr delay

1007 00 40 00 load # 01
1008 00 10 00 jsr send
1009 00 10 40 jsr delay

1007 00 10 00 jsr delay
1008 00 10 00 jsr delay
1009 00 10 00 rts
    
```


NAV.A PROGRAM LISTING (continued)

```

DATE: 07 11 15
VERSION: 1.1
NAME: stop
MODULE NUMBER: 1.11
DESCRIPTION: This module sends the command sequence to stop
the drive and steering motors.
PASSED VARIABLES: None
RETURNS: None
GLOBAL VARIABLES USED: None
LOCAL VARIABLES IN scope: None
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: send, sdelay, delay
CALLING MODULES: lock
AUTHOR: CAPT WILLIAM D. RAMEY JR.
HISTORY: None

```

```

1006 06 41 stop      1tab #1A1      1send to drive computer (MENOS)
1008 00 13 00      1sr send      1stop steering and drive motor command
1009 00 13 40      1sr sdelay     1sequence
1010 06 44          1tab #01
1011 00 13 00      1sr send
1013 00 13 40      1sr sdelay
1014 06 38          1tab #0
1015 00 13 00      1sr send
1016 00 10 40      1sr sdelay
1017 06 31          1tab #11
1018 00 13 00      1sr send
1019 00 13 40      1sr sdelay
1020 06 21          1tab #11
1021 00 13 00      1sr send
1022 00 13 40      1sr sdelay
1023 06 10          1tab #01
1024 00 13 00      1sr send
1025 00 13 40      1sr sdelay
1026 06 11 50      1sr delay
1027 00 11 50      1sr delay
1028 00          rts

```

NAV.A PROGRAM LISTING (continued)

```

DATE: 12/12/66
VERSION: 1.0
NAME: 1001
MODULE NUMBER: 1.0
DESCRIPTION: This module will initiate the robot until it is
              within 1.5 degrees of the given heading.
PASSED VARIABLES: None
RETURNS: None
LOCAL VARIABLES USED: cheadn, cheadl
LOCAL VARIABLES CHANGED: cheadn, cheadl
REGISTERS USED: A, B
FILES READ: None
FILES WRITTEN: None
MODULES CALLED: center
                 moveh
                 geth
                 sendt
                 rctt
                 rctf
                 stop
                 delay
CALLING MODULES: main
AUTHOR: CAPT WILLIAM G. RAMEY JR.
DISTRIB: None

```

```

1042  00 17 00 1001  jsr  geth      ;Get G FAC heading
1043  00 11 07      jsr  sendt     ;Send heading to external computer
1044  00 10 00      ldaa  cheadn    ;Get MSB of current heading
1045  01 03      cmpa  #theadm     ;If given & current heading are equal
1046  07 07      beq   1044        ;Then test LSB MSB of heading
1047  00 17      bmi  1045        ;Else if current < given Go left
1048  00 11 40     jsr  rctt      ;Else move steering wheel 40 right
1049  00 17      bra  1045

1050  00 10 07 1001  ldaa  cheadl    ;Get LSB of current heading
1051  01 04      lsra             ;Use only the MS nibble
1052  01 04      lsra
1053  01 04      lsra
1054  01 03      cmpa  #theadl     ;If current heading = given heading
1055  07 07      beq   1044        ;Then then move forward
1056  00 17      bmi  1045        ;Else if current < given head Go left
1057  00 11 40     jsr  rctf      ;Else move steering wheel 40 left
1058  00 17      bra  1045

1059  00 10 00 1001  jsr  rctf      ;Move steering wheel 40 left
1060  00 17      bra  1045

1061  00 10 00 1001  jsr  moveh     ;Start move with no speed
1062  01 07 00 1001  jsr  geth      ;Get G FAC heading
1063  01 07 00 1001  ldaa  cheadn    ;Get MSB of current heading
1064  01 03      cmpa  #theadm     ;If given & current heading are equal
1065  07 07      beq   1061        ;Then test LSB MSB of heading
1066  00 17      bmi  1062        ;Else go get next heading update

```

NAV.A PROGRAM LISTING (continued)

[illegible]

NAV.A PROGRAM LISTING (continued)

alpha1	1001	alpha sta	1001	alpha1 sat	1001	alpha1 data	1001
alpha2	1002	alpha sta	1002	alpha2 sat	1002	alpha2 data	1002
beta1	1003	beta sta	1003	beta1 sat	1003	beta1 data	1003
beta2	1004	beta sta	1004	beta2 sat	1004	beta2 data	1004
gamma1	1005	gamma sta	1005	gamma1 sat	1005	gamma1 data	1005
gamma2	1006	gamma sta	1006	gamma2 sat	1006	gamma2 data	1006
delta1	1007	delta sta	1007	delta1 sat	1007	delta1 data	1007
delta2	1008	delta sta	1008	delta2 sat	1008	delta2 data	1008
epsilon1	1009	epsilon sta	1009	epsilon1 sat	1009	epsilon1 data	1009
epsilon2	1010	epsilon sta	1010	epsilon2 sat	1010	epsilon2 data	1010
zeta1	1011	zeta sta	1011	zeta1 sat	1011	zeta1 data	1011
zeta2	1012	zeta sta	1012	zeta2 sat	1012	zeta2 data	1012
eta1	1013	eta sta	1013	eta1 sat	1013	eta1 data	1013
eta2	1014	eta sta	1014	eta2 sat	1014	eta2 data	1014
theta1	1015	theta sta	1015	theta1 sat	1015	theta1 data	1015
theta2	1016	theta sta	1016	theta2 sat	1016	theta2 data	1016
iota1	1017	iota sta	1017	iota1 sat	1017	iota1 data	1017
iota2	1018	iota sta	1018	iota2 sat	1018	iota2 data	1018
kappa1	1019	kappa sta	1019	kappa1 sat	1019	kappa1 data	1019
kappa2	1020	kappa sta	1020	kappa2 sat	1020	kappa2 data	1020
lambda1	1021	lambda sta	1021	lambda1 sat	1021	lambda1 data	1021
lambda2	1022	lambda sta	1022	lambda2 sat	1022	lambda2 data	1022
mu1	1023	mu sta	1023	mu1 sat	1023	mu1 data	1023
mu2	1024	mu sta	1024	mu2 sat	1024	mu2 data	1024
nu1	1025	nu sta	1025	nu1 sat	1025	nu1 data	1025
nu2	1026	nu sta	1026	nu2 sat	1026	nu2 data	1026
xi1	1027	xi sta	1027	xi1 sat	1027	xi1 data	1027
xi2	1028	xi sta	1028	xi2 sat	1028	xi2 data	1028
omicron1	1029	omicron sta	1029	omicron1 sat	1029	omicron1 data	1029
omicron2	1030	omicron sta	1030	omicron2 sat	1030	omicron2 data	1030
pi1	1031	pi sta	1031	pi1 sat	1031	pi1 data	1031
pi2	1032	pi sta	1032	pi2 sat	1032	pi2 data	1032
rho1	1033	rho sta	1033	rho1 sat	1033	rho1 data	1033
rho2	1034	rho sta	1034	rho2 sat	1034	rho2 data	1034
sigma1	1035	sigma sta	1035	sigma1 sat	1035	sigma1 data	1035
sigma2	1036	sigma sta	1036	sigma2 sat	1036	sigma2 data	1036
tau1	1037	tau sta	1037	tau1 sat	1037	tau1 data	1037
tau2	1038	tau sta	1038	tau2 sat	1038	tau2 data	1038
upsilon1	1039	upsilon sta	1039	upsilon1 sat	1039	upsilon1 data	1039
upsilon2	1040	upsilon sta	1040	upsilon2 sat	1040	upsilon2 data	1040
phi1	1041	phi sta	1041	phi1 sat	1041	phi1 data	1041
phi2	1042	phi sta	1042	phi2 sat	1042	phi2 data	1042
chi1	1043	chi sta	1043	chi1 sat	1043	chi1 data	1043
chi2	1044	chi sta	1044	chi2 sat	1044	chi2 data	1044
psi1	1045	psi sta	1045	psi1 sat	1045	psi1 data	1045
psi2	1046	psi sta	1046	psi2 sat	1046	psi2 data	1046
omega1	1047	omega sta	1047	omega1 sat	1047	omega1 data	1047
omega2	1048	omega sta	1048	omega2 sat	1048	omega2 data	1048

MARRS.NAV PROGRAM LISTING

Address (HEX)	Instruction Code (HEX)	Comment
NONE	AA	Put MARRS-1 into
NONE	01	Instruction Input
NONE	00	Mode at Address 0100
0000	CC	Start Drive Motor
0101	1B	Forward at Fast
0102	FF	Speed (movef)
0103	3A	
0004	03	Turn Steering Motor
0105	DC	One Increment to the
0106	EC	Left (left)
0107	01	
0108	3A	
0009	03	Turn Steering Motor
010A	DC	One Increment to the
010B	E8	RIGHT (right)
010C	01	
010D	3A	
010E	03	Turn Steering Motor
010F	CC	to the Center,
0110	E8	straight, Position
0111	49	(center)
0112	3A	
0113	03	Turn Steering Motor
0114	CC	to the Full Right
0115	E8	Position (rotr)
0116	90	
0117	3A	
0118	02	Stop all Motor
0119	03	Movement (stop)
011A	3A	
010B	CC	Start Drive Motor
010C	13	Forward at Medium
010D	FF	Speed (movem)
010E	3A	
011F	03	Turn Steering Motor
0120	CC	to the Full Left
0121	EC	Position (rotl)
0122	00	
0123	3A	
NONE	R	Reset to Input Mode

NAV.A OPERATING INSTRUCTIONS

STEP 1: Power up the H89 computer system. Place the System disk in Drive A and the gyro program disk in Drive B. Boot the system (type "B 29") and change the mode of Drive B to single sided double density (type "mode B:ss,dd"). Change the working drive to Drive B (type "B:").

STEP 2: Connect the H89 to Robot RS-232 cable between the H89 DCE connector and the MARRS-1 Drive Computer (MENOS) connector. Power up the MARRS-1 Robot. Make sure the batteries are fully charged and the charger power line is connected and turned on. Press both the system reset key on the keypad and the Nav computer reset button.

STEP 3: Load and run the M72 modem program on the H89 by typing "M72". When this program is running type "SPD" to change the transmission time delays. When prompted for time delays reply with a "1" for both character and line delay times. Set the H89 keyboard caps lock on.

STEP 4: Load and transmit the MARRS.NAV file to the robot's Drive computer. This is done by typing "T" to enter the M72 Terminal mode. Next type "control shift " followed by "control T" and then "MARRS.NAV" to send the program file. When asked if time delays are desired, answer Yes. The file will be displayed as it is being transmitted. When it is finished you will see the data stop and hear the robot say "READY". Type "control shift " followed by "control T" to return to the M72 command mode. This entire step can be skipped if the program is hand keyed directly into MARRS-1 via the onboard keypad (which is the recommended way since it avoids moving cables). NOTE: MARRS-1 system resets do not erase this program.

STEP 5: Connect the H89 to Robot RS-232 cable between the H89 DCE connector and the MARRS-1 Nav T connector. Connect the Drive Computer to Nav Computer RS-232 cable between the Drive Computer connector and the Nav X connector on MARRS-1. Connect the GYRAC to Nav Computer RS-232 cable between the GYRAC connector and the Nav L connector on MARRS-1. Connect the external power cable to the GYRAC and turn on the power supplies. Flip the GYRAC power switch to the on position and press the GYRAC computer reset button. Flip the gyro control switch to the slaved mode.

STEP 6: Load and transmit the NAV.HEX file to the robot's navigation computer. This is done by typing "L,1000,1369"

NAV.A OPERATING INSTRUCTIONS (continued)

to load the file at Nav computer memory address 1000 (HEX). Next type "T filename". This will place the CRT in terminal mode and create an input buffer to store incoming data in disk file filename. Follow this by typing "control shift " then "control T" and "NAV.HEX" to transmit the program file to the Nav computer. Reply with Yes when asked for time delays. The program data will again be displayed as it is transmitted. If an error is made in STEP 6, the navigation reset button must be pressed and the entire step done over.

STEP 7: Begin program execution. First type "control shift " and then "control Y" to open the input data buffer. Now type "G,1000" to begin program execution. During execution time the robot will send to the H89 two bytes of heading data each time it considers a course change. This data will be displayed on the CRT and stored in the input buffer.

STEP 8: When the robot run is completed (i.e. you have manually stopped it with the MARRS-1 system reset button) the data stored in the input buffer may be written to disk. To do this press the Nav Computer reset button on MARRS-1. Next type "control shift " followed by "control E". Now type "WRT" to save the data to disk ("del" may also be typed to dump buffered data). If additional runs are required continue with STEP 6 and press all three reset buttons on MARRS-1.

STEP 9: Shutdown all systems. To exit M72, type "CPM". Remove both disks from the drives and turn off the power to the H89 system. Turn off power to the robot, GYRAC, and external power supplies.

NOTE: All references to "control" and "shift" in the H89 command lines refer to the control and shift keys and not the words control and shift. Care must be taken to ensure the various cables to MARRS-1 do not become tangled during movement. In addition, it is assumed that the robot has been "pointed" to the desired initial heading before movement commences. The actual direction of travel is set into the NAV.A program at assembly time. Also, the NAV.HEX program must be loaded each time a run is attempted, since the program is cleared on navigation computer reset.

APPENDIX I

CONVERT.BAS Program Listing	I-2
POSITION.BAS Program Listing	I-4

CONVERT.BAS PROGRAM LISTING

```

REM*****
REM*
REM*   DATE: 30SEP85
REM*   VERSION: 1.0
REM*   TITLE: CONVERT
REM*   FILENAME: CONVERT.BAS
REM*   AUTHOR: CAPT ROLAND J. BLOOM
REM*   PROJECT: GYRO AND ACCELEROMETER BASED NAVIGATION
REM*             SYSTEM FOR A MOBILE AUTONOMOUS ROBOT
REM*             (THESIS)
REM*   OPERATING SYSTEM: Z89/Z90 CP/M V2.242 MAGNOLIA
REM*                     MICROSYSTEM 1982
REM*   LANGUAGE: MBASIC
REM*   USE: This program is used to convert raw hex-
REM*        adecimal data obtained from the NAV computer
REM*        into integer data. The whole data string
REM*        is read and converted to integer format.
REM*        The program interactively asks for the name
REM*        of the hex data file and asks for the name
REM*        of the file where the integer data is to be
REM*        stored.
REM*
REM*****
REM*
REM*           MAIN PROGRAM
REM*
REM*****
REM*
10 PRINT "INPUT THE NAME OF THE HEX DATA FILE TO CONVERT"
20 INPUT "INCLUDE THE DISK DRIVE AND ENCLOSE IN QUOTES",READFILE$
30 PRINT " "
40 PRINT "INPUT THE NAME OF THE FILE TO STORE THE INTEGER DATA"
50 INPUT "INCLUDE THE DISK DRIVE AND ENCLOSE IN QUOTES",PRINTFILE$
60 OPEN "I",#1,READFILE$
70 OPEN "O",#2,PRINTFILE$
REM*
REM*   A DATA STRING IS READ AND THEN EACH DATA SEGMENT IS
REM*   CONVERTED AND STORED ON DISK
REM*
80 INPUT#1,DATALINE$
90 IF EOF(1) THEN END
100 WORD$ = MID$(DATALINE$,3,4)
110 GOSUB 300
120 PRINT#2,VALUE%," ";
130 FOR I% = 11 TO 26 STEP 5
140 WORD$ = MID$(DATALINE$,I%,4)
150 GOSUB 300
160 GOSUB 450

```

CONVERT.BAS PROGRAM LISTING (continued)

```

170 NEXT I%
180 FOR I% = 40 TO 47 STEP 7
190 WORD$ = MID$(DATALINE$, I%, 4)
200 GOSUB 300
210 GOSUB 450
220 NEXT I%
230 GOTO 80
REM*****
REM*
REM*          SUBROUTINES    FOLLOW
REM*
REM*****
REM*
REM*  THIS SUBROUTINE CONVERTS THE HEX VALUE TO INTEGER
REM*
300 VALUE% = 0
310 FOR J = 2 TO 4
320 CHAR$ = MID$(WORD$, J%, 1)
330 DIGIT = VAL(CHAR$)
340 IF CHAR$ = "A" THEN DIGIT = 10
350 IF CHAR$ = "B" THEN DIGIT = 11
360 IF CHAR$ = "C" THEN DIGIT = 12
370 IF CHAR$ = "D" THEN DIGIT = 13
380 IF CHAR$ = "E" THEN DIGIT = 14
390 IF CHAR$ = "F" THEN DIGIT = 15
400 VALUE% = VALUE% * 16 + DIGIT
410 NEXT J%
420 RETURN
REM*
REM*  THIS SUBROUTINE STORES THE INTEGER VALUES ON DISK
REM*
450 IF I% = 47 GOTO 480
460 PRINT#2, VALUE%, ", ";
470 GOTO 490
480 PRINT#2, VALUE%
490 RETURN

```

POSITION.BAS PROGRAM LISTING

```

REM*****
REM*
REM*   DATE:  30SEP85
REM*   VERSION: 1.0
REM*   TITLE:  POSITION
REM*   FILENAME: POSITION.BAS
REM*   AUTHOR:  CAPT ROLAND J. BLOOM
REM*   PROJECT: GYRO AND ACCELEROMETER BASED NAVIGATION
REM*             SYSTEM FOR A MOBILE AUTONOMOUS ROBOT
REM*             (THESIS)
REM*   OPERATING SYSTEM:  Z89/Z90 CP/M V2.242 MAGNOLIA
REM*                     MICROSYSTEM  1982
REM*   LANGUAGE:  MBASIC
REM*   USE:  This program is used to compute the position
REM*         of the MARRS-1 robot based on heading and
REM*         velocity data from the GYRAC.  The raw data
REM*         from the NAV computer (which gathers the
REM*         GYRAC data) must first be converted to
REM*         integer format by the CONVERT program.
REM*         The computed position will be in terms of
REM*         x and y coordinates.  An initial (x,y)
REM*         position is provided to the program
REM*         interactively.  Program output is sent
REM*         to the printer where time, x-coordinate
REM*         and y-coordinate are printed.
REM*
REM*****
REM*
REM*               DEFINITION OF VARIABLES
REM*
REM*****
REM*
REM*           X = X-COORDINATE
REM*           Y = Y-COORDINATE
REM*           DELTAX = INCREMENT OF MOVEMENT IN X-DIRECTION
REM*           DELTAY = INCREMENT OF MOVEMENT IN Y-DIRECTION
REM*           DISTANCE = LINEAR DISTANCE TRAVELLED IN T SECONDS
REM*           T = 0.1 SECONDS (WHICH IS THE SAMPLE TIME)
REM*           HEADING = THE HEADING OF THE ROBOT IN DEGREES
REM*           VELOCITY = THE VELOCITY OF THE ROBOT (FT/SEC)
REM*           WEIGHT = WEIGHT OF EACH BIT OF VELOCITY,
REM*                   1024 BITS REPRESENT 10 VOLTS
REM*                   THEREFORE WEIGHT = 0.00977 VOLTS/BIT
REM*           CONV = CONVERSION FACTOR FOR CONVERTING THE
REM*                   VELOCITY MEASUREMENT FROM VOLTS TO
REM*                   FT/SEC.  THIS VALUE IS BASED ON LOCAL
REM*                   ACCELERATION DUE TO GRAVITY OF 32.174
REM*                   FT/S/S/G AND THE SENSITIVITY OF THE
REM*                   ACCELEROMETER (VOLTS/G).

```

POSITION.BAS PROGRAM LISTING (continued)

```

REM*          GF = GAIN FACTOR. THIS IS THE GAIN IN THE      *
REM*          INTEGRATOR CIRCUIT.                            *
REM*          TIME = TIME OF MEASUREMENT (SECONDS)           *
REM*          RAWTIME = INTEGER VALUE OF TIME. THIS VALUE IS  *
REM*          A FACTOR OF 10 TIMES THE REAL TIME.            *
REM*          RAWVEL = INTEGER VALUE FOR VELOCITY (BITS)      *
REM*          RAWHEAD = INTEGER VALUE FOR HEADING (BITS)      *
REM*          LEFTREV = REVERSE WHEEL COUNTS FROM OPTICAL     *
REM*          SHAFT ENCODER ON LEFT REAR WHEEL.              *
REM*          LEFTFOR = FORWARD WHEEL COUNTS FROM OPTICAL     *
REM*          SHAFT ENCODER ON LEFT REAR WHEEL.              *
REM*          RIGHTREV = REVERSE WHEEL COUNTS FROM OPTICAL    *
REM*          SHAFT ENCODER ON RIGHT REAR WHEEL.             *
REM*          RIGHTFOR = FORWARD WHEEL COUNTS FROM OPTICAL    *
REM*          SHAFT ENCODER ON RIGHT REAR WHEEL.             *
REM*
REM*          NOTE - WHEEL COUNTS ARE NOT USED BY THIS        *
REM*          PROGRAM BUT COULD BE INCORPORATED              *
REM*          TO PROVIDE A SEPARATE POSITION                    *
REM*          CALCULATION.                                    *
REM*
REM*****
REM*
REM*          MAIN PROGRAM FOLLOWS
REM*
REM*****
10 INPUT "INPUT THE NAME OF THE DATA FILE(INCLUDE DISK DRIVE)",
    FILE$
20 OPEN "I",#1,FILE$
30 WEIGHT = 0.00977
33 CONV = 32.174/0.6
37 GF = 1/19.7
39 T = 0.1
40 INPUT "INPUT THE INITIAL POSITION (X,Y) IN FEET,XO,YO
50 X = XO
60 Y = YO
70 INPUT "INPUT THE TEST DESIGNATION",TEST$
80 LPRINT TEST$
90 LPRINT " "
100 LPRINT "          POSITION"
110 LPRINT " TIME(SEC)          X(FT)          Y(FT)"
120 LPRINT "*****          *****          *****"
130 LPRINT " "
140 INPUT#1,RAWTIME,LEFTREV,LEFTFOR,RIGHTREV,RIGHTFOR,RAWVEL,
    RAWHEAD
150 IF EOF(1) THEN END

```

POSITION.BAS PROGRAM LISTING (continued)

```
160 TIME = RAWTIME * 0.1
165 PRINT TIME
170 VELOCITY = (RAWVEL - 512) * WEIGHT * CONV * GF
180 HEADING = RAWHEAD * 0.001534
190 DISTANCE = VELOCITY * T
200 DELTAX = DISTANCE * COS(HEADING)
210 DELTAY = DISTANCE * SIN(HEADING)
220 X = X + DELTAX
230 Y = Y + DELTAY
240 LPRINT USING "    ###.##    ##.##    ##.##";TIME,X,Y
250 GOTO 140
```

APPENDIX J

Phase II Sample Test Data..... J-2

GYRAC Phase II Sample Test Data

TEST #2B (SENSITIVITY = 0.393)

TIME(SEC)	POSITION	
	X(FT)	Y(FT)
0.0	6.75	4.58
0.1	6.75	4.58
0.2	6.75	4.58
0.3	6.75	4.59
0.4	6.75	4.59
0.5	6.75	4.59
0.6	6.75	4.59
0.7	6.75	4.59
0.8	6.75	4.60
0.9	6.75	4.60
1.0	6.75	4.60
1.1	6.75	4.60
1.2	6.75	4.60
1.3	6.75	4.61
1.4	6.75	4.61
1.5	6.75	4.61
1.6	6.75	4.61
1.7	6.75	4.62
1.8	6.75	4.62
1.9	6.75	4.62
2.0	6.75	4.64
2.1	6.75	4.65
2.2	6.75	4.66
2.3	6.74	4.67
2.4	6.74	4.68
2.5	6.74	4.69
2.6	6.74	4.70
2.7	6.74	4.70
2.8	6.74	4.71
2.9	6.74	4.72
3.0	6.74	4.73
3.1	6.74	4.74
3.2	6.74	4.75
3.3	6.74	4.76
3.4	6.74	4.77
3.5	6.74	4.78
3.6	6.74	4.79
3.7	6.73	4.80
3.8	6.73	4.81
3.9	6.73	4.82
4.0	6.73	4.83
4.1	6.73	4.84
4.2	6.73	4.85

APPENDIX K

Phase III Test Data

Gyro Navigation Test Run Number 1	K-2
Gyro Navigation Test Run Number 2	K-3
Gyro Navigation Test Run Number 3	K-4

GYRO NAVIGATION TEST RUN NUMBER 1

Given: 33 Foot Course
 Heading of 3EC (hex) = 1004 (integer) = 88.2421875
 (degrees)

Steering Window: 3E0 to 3EF (Hex)
 992 to 1007 (integer)
 87.1875 to 88.50585938 (Degrees)

Measured: Heading at each course change decision point

HEADING (Hex)	HEADING (Integer)	HEADING (Degrees)	DEVIATION (Integer)	DEVIATION (Degrees)
3EB	1003	88.15429688	- 1	-0.087890625
3EB	1003	88.15429688	- 1	-0.087890625
3EB	1003	88.15429688	- 1	-0.087890625
401	1025	90.08789063	+21	+1.845703125
410	1040	91.40625	+36	+3.1640625
40F	1039	91.31835938	+35	+3.076171875
3FF	1023	89.91210938	+19	+1.669921875
3D8	984	86.484375	-20	-1.7578125
3B0	944	82.96875	-60	-5.2734375
3A2	930	81.73828125	-74	-6.50390625
3A2	930	81.73828125	-74	-6.50390625
3B8	952	83.671875	-52	-4.5703125
3E0	992	87.1875	-12	-1.0546875
3FF	1023	89.91210938	+19	+1.669921875
40D	1037	91.14257813	+33	+2.900390625
410	1040	91.40625	+36	+3.1640625
3FC	1020	89.6484375	+16	+1.40625
3DB	987	86.74804688	-17	-1.494140625
3B9	953	83.75976563	-51	-4.482421875
3AE	942	82.79296875	-62	-5.44921875
3AA	938	82.44140625	-66	-5.80078125
3BA	954	83.84765625	-50	-4.39453125
3E2	994	87.36328125	-10	-0.87890625
3FF	1023	89.91210938	+19	+1.669921875
40A	1034	90.87890625	+30	+2.63671875

GYRO NAVIGATION TEST RUN NUMBER 2

Given: 33 Foot Course
Heading of 3EC (hex) = 1004 (integer) = 88.2421875
(degrees)

Steering Window: 3E0 to 3EF (Hex)
992 to 1007 (Integer)
87.1875 to 88.50585938 (Degrees)

Measured: Heading at each course change decision point

HEADING (Hex)	HEADING (Integer)	HEADING (Degrees)	DEVIATION (Integer)	DEVIATION (Degrees)
2F5	757	66.53320313	-247	-21.70898438
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3EB	1003	88.15429688	- 1	- 0.087890625
3E7	999	87.80273438	- 5	- 0.439454125
3E7	999	87.80273438	- 5	- 0.439454125
3E7	999	87.80273438	- 5	- 0.439454125
3E4	996	87.5390625	- 8	- 0.703125
3E1	993	87.27539063	- 11	- 0.966796875
3DE	990	87.01171875	- 14	- 1.23046875
3E2	994	87.36328125	- 10	- 0.87890625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E8	1000	87.890625	- 4	- 0.3515625
3E7	999	87.80273438	- 5	- 0.439454125
3E7	999	87.80273438	- 5	- 0.439454125

GYRO NAVIGATION TEST RUN NUMBER 3

Given: 33 Foot Course
 Heading of 3EC (hex) = 1004 (integer) = 88.2421875
 (degrees)

Steering Window: 3e0 to 3EF (Hex)
 992 to 1007 (Integer)
 87.1875 to 88.50585938 (Degrees)

Measured: Heading at each course change decision point

HEADING (Hex)	HeADING (Integer)	HEADING (Degrees)	DEVIATION (Integer)	DEVIATION (Degrees)
502	1282	112.6757813	+278	+24.43359375
3E4	996	87.5390625	- 8	- 0.703125
3E1	993	87.27539063	- 11	- 0.966796875
3E1	993	87.27539063	- 11	- 0.966796875
3E5	997	87.62695313	- 7	- 0.615234375
3E8	1000	87.890625	- 4	- 0.3515625
3EE	1006	88.41796875	+ 2	+ 0.17578125
3F4	1012	88.9453125	+ 8	+ 0.703125
3F9	1017	89.38476563	+ 13	+ 1.142578125
3E1	993	87.27539063	- 11	- 0.966796875
3DB	989	86.92382813	- 15	- 1.318359375
3F1	1009	88.68164063	+ 5	+ 0.439453125
3F9	1017	89.38476563	+ 13	+ 1.142578125
3FB	1019	89.56054688	+ 15	+ 1.318359375
3E7	999	87.80273438	- 5	- 0.439454125
3DE	990	87.01171875	- 14	- 1.23046875
3F1	1009	88.68164063	+ 5	+ 0.439453125
3FC	1020	89.6484375	+ 16	+ 1.40625
3FF	1023	89.91210938	+ 19	+ 1.669921875
3EA	1002	88.06640625	- 2	- 0.017578125
3E1	993	87.27539063	- 11	- 0.966796875
3E8	1000	87.890625	- 4	- 0.3515625
3EB	1003	88.15429688	- 1	- 0.087890625
3F1	1009	88.68164063	+ 5	+ 0.439453125
3F4	1012	88.9453125	+ 8	+ 0.703125
3DE	990	87.01171875	- 14	- 1.23046875
3D2	977	85.86914063	- 27	- 2.373046875

APPENDIX L

Lab Equipment, Computer Hardware, and Software L-2

LAB EQUIPMENT, COMPUTER, HARDWARE, and SOFTWARE

LAB EQUIPMENT

Quantity	Description
1	AFIT Mobile Robotics Laboratory
1	AFIT MARRS-1 Robot
1	1607 Eldorado Frequency Counter
1	186 Wavetek Waveform Generator
1	1610A Hewlitt Packard Logic State Analyzer
1	465M Tektronics Oscilloscope
1	3466A Hewlitt Packard Digital Multimeter
1	M-15 Trygon Power Supply
1	M-36 Trygon Power Supply
1	6C3000 Powertec Power Supply
1	S-10 Bytek EEPROM Programmer
1	S-52 Ultra Violet Products EEPROM Eraser

COMPUTER HARDWARE

Quantity	Description
1	MARRS-1 Navigation Computer
1	MARRS-1 Drive Computer
1	MARRS-1 GYRAC Computer
1	Heath H89 Computer
1	Heath H27 Eight Inch Dual Disk Drive System
1	Heath H125 Dot Matrix Printer

LAB EQUIPMENT, COMPUTER, HARDWARE, and SOFTWARE (continued)

COMPUTER SOFTWARE

Quantity	Description
1	Wordmaster Word Processor
1	Wordstar Word Processor
1	Virtual Devices Robo A 6802 Cross Assembler
1	Modem 720 Communication Program
1	CP/M Operating System
1	MBASIC Compiler
1	MARRS-1 Drive Computer ROM Software
1	MARRS-1 Navigation Computer ROM Software

AD A164036

REPORT DOCUMENTATION PAGE

1a. REPORT SECURITY CLASSIFICATION Unclassified			1b. RESTRICTIVE MARKINGS		
2a. SECURITY CLASSIFICATION AUTHORITY			3. DISTRIBUTION/AVAILABILITY OF REPORT Approved for Public Release; Distribution Unlimited		
2b. DECLASSIFICATION/DOWNGRADING SCHEDULE					
4. PERFORMING ORGANIZATION REPORT NUMBER(S) AFIT/GA/GE/ENG/85D-33			5. MONITORING ORGANIZATION REPORT NUMBER(S)		
6a. NAME OF PERFORMING ORGANIZATION Air Force Institute of Technology		6b. OFFICE SYMBOL (If applicable) EN		7a. NAME OF MONITORING ORGANIZATION	
6c. ADDRESS (City, State and ZIP Code) Wright-Patterson AFB, OH 45433				7b. ADDRESS (City, State and ZIP Code)	
8a. NAME OF FUNDING/SPONSORING ORGANIZATION		8b. OFFICE SYMBOL (If applicable)		9. PROCUREMENT INSTRUMENT IDENTIFICATION NUMBER	
8c. ADDRESS (City, State and ZIP Code)				10. SOURCE OF FUNDING NOS.	
				PROGRAM ELEMENT NO.	
11. TITLE (Include Security Classification) on back				PROJECT NO.	
				TASK NO.	
12. PERSONAL AUTHOR(S) Bloom, Roland J. and Ramey, William J. Jr.				WORK UNIT NO.	
				10. SOURCE OF FUNDING NOS.	
13a. TYPE OF REPORT Final		13b. TIME COVERED FROM 85 Jan to 85 Dec		14. DATE OF REPORT (Yr., Mo., Day) 1985/12/2	
15. PAGE COUNT 251		16. SUPPLEMENTARY NOTATION			
17. COSATI CODES		18. SUBJECT TERMS (Continue on reverse if necessary)			
FIELD		GROUP		SUB. GR.	
17		07			
19. ABSTRACT (Continue on reverse if necessary and identify by block number)					
<p>A navigation system for a mobile autonomous robot is presented. The navigation system is based upon a directional gyroscope and a single axis accelerometer which enables a robot to navigate independent of wheel optical shaft encoders and other commonly used positioning apparatus. The computer controlled navigation system is capable of providing absolute heading, heading rate (angular velocity), and linear velocity to a user computer. These data from the navigation system (heading and velocity) are used to compute the present location of the robot. In addition, the heading data is used to form a closed loop feedback control system for maintaining the robot on a desired course. The navigation system was designed</p>					
20. DISTRIBUTION/AVAILABILITY OF ABSTRACT UNCLASSIFIED/UNLIMITED <input checked="" type="checkbox"/> SAME AS RPT. <input type="checkbox"/> DTIC USERS <input type="checkbox"/>			21. ABSTRACT SECURITY CLASSIFICATION Unclassified		
22a. NAME OF RESPONSIBLE INDIVIDUAL Mathew Kabrisky		22b. TELEPHONE NUMBER (Include Area Code) (513) 255-5276		22c. OFFICE SYMBOL AFIT/EN	

Block 11. Gyro and Accelerometer Based Navigation System for a Mobile Autonomous Robot

Block 19. (continued) specifically for application on an existing Air Force Institute of Technology (AFIT) robot; however, it could be easily adapted to any robot system with a standard IEEE RS-232 serial communication interface. Test results are provided which demonstrate the use of closed loop heading control on the AFIT robot and which identify problems associated with the use of an accelerometer system for distance measurement. This thesis includes all schematics, parts lists, software listings, and operating instructions for the navigation system. A new robot world modeling and path planning technique is also presented. (746000) F

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